

# **GATE** 2025

# Electronics Engineering

## **Questions & Solutions**

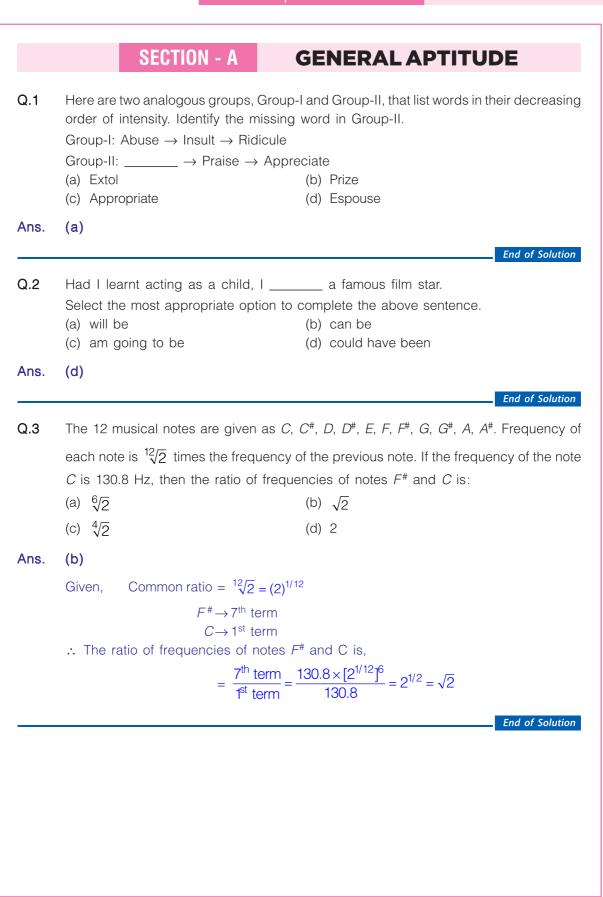
Exam held on 15/02/2025 (Afternoon Session)

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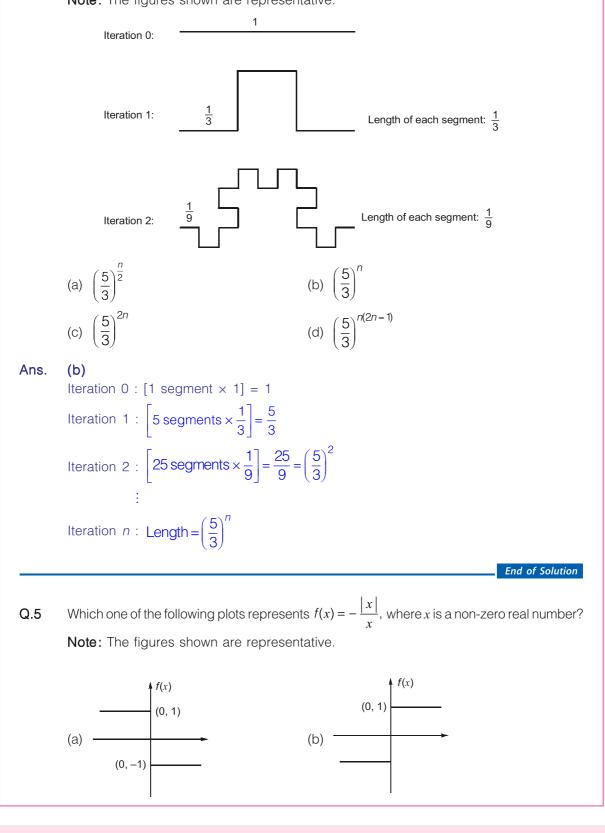








Q.4 The following figures show three curves generated using an iterative algorithm. The total length of the curve generated after 'Iteration *n*' is:Note: The figures shown are representative.





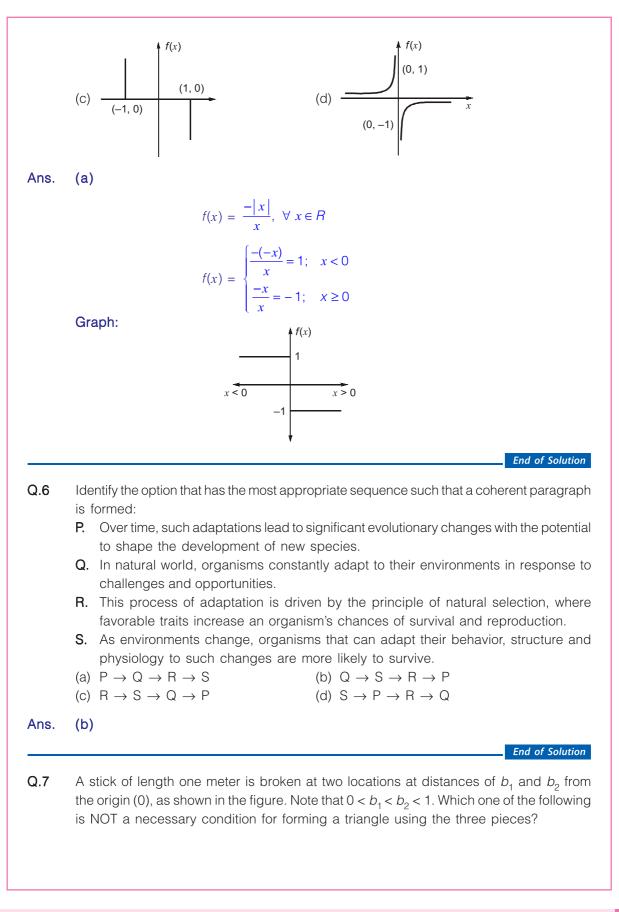
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by





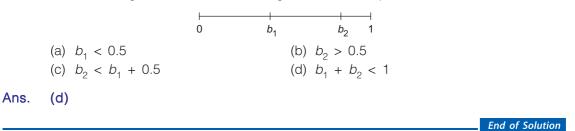








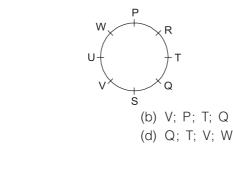
Note: All lengths are in meter. The figure shown is representative.



**Q.8** Eight students (P, Q, R, S, T, U, V, and W) are playing musical chairs. The figure indicates their order of position at the start of the game. They play the game by moving forward in a circle in the clockwise direction.

After the 1<sup>st</sup> round, 4<sup>th</sup> student behind P leaves the game. After 2<sup>nd</sup> round, 5<sup>th</sup> student behind Q leaves the game. After 3<sup>rd</sup> round, 3<sup>rd</sup> student behind V leaves the game. After 4<sup>th</sup> round, 4<sup>th</sup> student behind U leaves the game. Who all are left in the game after the 4<sup>th</sup> round?

Note: The figure shown is representative.



Ans. (d)

According to GATE key.

(a) P; T; Q; S

(c) W; R; Q; V

End of Solution

Q.9 The table lists the top 5 nations according to the number of gold medals won in a tournament; also included are the number of silver and the bronze medals won by them. Based only on the data provided in the table, which one of the following statements is INCORRECT?

Nation	Gold	Silver	Bronze	
USA	40	44	41	
Canada	39	27	24	
Japan	20	12	13	
Australia	17	19	16	
France	16	26	22	

- (a) France will occupy the third place if the list were made on the basis of the total number of medals won.
- (b) The order of the top two nations will not change even if the list is made on the basis of the total number of medals won.





- (c) USA and Canada together have less than 50% of the medals awarded to the nations in the above table.
- (d) Canada has won twice as many total medals as Japan.

#### Ans. (c)

Nation	Gold	Silver	Bronze	
USA	40 44		41	
Canada	39	27	24	
Japan	20	12	13	
Australia	17	19	16	
France	16	26	22	

Total medals awarded to the nations = 376Medals awarded to USA = 125Medals awarded to Canada = 90

Medals awarded to USA and Canada = 90 + 125 = 215

$$\therefore \qquad \frac{215}{376} \times 100 = 57.18\%$$

 $\Rightarrow$  USA and Canada together have 57.18% of the total medals awarded to the nations.

End of Solution

Q.10 An organization allows its employees to work independently on consultancy projects but charges an overhead on the consulting fee. The overhead is 20% of the consulting fee, if the fee is up to . 5,00,000. For higher fees, the overhead is . 1,00,000 plus 10% of the amount by which the fee exceeds . 5,00,000. The government charges a Goods and Services Tax of 18% on the total amount (the consulting fee plus the overhead). An employee of the organization charges this entire amount, i.e., the consulting fee, overhead, and tax, to the client. If the client cannot pay more than . 10,00,000, what is the maximum consulting fee that the employee can charge?

(a)	7,01,438	(b)	7,24,961
(C)	7,51,232	(d)	7,75,784

#### Ans. (b)

Let, consultation fee = X

#### Case (i):

 $\Rightarrow$ 

If  $X \le 5.00.000$ 

Overhead = 20% of X = 0.2XTotal cost = X + 0.2X = 1.2X

Tax = 18% of 
$$1.2X = \left(\frac{18}{100}\right) 1.2X = 0.216X$$

Total amount paid by client = 1.2X + 0.216X = 1.416XGiven that, the client can only pay 10,00,000. 1.416 X = 10,00,000

 $\Rightarrow$ 

X = 7,06,215

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Detailed Solutions Exam held on: 15-02-2025

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Case (ii): If X > 500000 Overhead = 1,00,000 + 10%[X - 500000] = 1,00,000 + 0.1[X - 500000] Total cost = X + 100000 + 0.1[X - 500000] = 1.1X + 50000 Tax =  $\frac{18}{100}(1.1X + 50000) = 0.198X + 9000$ Total amount paid by client = 1.1X + 50000 + 0.198X + 9000 = 1.298X + 59000 Given that, the client can only pay 1000000 ⇒ 1.298X + 59000 = 1000000 X = 724,961 ∴ Maximum consultation fee that the client can afford = 724961.

End of Solution



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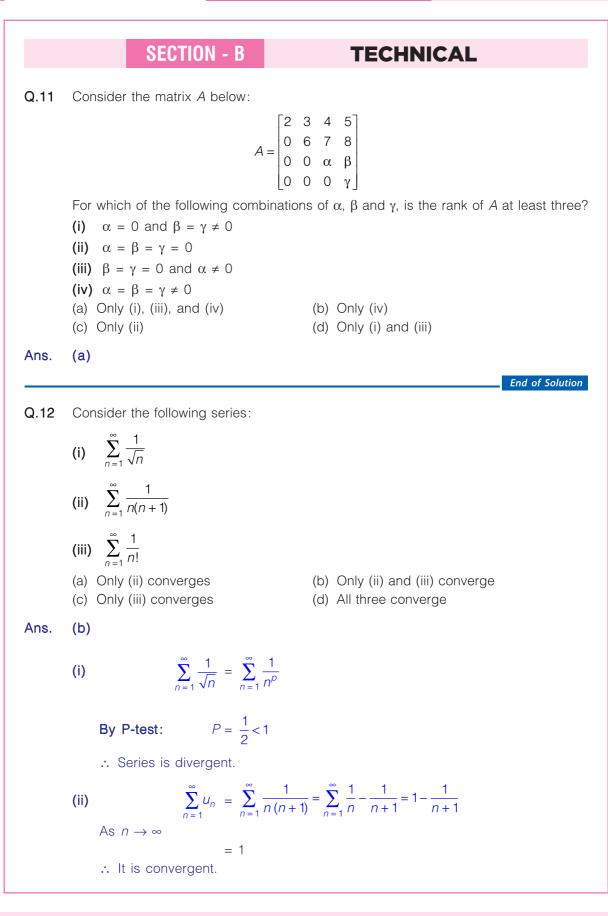
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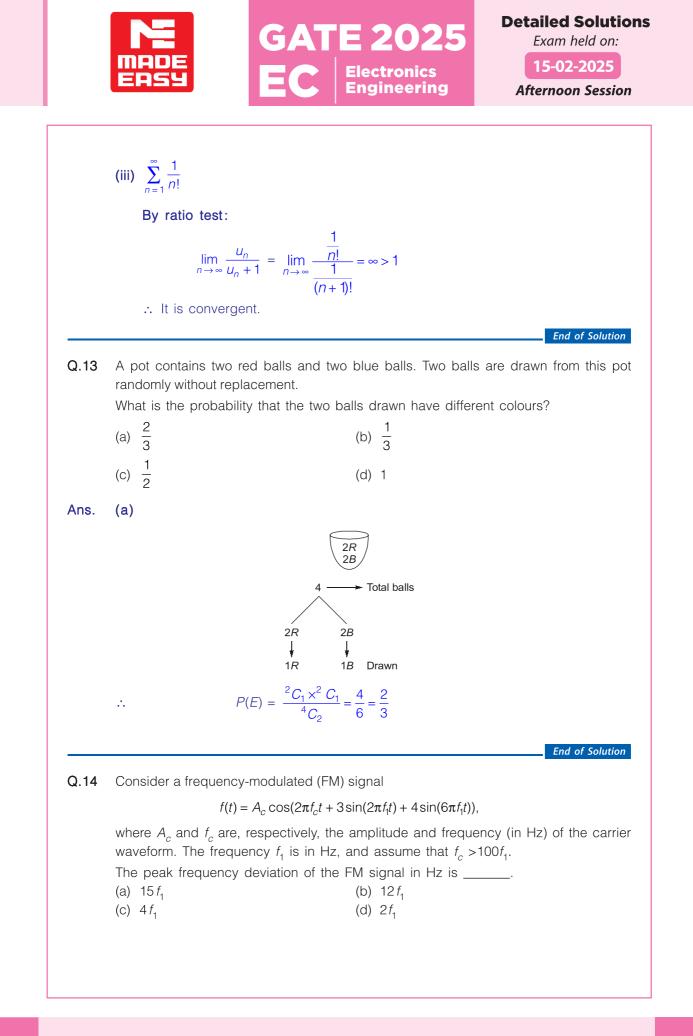
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Ans.

(a)

 $f(t) = A_{c} \cos[2\pi f_{c}t + 3\sin(2\pi f_{1}t) + 4\sin(6\pi f_{1}t)]$ 

General expression of angle modulated signal,

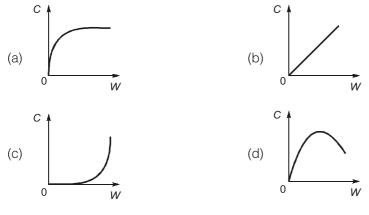
 $s(t) = A_c \cos[2\pi f_c t + \phi(t)]$ 

From comparison:  $\phi(t) = 3\sin 2\pi f_1 t + 4\sin 6\pi f_1 t$ Peak frequency deviation:

$$\Delta f = \left| \frac{1}{2\pi} \frac{d}{dt} \phi(t) \right|_{\max}$$
  
=  $\left| \frac{1}{2\pi} [3 \times 2\pi f_1 \cos 2\pi f_1 t + 4 \times 6\pi f_1 \cos 6\pi f_1 t] \right|_{\max}$   
=  $\left| 3f_1 \cos 2\pi f_1 t + 12f_1 \cos 6\pi f_1 t \right|_{\max} = 3f_1 + 12f_1 = 15f_1$ 

End of Solution

**Q.15** Consider an additive white Gaussian noise (AWGN) channel with bandwidth *W* and noise power spectral density  $\frac{N_o}{2}$ . Let  $P_{av}$  denote the average transmit power constraint. Which one of the following plots illustrates the dependence of the channel capacity *C* on the bandwidth *W* (keeping  $P_{av}$  and  $N_0$  fixed)?



Ans. (a)

We know that,  $C = B \log_2 \left( 1 + \frac{S}{N} \right)$ Given average transmitted power =  $P_{av}$  and channel bandwidth = W.

$$C = W \log_2 \left( 1 + \frac{P_{av}}{N_o W} \right)$$

Given  $P_{av}$  and  $N_o$  are fixed.

Let,

 $\frac{P_{\text{av}}}{N_o} = K \text{ (fixed)}$  $C = W \log_2 \left( 1 + \frac{K}{W} \right)$ 

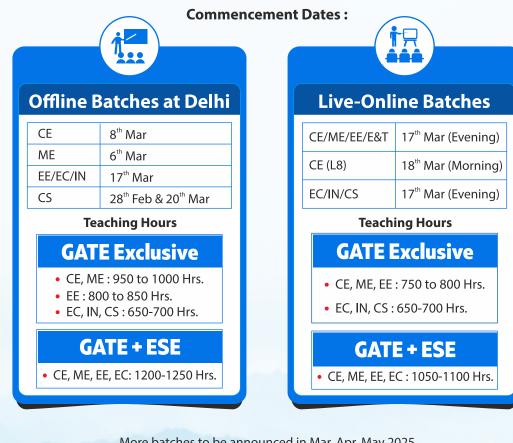


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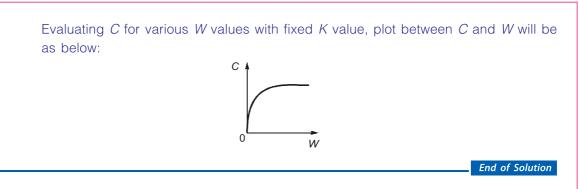
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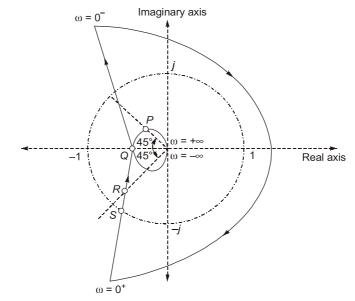
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**Q.16** The Nyquist plot of a system is given in the figure below. Let  $\omega_{P}$ ,  $\omega_{Q}$ ,  $\omega_{R}$ , and  $\omega_{S}$  be the positive frequencies at the points *P*, *Q*, *R*, and *S*, respectively. Which one of the following statements is TRUE?



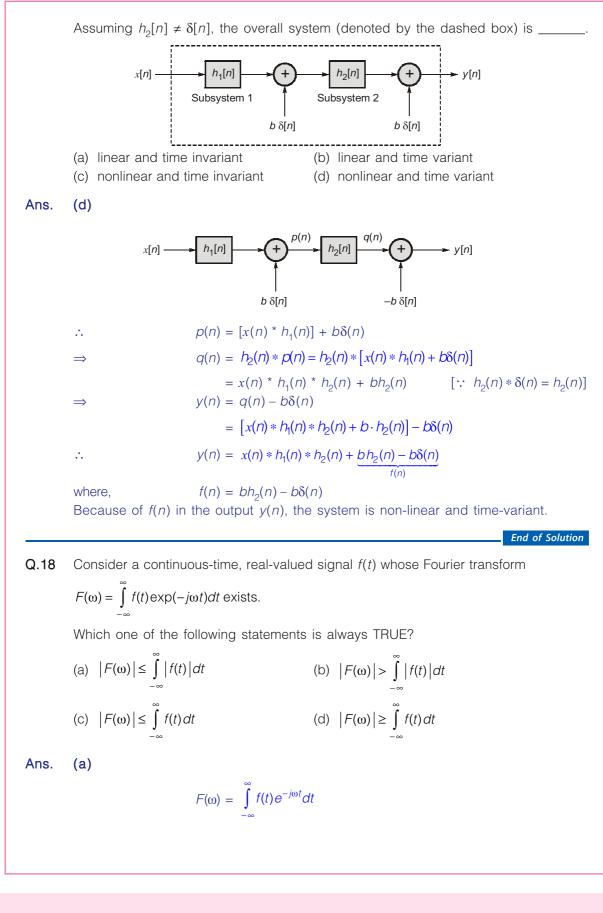
(a)  $\omega_S$  is the gain crossover frequency and  $\omega_P$  is the phase crossover frequency (b)  $\omega_Q$  is the gain crossover frequency and  $\omega_R$  is the phase crossover frequency (c)  $\omega_Q$  is the gain crossover frequency and  $\omega_S$  is the phase crossover frequency (d)  $\omega_S$  is the gain crossover frequency and  $\omega_Q$  is the phase crossover frequency

#### Ans. (d)

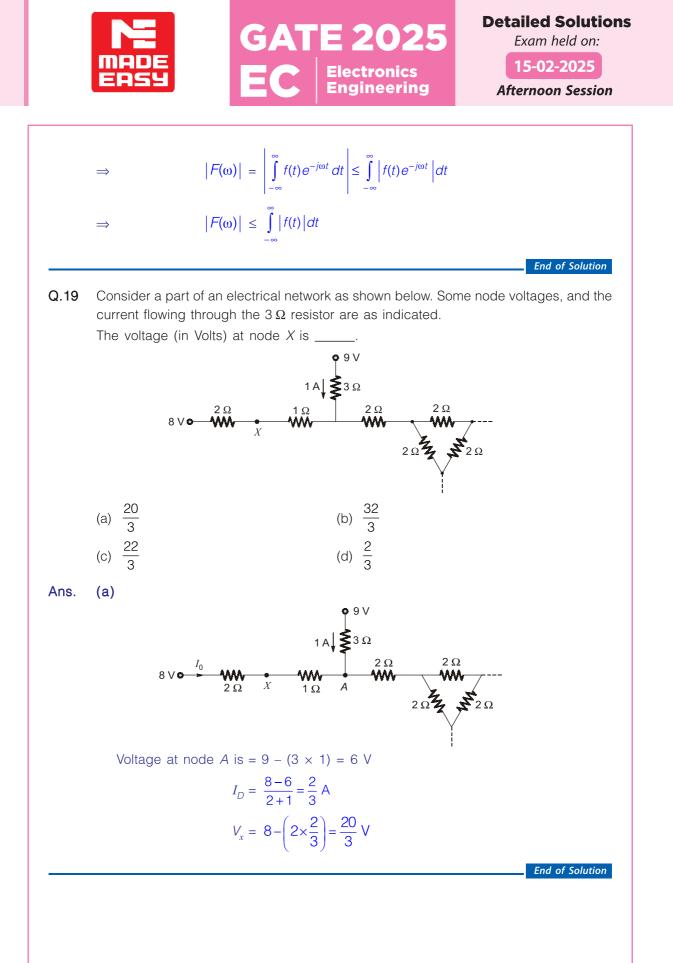
End of Solution

**Q.17** Consider the discrete-time system below with input x[n] and output y[n]. In the figure,  $h_1[n]$  and  $h_2[n]$  denote the impulse responses of LTI Subsystems 1 and 2, respectively. Also,  $\delta[n]$  is the unit impulse, and b > 0.





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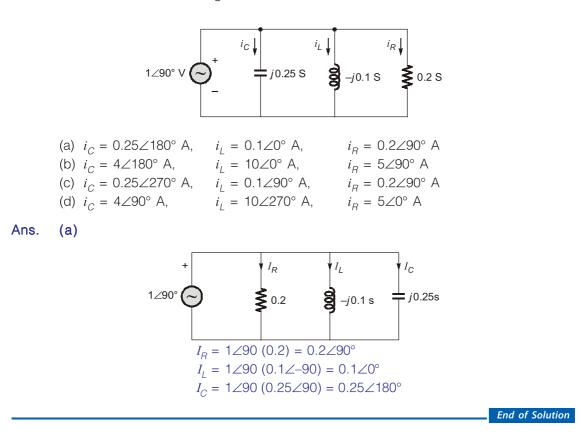
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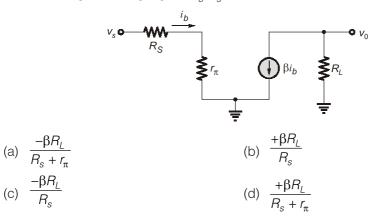


Let  $i_{C}$ ,  $i_{L}$ , and  $i_{R}$  be the currents flowing through the capacitor, inductor, and resistor, Q.20 respectively, in the circuit given below. The AC admittances are given in Siemens(S). Which one of the following is true?

Electronics



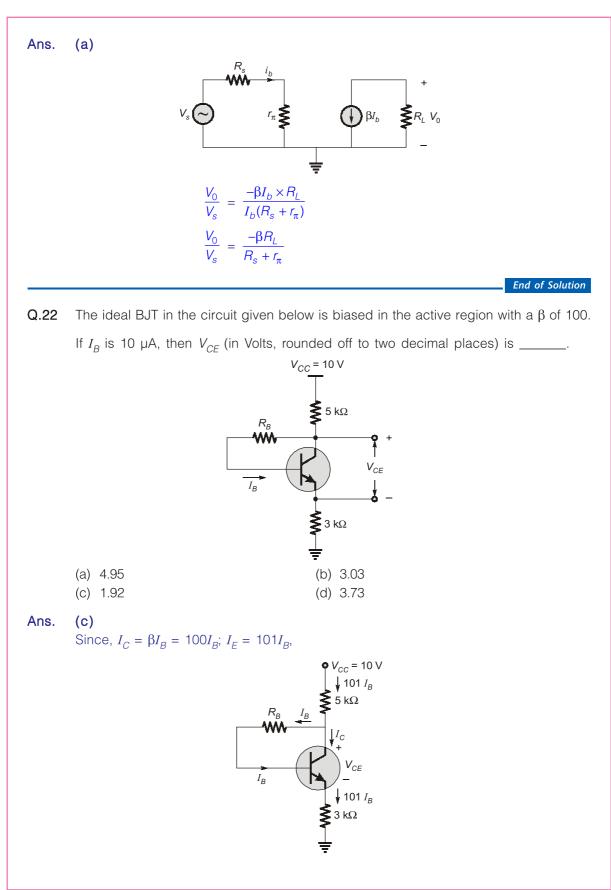
Q.21 A simplified small-signal equivalent circuit of a BJT-based amplifier is given below. The small-signal voltage gain  $V_o/V_s$  (in V/V) is \_\_\_\_\_

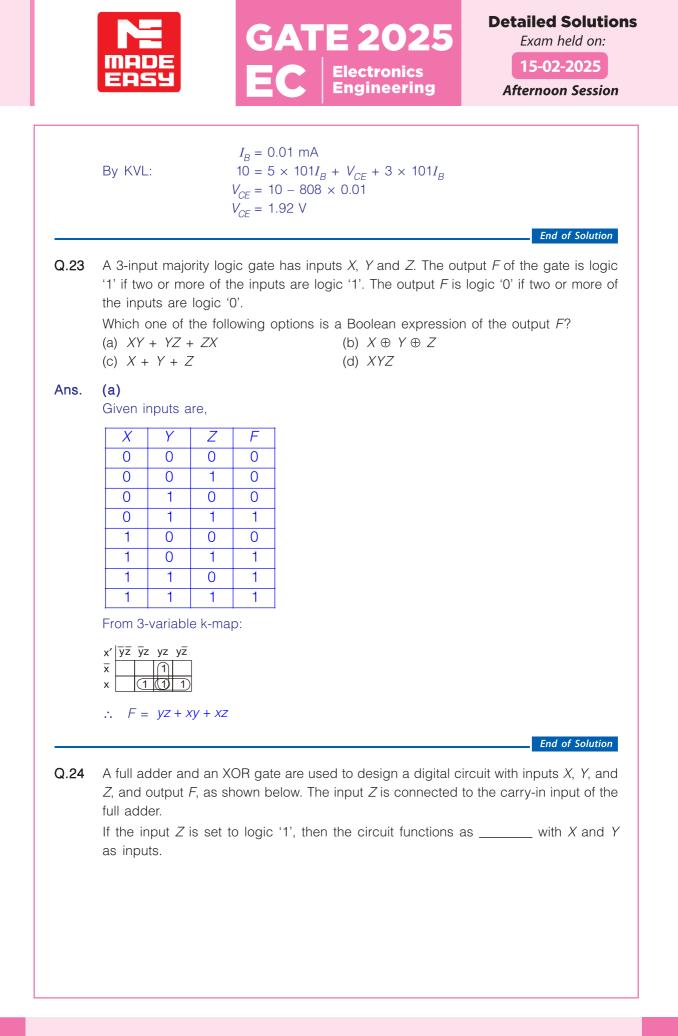


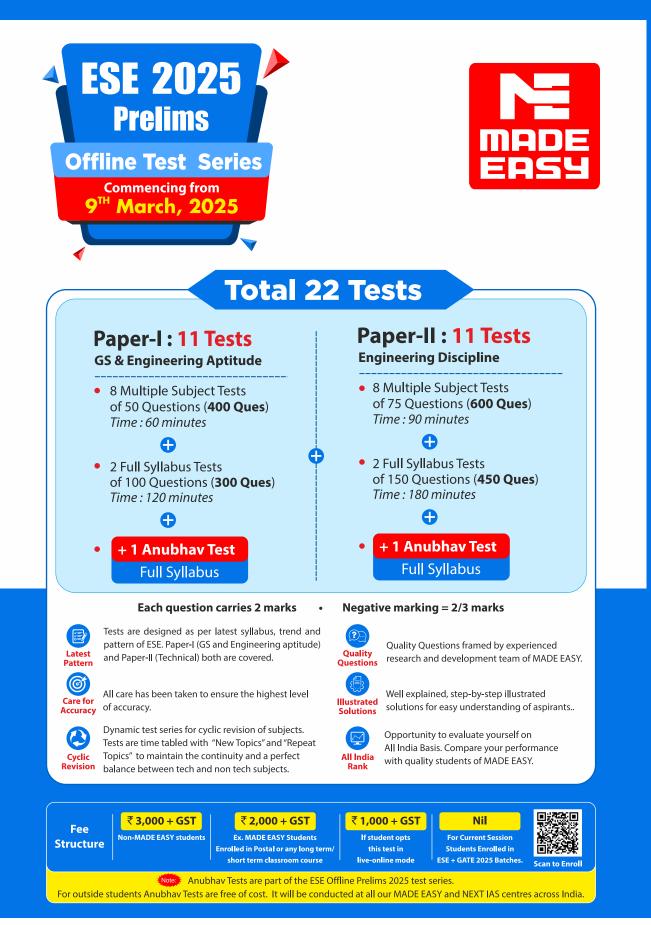
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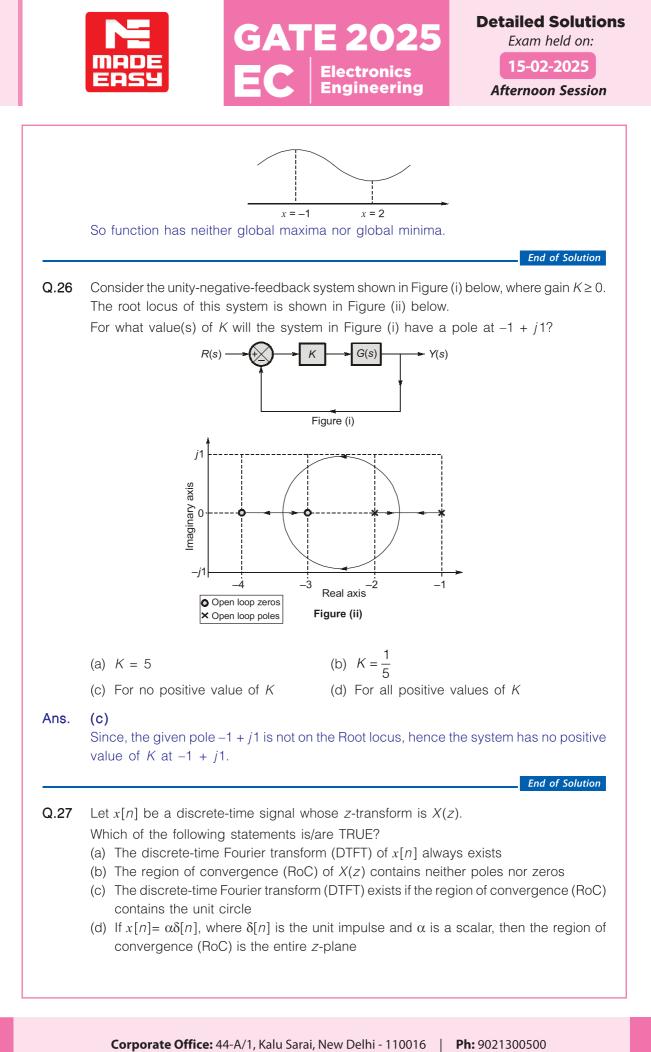


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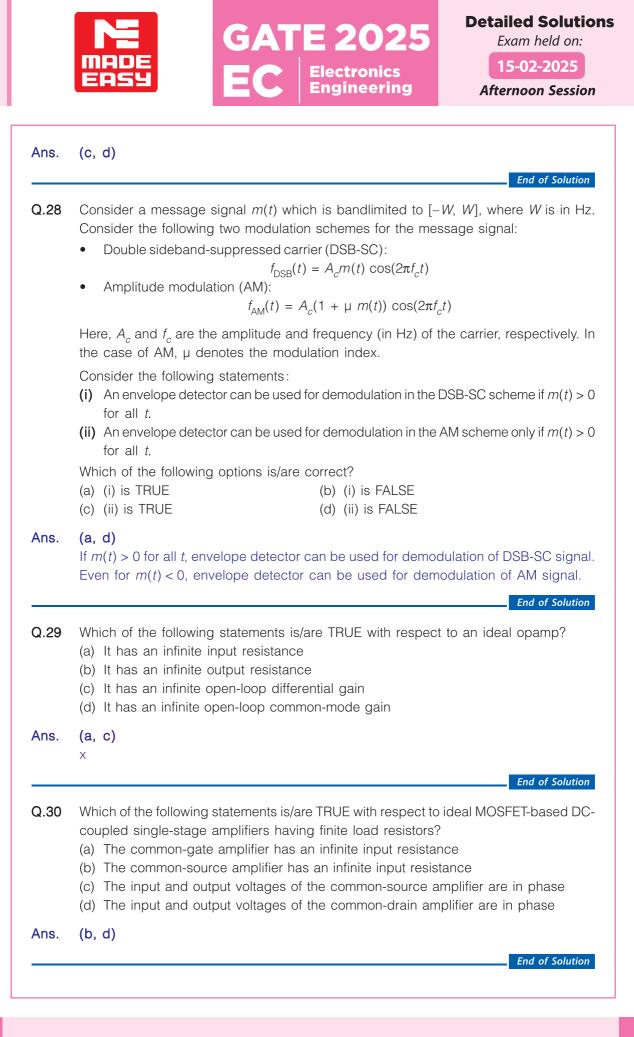


Full adder †2 (a) an adder (b) a subtractor (c) a multiplier (d) a binary to Gray code converter Ans. (b) Х End of Solution Consider the function  $f: \mathbb{R} \to \mathbb{R}$  defined as Q.25  $f(x) = 2x^3 - 3x^2 - 12x + 1$ Which of the following statements is/are correct? (Here,  $\mathbb{R}$  is the set of real numbers.) (a) *f* has no global maximizer (b) f has no global minimizer (c) x = -1 is a local minimizer of f (d) x = 2 is a local maximizer of f Ans. (a, b)  $f(x) = 2x^3 - 3x^2 - 12x + 1$  $f'(x) = 6x^2 - 6x - 12$ f'(x) = 0 $6x^2 - 6x - 12 = 0$  $6(x^2 - x - 2) = 0$ 6(x - 2)(x + 1) = 0x = 2x = -1f''(x) = 12x - 6 $f''(x)\Big|_{x=2} = 12 \times 2 - 6$ = 18 > 0  $x = 2 \implies$  local minima  $f''(x)|_{x=-1} = 12 \times -1 - 6$ = -12 - 6 = -18 < 0  $x = -1 \Rightarrow$  local maxima



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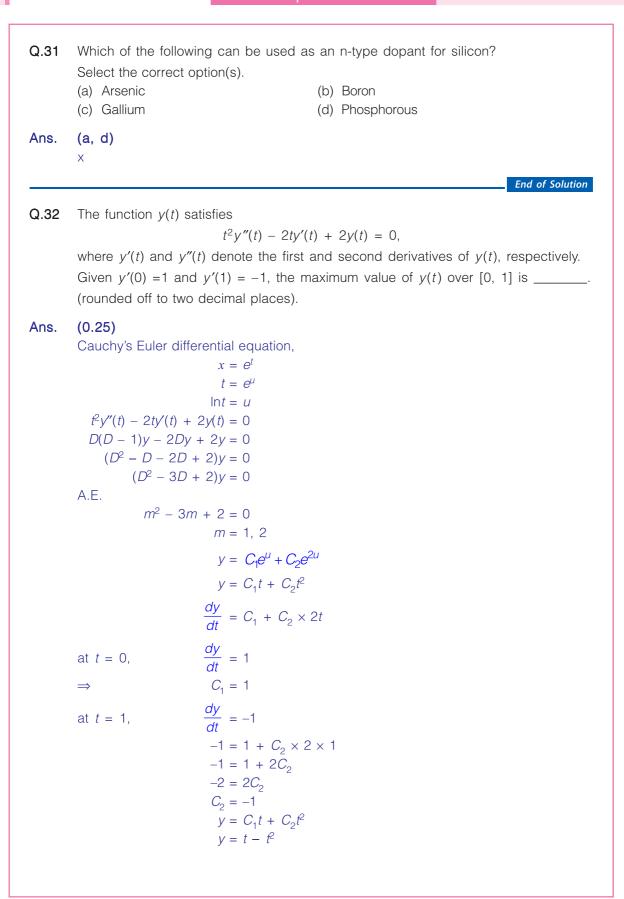
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## **SATE 2025 EC** Electronics Engineering

Detailed Solutions Exam held on: 15-02-2025

Afternoon Session





## GATE 2025 EC Electronics

$$\frac{dy}{dt} = 1 - 2t = 0$$

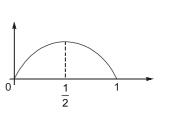
$$t = \frac{1}{2}$$

$$\frac{d^2y}{dt^2} = -2 < 0$$

$$t = \frac{1}{2} \implies \text{maxima point}$$

$$y\left(\frac{1}{2}\right) = \frac{1}{2} - \left(\frac{1}{2}\right)^2$$

$$= \frac{1}{2} - \frac{1}{4} = \frac{1}{4}$$



End of Solution

**Detailed Solutions** 

Exam held on: 15-02-2025

Afternoon Session

Q.33 The generator matrix of a (6,3) binary linear block code is given by

	[1	0	0	1	0	1]
G =	0	1	0	0	1	1
G=	0	0	1	1	1	0

The minimum Hamming distance  $d_{\min}$  between codewords equals \_\_\_\_\_ (answer in integer).

Ans. (3)

$$G = \begin{bmatrix} 1 & 0 & 0 & 1 & 0 & 1 \\ 0 & 1 & 0 & 0 & 1 & 1 \\ 0 & 0 & 1 & 1 & 1 & 0 \end{bmatrix}_{\substack{3 \times 6 \\ K \times n}}^{3 \times 6}$$

#### Method 1:

Since, K = 3,  $2^{K} = 8$  distinct message blocks possible i.e. 000 to 111. Find corresponding codewords  $[c] = [d] \cdot [G]$ . Find Hamming weight of each of the codewords. Smallest possible Hamming weight of non zero codeword equals to  $d_{\min}$ . By evaluating above procedure  $d_{\min} = 3$ .

#### Method 2:

Minimum number of columns of *G* that sum to zero equals to  $d_{min}$ . Sum of 1<sup>st</sup>, 2<sup>nd</sup> and 6<sup>th</sup> columns (or) 4<sup>th</sup>, 5<sup>th</sup> and 6<sup>th</sup> columns (or) 2<sup>nd</sup>, 3<sup>rd</sup> and 5<sup>th</sup> columns equals to zero.

For above G, minimum number of columns of G matrix that sum to zero equals to 3 so that  $d_{\min} = 3$ .

End of Solution

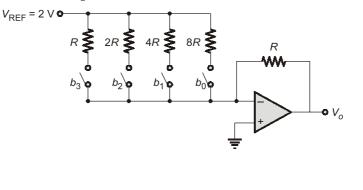


## **SATE 2025 EC** Electronics Engineering

Detailed Solutions Exam held on: 15-02-2025 Afternoon Session

- All the components in the bandpass filter given below are ideal. The lower -3 dB Q.34 frequency of the filter is 1 MHz. The upper -3 dB frequency (in MHz, rounded off to the nearest integer) is \_ 2R ₩₩• 0.1C ┨┠ *R* 10 C *V*<sub>s</sub> **●** - **W** - - **I |** -(50)Ans.  $f_L = \frac{1}{2\pi R_1 C_1}$ We know that,  $10^6 = \frac{1}{2\pi \times R \times 10C}$  $\frac{1}{BC} = 2\pi \times 10^7$  $f_{H} = \frac{1}{2\pi R_{2}C_{2}} = \frac{1}{2\pi \times 2R \times 0.1C}$  $= \frac{1}{0.4\pi} \times \frac{1}{RC} = \frac{2\pi \times 10^7}{0.4\pi}$  $f_{\rm H} = 5 \times 10^7 \, {\rm Hz} = 50 \, {\rm MHz}$ End of Solution
- **Q.35** A 4-bit weighted-resistor DAC with inputs  $b_3$ ,  $b_2$ ,  $b_1$ , and  $b_0$  (MSB to LSB) is designed using an ideal opamp, as shown below. The switches are closed when the corresponding input bits are logic '1' and open otherwise.

When the input  $b_3 b_2 b_1 b_0$  changes from 1110 to 1101, the magnitude of the change in the output voltage  $V_O$  (in mV, rounded off to the nearest integer) is \_\_\_\_\_.





## GATE 2025 EC Electronics Engineering

Detailed Solutions Exam held on: 15-02-2025 Afternoon Session

#### Ans. (250)

For the given 4-bit weighted-resistor DAC with inputs  $b_3$ ,  $b_2$ ,  $b_1$  and  $b_0$  (MSB to LSB), the output voltage is,

 $V_{o} = -R \left[ \frac{b_{0}}{8R} + \frac{b_{1}}{4R} + \frac{b_{2}}{2R} + \frac{b_{3}}{R} \right] V_{\text{REF}}$  $V_{o} = - \left[ \frac{b_{0}}{8} + \frac{b_{1}}{4} + \frac{b_{2}}{2} + \frac{b_{3}}{1} \right] \times 2$ 

 $\Rightarrow$ 

For input 1110:

$$V_{o_1} = -\left[0 + \frac{1}{4} + \frac{1}{2} + \frac{1}{1}\right] \times 2 = -3.5$$

For input 1101:

$$V_{o_2} = -\left[\frac{1}{8} + 0 + \frac{1}{2} + \frac{1}{1}\right] \times 2 = -3.25$$

The magnitude of the change in the output voltage is,

$$V_{o} = |-3.5 - (-3.25)| = 0.25 = 250 \,\mathrm{mV}$$

End of Solution

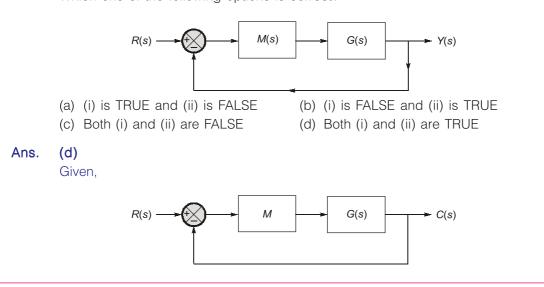
**Q.36** Let  $G(s) = \frac{1}{10s^2}$  be the transfer function of a second-order system. A controller M(s)

is connected to the system G(s) in the configuration shown below. Consider the following statements.

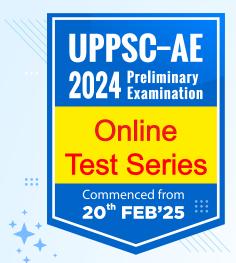
(i) There exists no controller of the form  $M(s) = \frac{K_I}{s}$ , where  $K_I$  is a positive real number,

such that the closed loop system is stable.

(ii) There exists at least one controller of the form M(s) = K<sub>P</sub> + sK<sub>D</sub>, where K<sub>P</sub> and K<sub>D</sub> are positive real numbers, such that the closed loop system is stable.
 Which one of the following options is correct?







### Total 10 Tests (Total 1125 Questions)

5 Part Syllabus Tests + 5 Full Syllabus Tests

#### **Paper Pattern:**

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- → There is a penalty of 0.66 Mark for every wrong answer.

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#### Stream : CE, ME, EE

#### **Test Series Schedule**

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1	20 <sup>th</sup> Feb 2025	75 Qs	1 Hour	Part Syllabus Test	General Principles of Design and Drawing, Industrial Safety and Safety Standards, Engineering Materials, Quality Control, Types of Machinery and Maintenance, Production and Construction, Handling and Storage of Products
2	27 <sup>th</sup> Feb 2025	75 Qs	1 Hour	Part Syllabus Test	Basics of project Management, Information and communication technologies, Ethics and values in engineering profession, intellectual property rights, Role of science and technology in daily life, recent developments in applied sciences, basics of artificial intelligence and robotics
3	6 <sup>th</sup> Mar 2025	75 Qs	1 Hour	Part Syllabus Test	Green Energy, Energy conversion principles, Climate change, Disaster Management, Basics of thermodynamics, Water resources and conservation processes, Basics of measurement and instrumentation, Human health and sanitation
4	13 <sup>th</sup> Mar 2025	75 Qs	1 Hour	Part Syllabus Test	General Hindi
5	20 <sup>th</sup> Mar 2025	75 Qs	1 Hour	Part Syllabus Test	Indian History, Indian Polity, Geography, GK & Miscellaneous and Current Affairs
6	27 <sup>th</sup> Mar 2025	150 Qs	2 Hours	Full Syllabus Test	Full Syllabus Test (100 Qs. Engineering Aptitude + 25 Hindi + 25 General Studies)
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and 
$$G(s) = \frac{1}{s^2}$$
(i) Given,  $M = \frac{K}{s}$ , then closed loop system,  

$$\frac{G(s)}{R(s)} = \frac{MG(s)}{1+MG(s)} = \frac{\frac{K}{s} \times \frac{1}{s^2}}{1+\frac{K}{s} + \frac{1}{s^2}} = \frac{K}{s^3 + K}$$
 $\therefore$  For all values of 'K', the closed loop system is unstable.  
(ii) For controller,  $M = K_p + K_{pS}$   
Let,  $M = 1 + s$   
 $\therefore \qquad \frac{G(s)}{R(s)} = \frac{(1+s)\times\frac{1}{s^2}}{1+\frac{S+1}{s^2}} = \frac{(s+1)}{s^2+s+1}$   
 $\therefore$  Closed loop system is stable.  
 $\therefore$  Both statements are correct.  
**End of Solution**  
**Q.37** Consider the polynomial  $p(s) = s^5 + 7s^4 + 3s^3 - 33s^2 + 2s - 40$ . Let  $(L, I, R)$  be defined as follows.  
 $L$  is the number of roots of  $p(s)$  with negative real parts.  
 $I$  is the number of roots of  $p(s)$  with positive real parts.  
 $I$  is the number of roots of  $p(s)$  that are purely imaginary.  
 $R$  is the number of roots of  $p(s)$  with positive real parts.  
 $I$  is the number of roots of  $p(s)$  with positive real parts.  
 $I$  is the number of roots of  $p(s)$  with  $p(s) = 3, I = 2$  and  $R = 0$   
(c)  $L = 1, I = 2$  and  $R = 1$  (d)  $L = 3, I = 2$  and  $R = 1$   
**Ans.** (a)  
Given  
 $p(s) = s^5 + 7s^4 + 3s^3 - 33s^2 + 2s - 40$   
By Routh's Hurwitz orderia  
 $s^5 = 1 = 3 = 2$  (d)  $L = 0, I = 4$  and  $R = 1$   
**Ans.** (a)  
 $G(s) = 0, S = 5^5 + 7s^4 + 3s^3 - 33s^2 + 2s - 40$   
By Routh's Hurwitz orderia  
 $s^5 = 1 = 3 = 2$  (d)  $L = 0, I = 4$  and  $R = 1$   
**Auxiliary equation is**  
 $A(s) = -40s^2 - 40$   
 $\frac{dA(s)}{ds} = -80s$ 

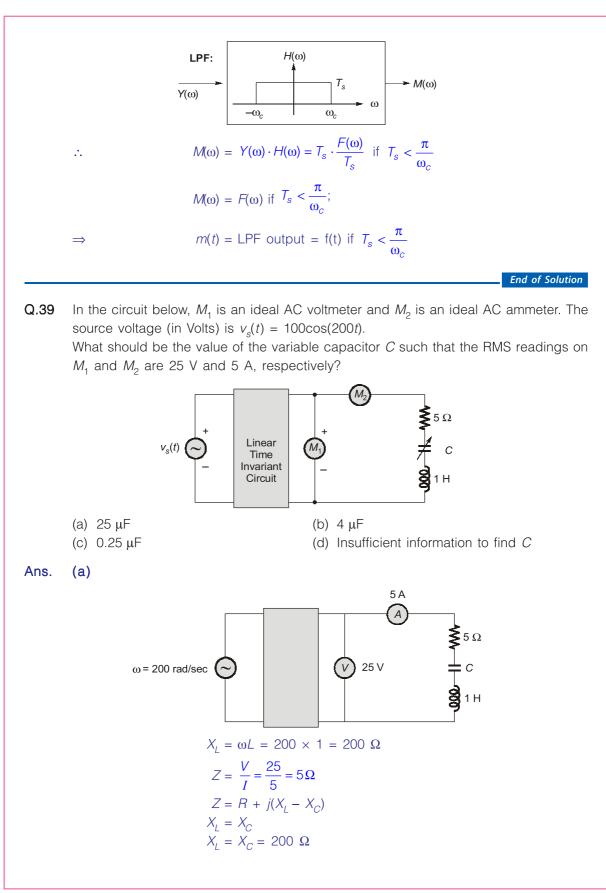




From the above table we get L = 2, I = 2 and R = 1End of Solution Q.38 Consider a continuous-time finite-energy signal f(t) whose Fourier transform vanishes outside the frequency interval  $[-\omega_c, \omega_c]$ , where  $\omega_c$  is in rad/sec. The signal f(t) is uniformly sampled to obtain y(t) = f(t) p(t). Here  $p(t) = \sum_{n=1}^{\infty} \delta(t - \tau - nT_s),$ with  $\delta(t)$  being the Dirac impulse,  $T_s > 0$ , and  $\tau > 0$ . The sampled signal y(t) is passed through an ideal lowpass filter  $h(t) = \omega_c T_s \frac{\sin(\omega_c t)}{\pi \omega_c t}$  with cutoff frequency  $\omega_c$  and passband gain  $T_{\rm s}$ The output of the filter is given by \_ (b)  $f(t - \tau)$  if  $T_s < \pi/\omega_c$ (d)  $T_s f(t)$  if  $T_s < 2\pi/\omega_c$ (a) f(t) if  $T_s < \pi/\omega_c$ (c)  $f(t - \tau)$  if  $T_s < 2\pi/\omega_c$ Ans. (a)  $p(t) = \sum_{s=1}^{\infty} \delta(t - \tau - nT_s)$ p(t)For p(t), FS-coefficient:  $C_n = \frac{1}{T_s} \int_{T_s} p(t) e^{-jn\omega_s t} dt$ =  $\frac{1}{T_s} \int_{T_s} \delta(t-\tau) e^{-jn\omega_s t} dt$   $\cdots$   $-\tau_s + \tau$  0  $=\frac{e^{-j\Omega\omega_{s}\tau}}{T_{c}}$  $P(\omega) = 2\pi \sum_{n=-\infty}^{\infty} C_n \delta(\omega - n\omega_s) = 2\pi \sum_{n=-\infty}^{\infty} \frac{e^{-jn\omega_s \tau}}{T_s} \delta(\omega - n\omega_s)$  $\Rightarrow$  $= \omega_s \sum e^{-jn\omega_s \tau} \delta(\omega - n\omega_s)$  $y(t) = f(t) \cdot p(t)$ Now,  $Y(\omega) = \frac{1}{2\pi} [F(\omega) * P(\omega)] = \frac{1}{2\pi} \left| F(\omega) * \omega_{s} \sum_{\rho = -\infty}^{\infty} e^{-j\rho\omega_{s}\tau} \delta(\omega - \rho\omega_{s}) \right|$  $\Rightarrow$  $=\frac{1}{T}\sum_{k=1}^{\infty}e^{-jn\omega_{s}\tau}F(\omega-n\omega_{s})$  $= \left[ \dots + \frac{e^{-j\omega_{s}\tau}}{T_{s}} F(\omega + \omega_{s}) + \frac{F(\omega)}{T_{s}} + \frac{e^{-j\omega_{s}\tau}}{T_{s}} F(\omega - \omega_{s}) + \dots \right]$ 









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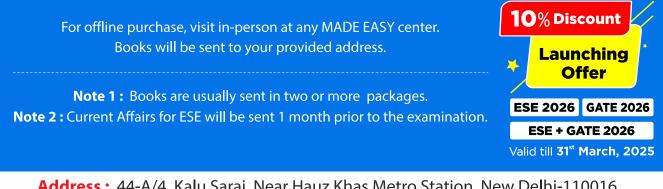
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$$\frac{1}{\omega C} = 200$$
$$\frac{1}{200 \times C} = 200$$
$$C = \frac{1}{200 \times 200} = 25 \mu F$$

End of Solution

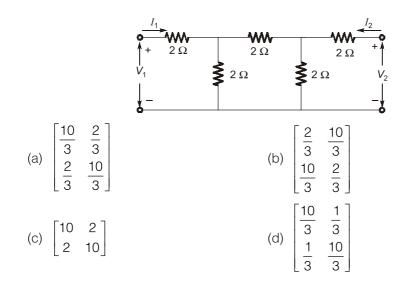
**Q.40** The *Z*-parameter matrix of a two port network relates the port voltages and port currents as follows:

$$\begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = Z \begin{bmatrix} I_1 \\ I_2 \end{bmatrix}$$

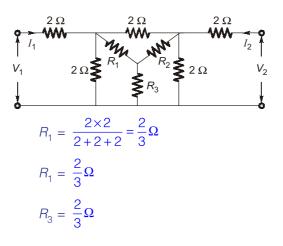
The Z-parameter matrix (with each entry in Ohms) of the network shown below is

E 202

Electronics Engineering

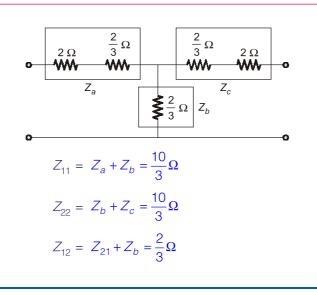


Ans. (a)









End of Solution

**Q.41** A source transmits symbol *S* that takes values uniformly at random from the set  $\{-2, 0, 2\}$ . The receiver obtains Y = S + N, where *N* is a zero-mean Gaussian random variable independent of *S*. The receiver uses the maximum likelihood decoder to estimate the transmitted symbol *S*.

Suppose the probability of symbol estimation error  $P_e$  is expressed as follows:

$$P_e = \alpha P(N > 1),$$

where P(N > 1) denotes the probability that N exceeds1. What is the value of  $\alpha$ ?

(a) 
$$\frac{1}{3}$$
 (b) 1  
(c)  $\frac{2}{3}$  (d)  $\frac{4}{3}$ 

#### Ans.

(d)

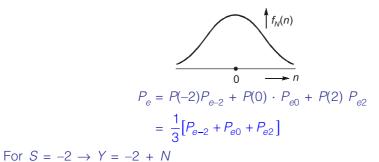
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S takes the values uniformly at random from the set {-2, 0, 2}

$$P(-2) = P(0) = P(2) = \frac{1}{3}$$

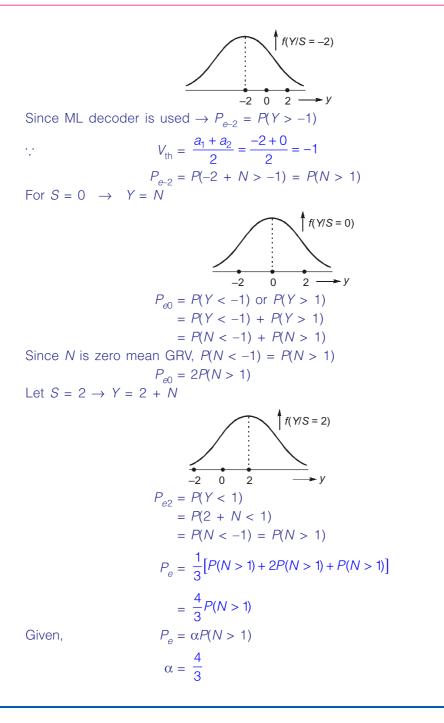
$$Y = S + N$$

Given N is zero mean Gaussian R.V









End of Solution

Q.42 Consider a real-valued random process

$$f(t) = \sum_{n=1}^{N} a_n p(t - nT),$$

where T > 0 and N is a positive integer. Here, p(t) = 1 for  $t \in [0, 0.5T]$  and 0 otherwise. The coefficients  $a_n$  are pairwise independent, zero-mean unit-variance random variables. Read the following statements about the random process and choose the correct option.

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	Total	150	150	

	Test No.	Activate Date	Total Marks	Total Questions	Total Time
	1	11 <sup>th</sup> Feb 2025	150 Marks	150 Qs	2 Hours
	2	14 <sup>th</sup> Feb 2025	150 Marks	150 Qs	2 Hours
	3	18 <sup>th</sup> Feb 2025	150 Marks	150 Qs	2 Hours
Test	4	21 <sup>st</sup> Feb 2025	150 Marks	150 Qs	2 Hours
Series	5	25 <sup>th</sup> Feb 2025	150 Marks	150 Qs	2 Hours
Schedule	6	28 <sup>th</sup> Feb 2025	150 Marks	150 Qs	2 Hours
	7	4 <sup>th</sup> Mar 2025	150 Marks	150 Qs	2 Hours
	8	7 <sup>th</sup> Mar 2025	150 Marks	150 Qs	2 Hours
	9	11 <sup>th</sup> Mar 2025	150 Marks	150 Qs	2 Hours
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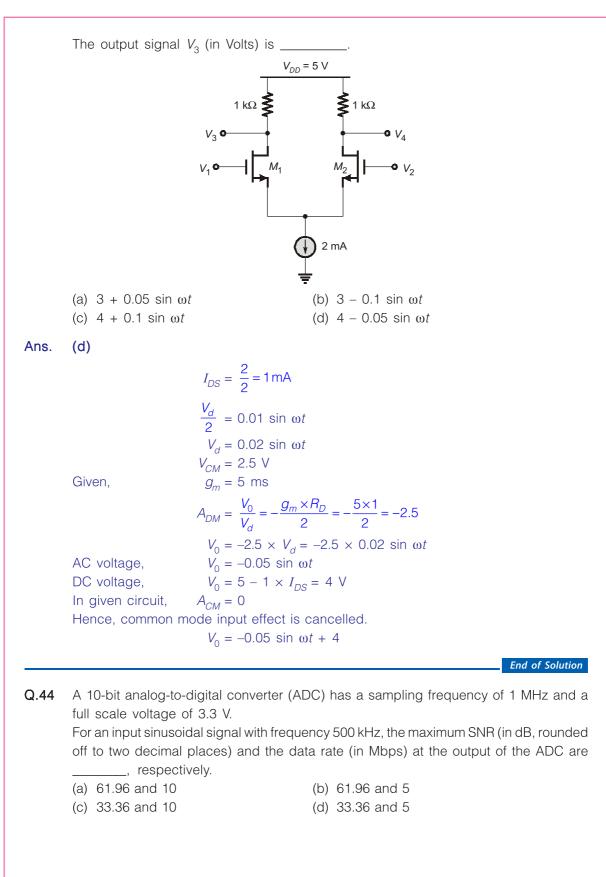




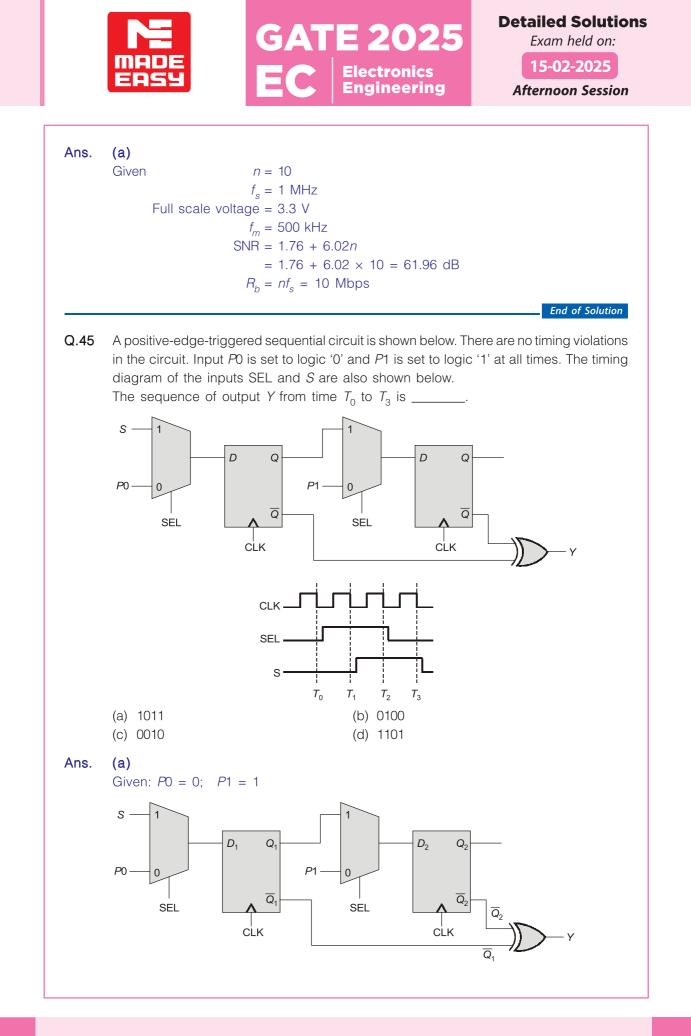
(i) The mean of the process f(t) is independent of time t. (ii) The autocorrelation function  $E[f(t) f(t + \tau)]$  is independent of time t for all  $\tau$ . (Here,  $E[\cdot]$  is the expectation operation.) (a) (i) is TRUE and (ii) is FALSE (b) Both (i) and (ii) are TRUE (c) Both (i) and (ii) are FALSE (d) (i) is FALSE and (ii) is TRUE Ans. (a) Given Random process  $f(t) = \sum_{n=1}^{N} a_n p(t - nT)$  $1 \xrightarrow{\dagger} p(t)$  $\{a_n\}$  specifies set of independent random variables each having zero mean and unit variance. Mean:  $E[f(t)] = \sum_{n=1}^{N} E[a_n]p(t - nT)$ Given that  $E[a_n] = 0$ E[f(t)] = 0E[f(t)] is independent of t. ACF:  $ACF[f(t)] = \sum_{n=1}^{n} ACF[a_n] \cdot ACF[p(t - nT)]$ Let p(t - nT) = g(t) where g(t) is a deterministic signal.  $\mathsf{ACF}[p(t-nT)] = \mathsf{ACF}[g(t)] = \int_{-\infty}^{\infty} g(t) \cdot g(t+\tau) dt$ From observation ACF[q(t)] is function of t. i.e., ACF[f(t)] is function of t. End of Solution The identical MOSFETs  $M_1$  and  $M_2$  in the circuit given below are ideal and biased in Q.43 the saturation region.  $M_1$  and  $M_2$  have a transconductance  $g_m$  of 5 mS. The input signals (in Volts) are:  $V_1 = 2.5 + 0.01 \sin \omega t$  $V_2 = 2.5 - 0.01 \sin \omega t$ 

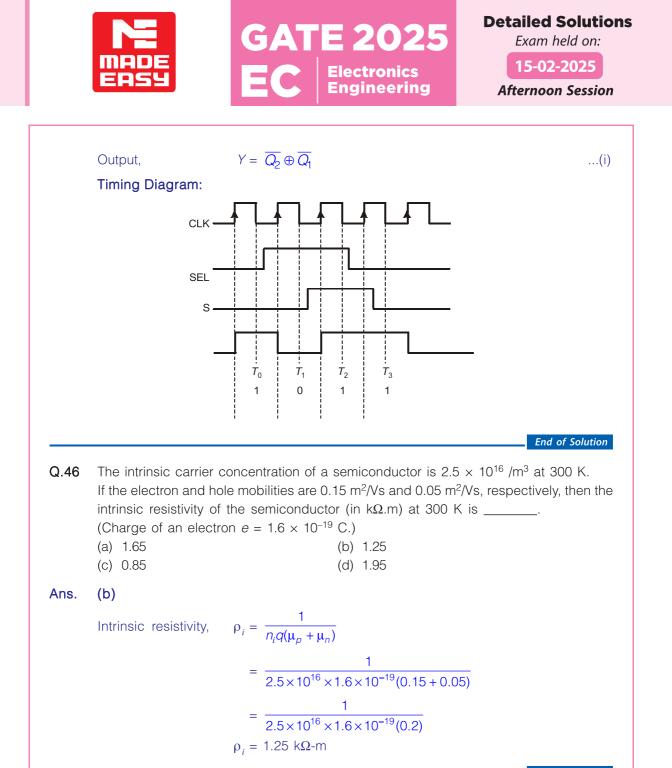






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End of Solution

**Q.47** In the circuit shown, the identical transistors Q1 and Q2 are biased in the active region with  $\beta$  =120. The Zener diode is in the breakdown region with  $V_Z$  = 5 V and  $I_Z$  = 25 mA. If  $I_L$  = 12 mA and  $V_{EB1} = V_{EB2}$  = 0.7 V, then the values of  $R_1$  and  $R_2$  (in k $\Omega$ , rounded off to one decimal place) are \_\_\_\_\_, respectively.

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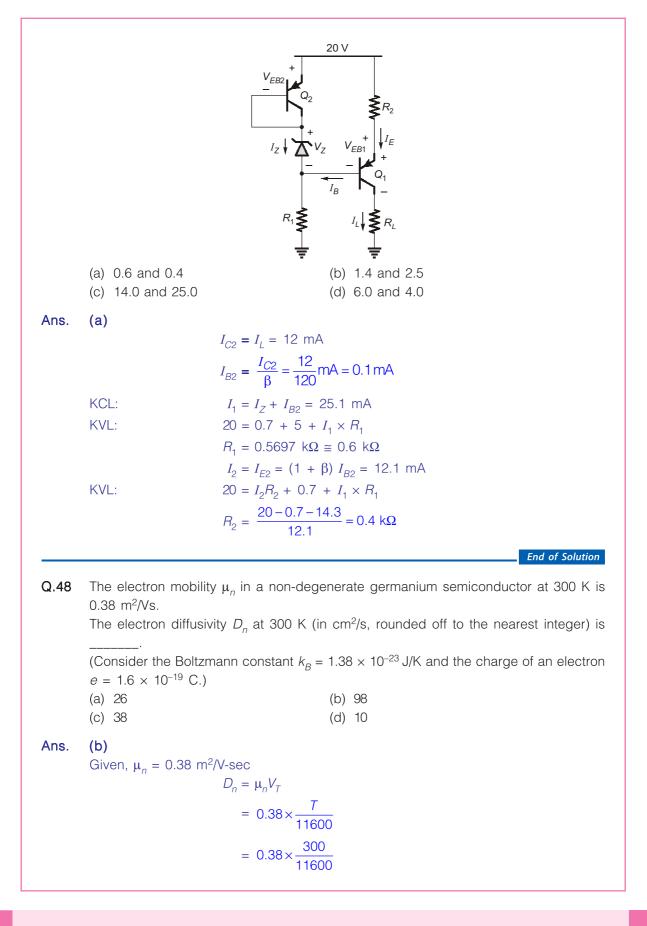
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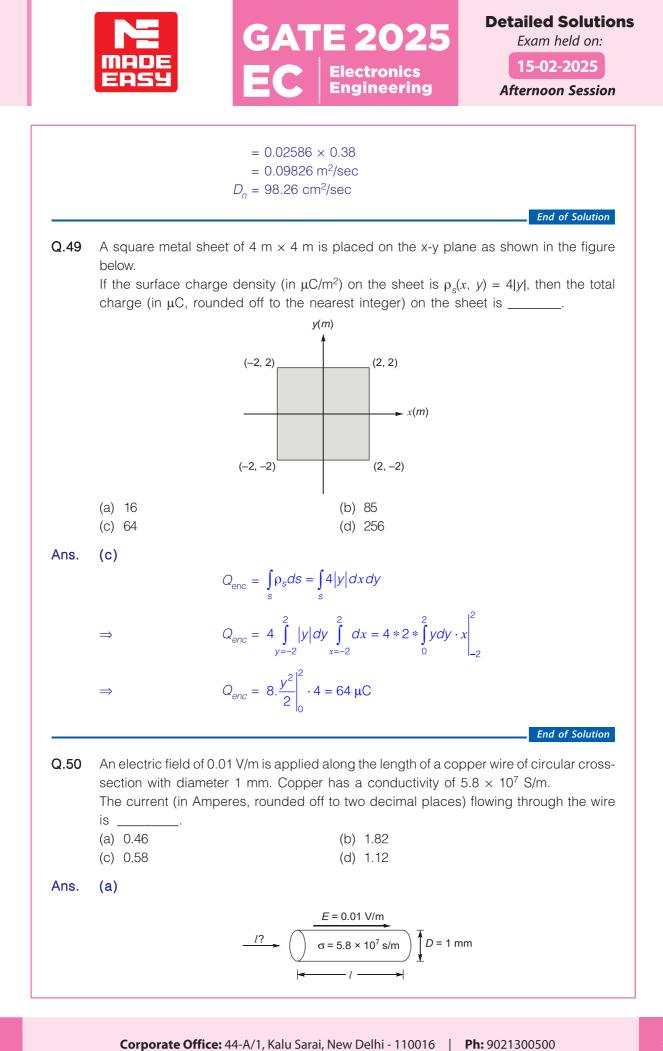
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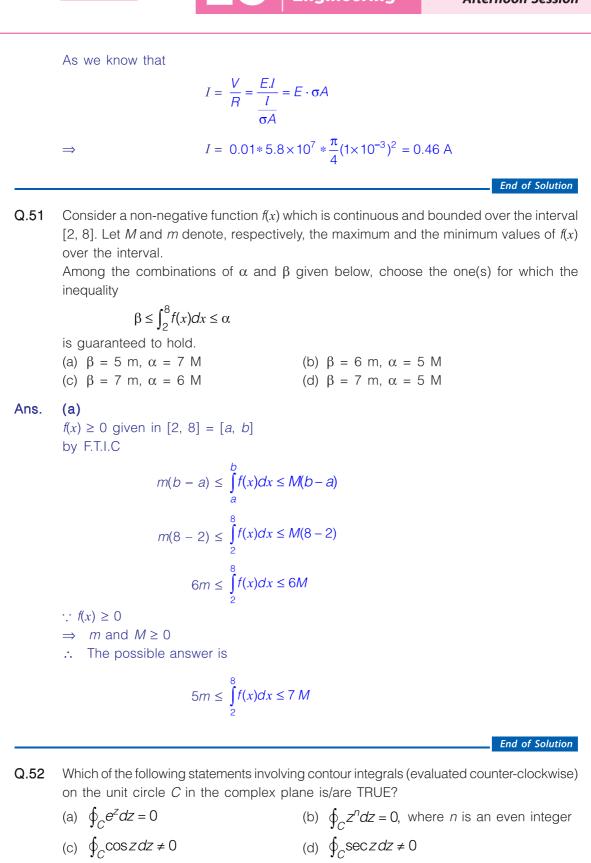
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Ans. (a, b) Given C: |z| = 1C: |z| = 10 (a)  $f = e^z$  is always analytic ∴ by C.I.T  $\oint_C f(z) dz = 0$  (a) is true (b)  $f(z) = z^n$  is always analytic : by C.I.T  $\oint_C z^n dz = 0$  (b) is true (c)  $f(z) = \cos z$  is also always analytic ∴ by C.I.T  $\oint_C \cos z \, dz = 0$ : (c) is false. (d)  $f(z) = \sec z$ , Singularities are given by  $\cos z = 0 \implies z = (2n+1)\frac{\pi}{2}$  $Z = \pm \frac{\pi}{2}, \pm \frac{3\pi}{2}, \dots$ i.e., : all nodes lies out 'C' :. by C.I.T  $\oint_C \sec z \, dz = 0$  (:. (d) is false) End of Solution

**Q.53** Consider a system where  $x_1(t)$ ,  $x_2(t)$ , and  $x_3(t)$  are three internal state signals and u(t) is the input signal. The differential equations governing the system are given by

$$\frac{d}{dt} \begin{bmatrix} x_1(t) \\ x_2(t) \\ x_3(t) \end{bmatrix} = \begin{bmatrix} 2 & 0 & 0 \\ 0 & -2 & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \\ x_3(t) \end{bmatrix} + \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix} u(t).$$

Which of the following statements is/are TRUE?

- (a) The signals  $x_1(t)$ ,  $x_2(t)$ , and  $x_3(t)$  are bounded for all bounded inputs.
- (b) There exists a bounded input such that at least one of the signals  $x_1(t)$ ,  $x_2(t)$ , and  $x_3(t)$  is unbounded.
- (c) There exists a bounded input such that the signals  $x_1(t)$ ,  $x_2(t)$  and  $x_3(t)$  are unbounded.
- (d) The signals  $x_1(t)$ ,  $x_2(t)$  and  $x_3(t)$  are unbounded for all bounded inputs.



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Ans. (b) From the given matrix, the eigen values are 2, -2, 0Hence system  $x_1(t)$  whose eigen value is positive, it is unbounded for bounded input. Similarly system  $x_3(t)$  is also unbounded for bounded input. Only system  $x_2(t)$  is bounded for bounded input (since its eigen value is negative). End of Solution Q.54 The random variable X takes values in  $\{-1, 0, 1\}$  with probabilities P(X = -1) = P(X = 1) and  $\alpha$  and  $P(X = 0) = 1 - 2\alpha$ , where  $0 < \alpha < \frac{1}{2}$ . Let  $g(\alpha)$  denote the entropy of X (in bits), parameterized by  $\alpha$ . Which of the following statements is/are TRUE? (a) g(0.4) > g(0.3)(b) g(0.3) > g(0.4)(c) g(0.3) > g(0.25)(d) g(0.25) > g(0.3)Ans. (b, c)  $X \in \{-1, 0, 1\}$  $P(X = -1) = P(X = 1) = \alpha$  and  $P(X = 0) = 1 - 2\alpha$  where  $0 < \alpha < \frac{1}{2}$  $q(\alpha) =$  Entropy of X  $= -\sum_{i} P(x_i) \log_2 P(x_i)$  $= -\{\alpha \log_2 \alpha + \alpha \log_2 \alpha + (1 - 2\alpha) \log_2 (1 - 2\alpha)\}$  $g(\alpha) = -\{2\alpha \log_2 \alpha + (1 - 2\alpha)\log_2(1 - 2\alpha)\}$  $g(0.25) = -\{0.5 \log_2 0.25 + 0.5 \log_2 0.5\} = 1.5$  $g(0.3) = -\{0.6 \log_2 0.3 + 0.4 \log_2 0.4\} = 1.57$  $g(0.4) = -\{0.8 \log_2 0.4 + 0.2 \log_2 0.2\} = 1.52$ g(0.3) > g(0.4) > g(0.25)End of Solution Q.55 Let f(t) be a periodic signal with fundamental period  $T_0 > 0$ . Consider the signal  $y(t) = f(\alpha t)$ , where  $\alpha > 1$ . The Fourier series expansions of f(t) and y(t) are given by  $f(t) = \sum_{k=1}^{\infty} c_k e^{j\frac{2\pi}{T_0}kT} \text{ and } y(t) = \sum_{k=1}^{\infty} d_k e^{j\frac{2\pi}{T_0}\alpha kT}.$ Which of the following statements is/are TRUE? (a)  $c_k = d_k$  for all k(b) y(t) is periodic with a fundamental period  $\alpha T_0$ (c)  $c_k = d_k / \alpha$  for all k (d) y(t) is periodic with a fundamental period  $T_0/\alpha$ Ans. (a, d)  $f(t) = c_k \qquad y(t) = d_k \qquad FTP: T_0$ Given,

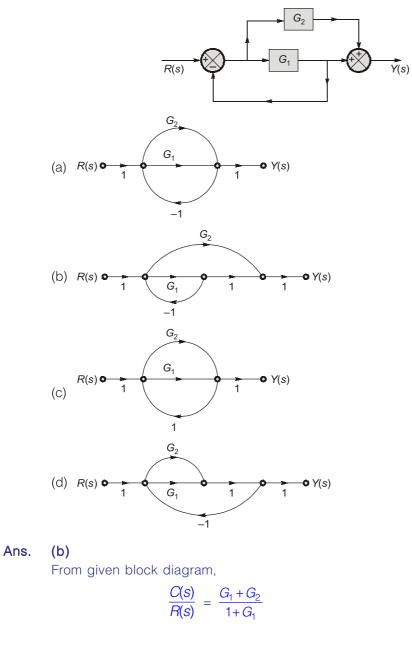




and By time-scaling	y(t) = f(at) property of FS,	
	$d_k = c_k$ if a is positive	
Given that,	a > 1	
<i>.</i>	$d_k = c_k$ for all 'k'	

End of Solution

**Q.56** Consider a system represented by the block diagram shown below. Which of the following signal flow graphs represent(s) this system? Choose the correct option(s).







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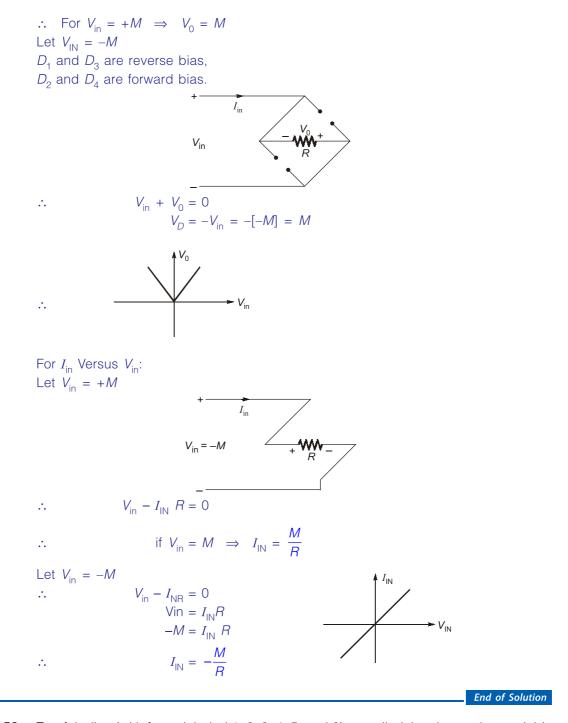




and in options, from option (a),  $\frac{Y(s)}{R(s)} = \frac{G_1 + G_2}{1 + G_1}$ : Option (b) only satisfies. End of Solution Q.57 All the diodes in the circuit given below are ideal. Which of the following plots is/are correct when  $V_1$  (in Volts) is swept from -M to M?  $I_I$ М (a) (C) (d) Ans. (a, d)  $I_{\rm in}$  $V_{in}$ Given,  $V_{\rm in}$  is steep of -M to +MFor  $V_0$  versus  $V_{\rm IN}$ : Let  $V_{\rm IN} = +M$ In the bridge rectifier,  $D_1$  and  $D_3$  are forward bias, and  $D_2$  and  $D_4$  are reverse bias, I.,  $V_{\rm in}$  $V_{\rm in} - V_0 = 0 \implies V_{\rm in} = V_0$ ...







Q.58 Two fair dice (with faces labeled 1, 2, 3, 4, 5, and 6) are rolled. Let the random variable X denote the sum of the outcomes obtained.The expectation of X is \_\_\_\_\_ (rounded off to two decimal places).





Detailed Solutions Exam held on: 15-02-2025

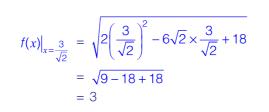
Afternoon Session

(7.0) (7.0 to 7.0) Ans. Sum of two dice = {2, 3, 4, 5, ...., 12} 5 6 7 8 9 10 11 | 12 2 3 4 2 36 1 P(x)36  $E(X) = \sum x \cdot P(x)$  $= 2 \times \frac{1}{36} + 3 \times \frac{2}{36} + 4 \times \frac{3}{36} + 5 \times \frac{4}{36} + 6 \times \frac{5}{36} + 7 \times \frac{6}{36}$  $+8\times\frac{5}{36}+9\times\frac{4}{36}+10\times\frac{3}{36}+11\times\frac{2}{36}+12\times\frac{1}{36}$  $=\frac{252}{36}=7$ End of Solution Q.59 Consider the vectors  $a = \begin{bmatrix} 1 \\ 1 \end{bmatrix}, b = \begin{bmatrix} 0 \\ 3\sqrt{2} \end{bmatrix}.$ For real-valued scalar variable x, the value of  $\min || ax - b ||_2$ is \_\_\_\_\_ (rounded off to two decimal places).  $\|\cdot\|_2$  denotes the Euclidean norm, i.e., for  $y = \begin{bmatrix} y_1 \\ y_2 \end{bmatrix}$ ,  $\|y\|_2 = \sqrt{y_1^2 + y_2^2}$ . Ans. (3.0) (3.0 to 3.0)  $ax - b = \begin{vmatrix} x \\ x - 3\sqrt{2} \end{vmatrix}$  $\begin{vmatrix} x \\ x - 3\sqrt{2} \end{vmatrix} = f(x) = \text{norm}$  $f(x) = \sqrt{x^2 + (x - 3\sqrt{2})^2}$  $= \sqrt{x^2 + x^2 + 18 - 6\sqrt{2}x}$  $f(x) = \sqrt{2x^2 - 6\sqrt{2}x + 18}$  $q(x) = 2x^2 - 6\sqrt{2}x + 18$ Let,  $q'(x) = 4x - 6\sqrt{2} = 0$  $x = \frac{6\sqrt{2}}{4} = \frac{3\sqrt{2}}{2} = \frac{3}{\sqrt{2}}$ Point of minima =  $\frac{3}{\sqrt{2}}$ 









End of Solution

**Q.60** *X* and *Y* are Bernoulli random variables taking values in {0, 1}. The joint probability mass function of the random variables is given by:

P(X = 0, Y = 0) = 0.06 P(X = 0, Y = 1) = 0.14 P(X = 1, Y = 0) = 0.24 P(X = 1, Y = 1) = 0.56The mutual information I(X; Y) is \_\_\_\_\_ (rounded off to two decimal places).

Ans. (0) (0.0 to 0.0)

Given, P(X = 0, Y = 0) = 0.06 P(X = 0, Y = 1) = 0.14 P(X = 1, Y = 0) = 0.24P(X = 1, Y = 1) = 0.56

$$[P(X, Y)] = \begin{array}{c} X = 0 \\ X = 1 \begin{bmatrix} 0.06 & 0.14 \\ 0.24 & 0.56 \end{bmatrix} \rightarrow P(X = 0) = 0.2 \\ \downarrow \\ P(Y = 0) = 0.3 \end{array}$$

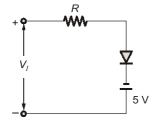
From above

 $P(X = 0, Y = 0) = P(X = 0) \cdot P(Y = 0)$   $P(X = 0, Y = 1) = P(X = 0) \cdot P(Y = 1)$   $P(X = 1, Y = 0) = P(X = 1) \cdot P(Y = 0)$   $P(X = 1, Y = 1) = P(X = 1) \cdot P(Y = 1)$ We conclude that X and Y are independent. So that I(X; Y) = 0

End of Solution

**Q.61** The diode in the circuit shown below is ideal. The input voltage (in Volts) is given by  $V_1 = 10 \sin 100 \pi t$ , where time *t* is in seconds. The time duration (in ms, rounded off to two decimal places) for which the diode is forward

biased during one period of the input is \_\_\_\_\_.







Ans. (13.33) (13.32 to 13.34)  $2\pi ft = 100\pi t$ Given,  $f = 50 \, \text{Hz}$  $T_0 = \frac{1}{f} = 20$  msec  $V_i = 10 \sin(100\pi t) = 10 \sin\alpha$ ,  $\alpha = 100\pi t$ Diode conducts if  $V_i > -5$  V Diode is OFF if  $V_i < -5$  V  $V_i$ 10 V –5 V –10 V 10 sin  $\alpha = -5$  $\alpha = \sin^{-1}\left(\frac{-5}{10}\right)$  $\alpha = \frac{7\pi}{6}, \frac{11\pi}{6}$  $100\pi t_1 = \frac{7\pi}{6}$  $t_1 = 11.66$  msec  $\alpha_2 = 100\pi t_2 = \frac{11\pi}{6}$  $t_2 = 18.33$  msec Diode is OFF from  $t_1$  to  $t_2$  $t_{\text{OFF}} = t_2 - t_1 = 6.66 \text{ msec}$  $t_{\rm ON} = T_0 - t_{\rm OFF} = 13.33$  msec End of Solution

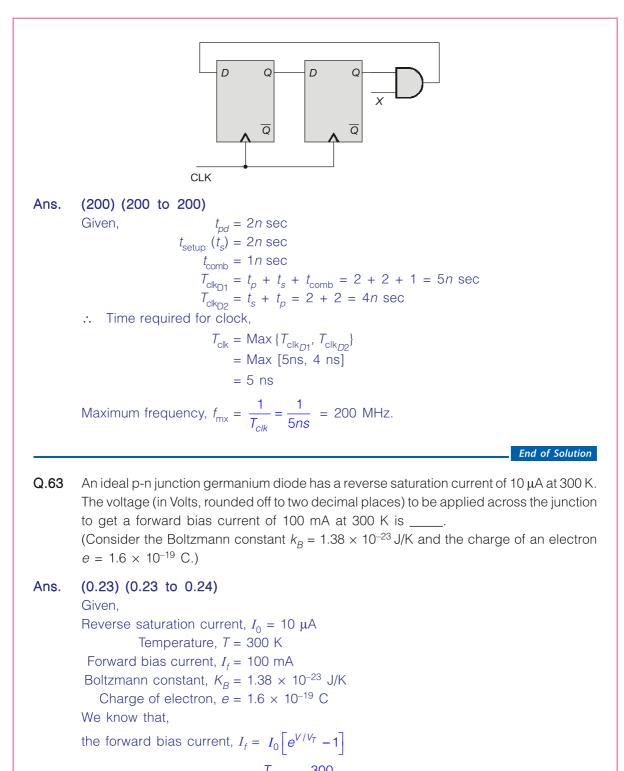
Q.62 In the circuit shown below, the AND gate has a propagation delay of 1 ns. The edgetriggered flip-flops have a set-up time of 2 ns, a hold-time of 0 ns, and a clock-to-Q delay of 2 ns.

The maximum clock frequency (in MHz, rounded off to the nearest integer) such that there are no setup violations is \_\_\_\_\_.

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where,

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 $V_{T} = \frac{T}{11600} = \frac{300}{11600} = 25.86 \text{ mV}$  $100 \times 10^{-3} = 10 \times 10^{-6} \left[ e^{\frac{V}{25.86 \times 10^{-3}}} - 1 \right]$  $10^{4} = e^{\frac{V}{25.86 \times 10^{-3}}}$ 

