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Important Questions  
for **GATE 2022**

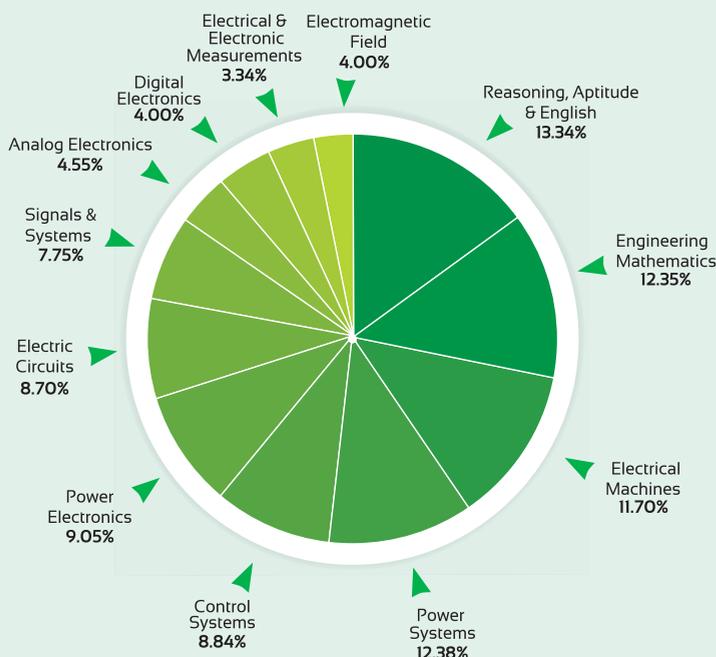
**ELECTRICAL  
ENGINEERING**

**Day 8 of 8**

**Q.176 - Q.200 (Out of 200 Questions)**

**Power Electronics &  
Electromagnetic Fields**

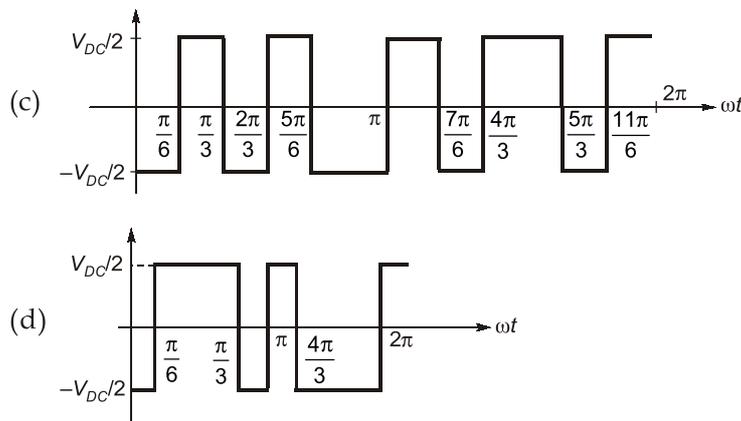
**SUBJECT-WISE WEIGHTAGE ANALYSIS OF GATE SYLLABUS**



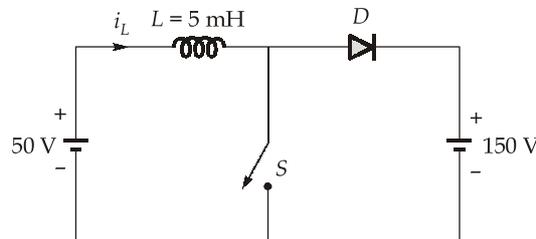
Subject	Average % (last 5 yrs)
Reasoning, Aptitude & English	13.34%
Engineering Mathematics	12.35%
Electrical Machines	11.70%
Power Systems	12.38%
Control Systems	8.84%
Power Electronics	9.05%
Electric Circuits	8.70%
Signals & Systems	7.75%
Analog Electronics	4.55%
Digital Electronics	4.00%
Electrical & Electronic Measurements	3.34%
Electromagnetic Fields	4.00%
<b>Total</b>	<b>100%</b>







**Q.181** In the circuit shown below :



If the switching frequency is 5 kHz, duty ratio is 0.4 and circuit reached to steady state then the average value of inductor current is \_\_\_\_\_ A.  
(Assuming the switch and diode to be ideal and discontinuous mode of current conduction)

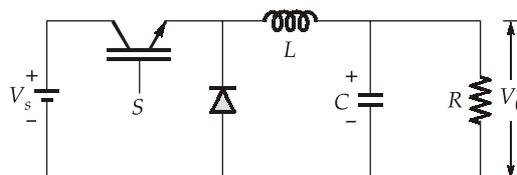
**Q.182** A full bridge inverter has bi-directional switches employed in a manner that their switching sequence produces a square wave voltage across a series  $R$ - $L$  load. If the switching frequency is 50 Hz, dc supply voltage,  $V_{dc} = 100 \text{ V}$ ,  $R = 10 \Omega$  and  $L = 25 \text{ mH}$ , then power absorbed by the load and rms value of current will be respectively

- (a) 543.62 W, 5.41 A
- (b) 337.97 W, 4.23 A
- (c) 516.96 W, 7.19 A
- (d) 243.62 W, 3.43 A

**Q.183** For a power diode, the reverse recovery time is  $4.2 \mu\text{s}$  and the rate of diode current decay is  $40 \text{ A}/\mu\text{s}$ . For softness factor 0.25, the storage charge is

- (a)  $350 \mu\text{C}$
- (b)  $282.24 \mu\text{C}$
- (c)  $200 \mu\text{C}$
- (d)  $210.6 \mu\text{C}$

**Q.184** In the chopper circuit shown in figure, the input voltage has a constant value  $V_s$ . The output voltage  $V_0$  is assumed ripple free. The switch  $S$  is operated with a switching time period  $T$  and a duty ratio  $\alpha$ . At the boundary condition of continuous and discontinuous condition of the inductor current  $i_L$ , the value of critical inductance  $L_C$  \_\_\_\_\_





**Q.190** A Buck-boost converter has the following parameters:

$$\begin{array}{lll} V_S = 24 \text{ V} & D = 0.4 & R = 5 \Omega \\ L = 20 \mu\text{H} & C = 80 \mu\text{F} & f = 100 \text{ kHz} \end{array}$$

The ratio of maximum inductor current,  $I_{L, \max}$  to minimum inductor current  $I_{L, \min}$  for above converter considering continuous inductor current will be

- (a) 1.50 (b) 2.63  
(c) 1.24 (d) 3.50

**Q.191** A single phase semi converter, connected to 230 V, 50 Hz, source is feeding a load  $R = 40 \Omega$  in series with large inductor that makes the load current ripple free. If the value of firing angle is  $45^\circ$ , then the current distortion factor is

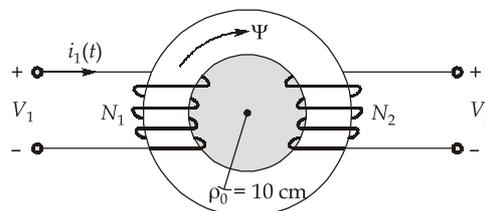
- (a) 0.850 (b) 0.700  
(c) 0.550 (d) 0.960

**Q.192** Surface  $x = 0$  separates two perfect dielectric with relative permittivities  $\epsilon_{r1} = 2$  and  $\epsilon_{r2} = 8$ .

If  $\vec{E}_1 = (100\hat{a}_x + 200\hat{a}_y - 50\hat{a}_z)$  V/m for  $x > 0$ . Then  $\vec{E}_2$  is

- (a)  $100\hat{a}_x + 50\hat{a}_y - 1.25\hat{a}_z$  (b)  $400\hat{a}_x + 200\hat{a}_y - 50\hat{a}_z$   
(c)  $25\hat{a}_x + 200\hat{a}_y - 50\hat{a}_z$  (d)  $100\hat{a}_x + 800\hat{a}_y - 200\hat{a}_z$

**Q.193** The magnetic circuit shown below has a uniform cross-section of  $10^{-3} \text{ m}^2$ . If the circuit is energized by a current  $i_1(t) = 3 \sin 100\pi t$  A in the coil of  $N_1 = 200$  turns, then the emf induced in the coil of  $N_2 = 100$  turns is  
(Assume  $\mu = 500 \mu_0$ )



- (a)  $-6\pi \sin 100\pi t$  V (b)  $-6\pi \sin 200\pi t$  V  
(c)  $-6\pi \cos 100\pi t$  V (d)  $8\pi \cos 100\pi t$  V

**Q.194** Two uniformly distributed line charges of  $\lambda = 5 \text{ nC/m}$  each are parallel to the X-axis, one at  $z = 0$  and  $y = -2 \text{ m}$  and the other  $z = 0$  and  $y = 4 \text{ m}$ . The field  $E$  at  $(4, 1, 3) \text{ m}$  is \_\_\_\_\_

- (a)  $30 a_x$  (b)  $30 a_y$   
(c)  $6 a_z$  (d)  $30 a_z$

**Q.195** The electric flux density vector of a certain charge distribution is given by

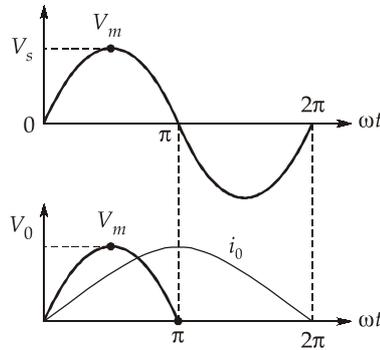
$$\vec{D} = 12xyz\hat{a}_x + (6x^2z + 6yz)\hat{a}_y + (6x^2y + 3y^2)\hat{a}_z \text{ C/m}^2.$$

The amount of charge enclosed in cube having a volume of  $10^{-5} \text{ m}^3$  located at  $(x = 1 \text{ m}, y = 1 \text{ m}, z = 1 \text{ m})$  is \_\_\_\_\_  $\mu\text{C}$ .



### Detailed Explanations

176. (d)  
In figure (a),



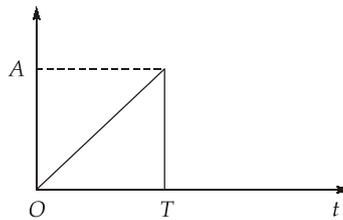
Inductor stores energy upto  $\pi$  radian and dissipates energy upto  $2\pi$  radians.

So, diode conducts for  $360^\circ$  in figure (a).

In figure (b), the source is a 'cosine' function, so capacitor charges to its maximum value instantaneously as switch is closed at  $t = 0$ .

So, diode conducts for  $0^\circ$  in figure (b).

177. (a)

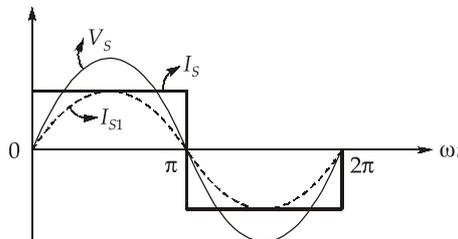


The rms value of the waveform is  $\frac{A}{\sqrt{3}}$

$$\begin{aligned} \text{Power loss, } P &= V_{\text{rms}} I_{\text{rms}} f \times t \\ &= \frac{300}{\sqrt{3}} \times \frac{100}{\sqrt{3}} \times 50 \times 2 \times 10^{-6} = 1 \text{ W} \end{aligned}$$

178. 0.90 (0.80 to 0.99)

The waveform of source current and fundamental component of source current ( $I_{s1}$ ) is



$$\text{Input p.f.} = \frac{I_{s1}}{I_s} \cos \alpha = \frac{4I_s}{\sqrt{2\pi} I_s} \cos 0^\circ$$

$$\text{Input p.f.} = 0.9 \text{ lagging}$$

179. (a)

The main thyristor is turned off when  $i_c$  is,

$$i_c = I_p \sin \omega t = I_0$$

$$\omega t = \sin^{-1}\left(\frac{I_0}{I_p}\right)$$

Peak value of current through capacitor,

$$I_p = V_s \sqrt{\frac{C}{L}} = 230 \times \sqrt{\frac{20 \times 10^{-6}}{5 \times 10^{-6}}} = 460 \text{ A}$$

$$\omega t = \sin^{-1}\left(\frac{300}{460}\right) = 40.70^\circ$$

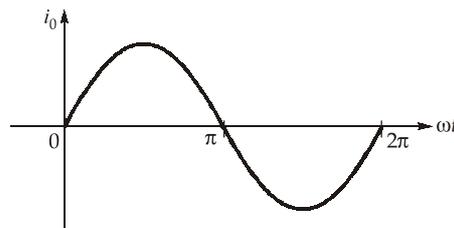
Voltage across main thyristor is

$$= V_s \cos \omega t$$

$$= 230 \times \cos(40.70^\circ) = 174.37 \text{ V}$$

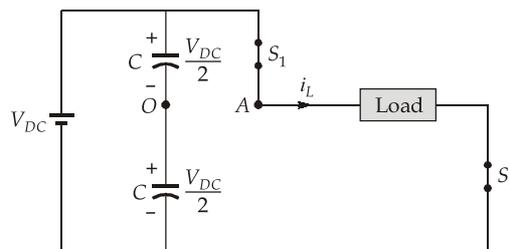
180. (c)

The load current waveform is



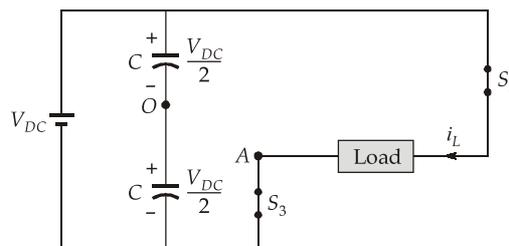
During the period switches are triggered the voltage across AO is

$S_1$  and  $S_4$  ON

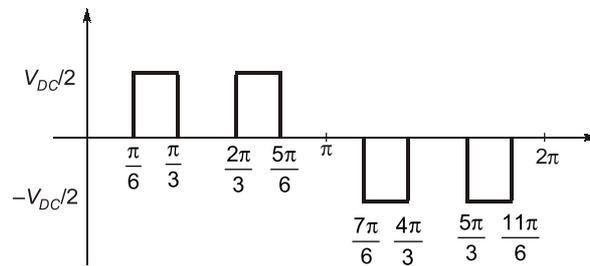


$$\text{Voltage } V_{AO} = \frac{V_{DC}}{2}$$

$S_2$  and  $S_3$  ON



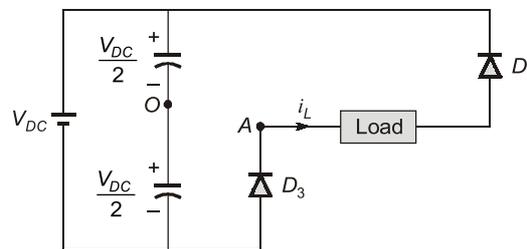
$$\text{Voltage } V_{AO} = -\frac{V_{DC}}{2}$$



When the switches are off.

The current direction from 0 to  $\pi$  is positive so it is in the direction given in the circuit.

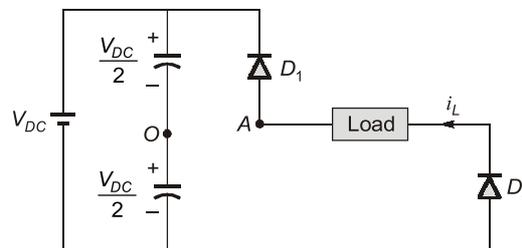
$$0 \leq \omega t \leq \pi$$



$$\text{Voltage } V_{AO} = -\frac{V_{DC}}{2}$$

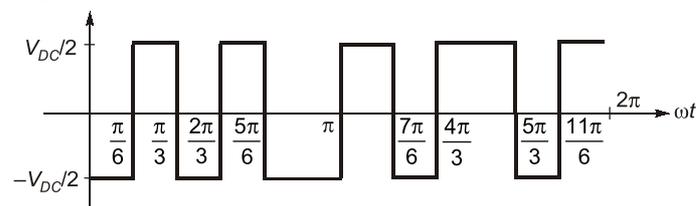
$$\pi \leq \omega t \leq 2\pi$$

The load current is negative so the current direction will be opposite to given direction.

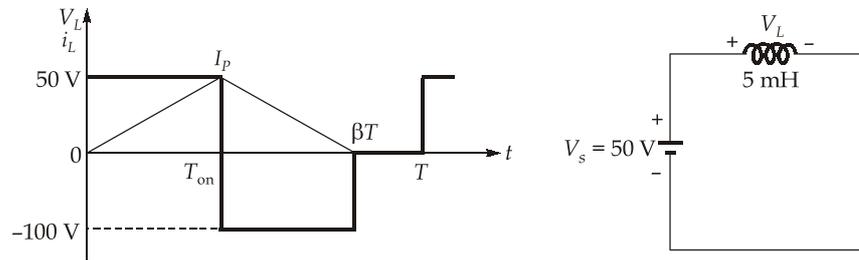


$$\text{Voltage } V_{AO} = \frac{V_{DC}}{2}$$

So, the complete voltage  $V_{AO}$  is



181. 0.24 (0.10 to 0.40)



During  $T_{on}$  the circuit behaves as,

$$V_s = L \frac{di}{dt}$$

$$di = \frac{V_s}{L} dt$$

Integrating on both sides, we get

$$I_p = \frac{V_s T_{on}}{L}$$

$$T_{on} = \alpha T = \frac{\alpha}{f} = \frac{0.4}{5 \times 10^3} = 80 \times 10^{-6} \text{ s}$$

$$I_p = \frac{50}{5 \times 10^{-3}} \times (80 \times 10^{-6}) = 0.8 \text{ A}$$

During  $T_{off}$  it is  $T_{on} \leq t \leq \beta T$

Apply KVL in the circuit,

$$V_L = V_s - V_0$$

$$(V_L)_{avg} = 0$$

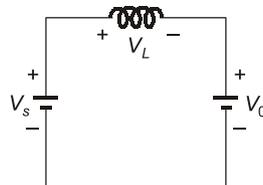
$$V_s T_{on} + (V_s - V_0) (\beta T - T_{on}) = 0$$

$$V_s \beta T = V_0 (\beta T - T_{on})$$

$$\frac{V_0}{V_s} = \frac{\beta}{\beta - \alpha}$$

$$\beta = 0.6$$

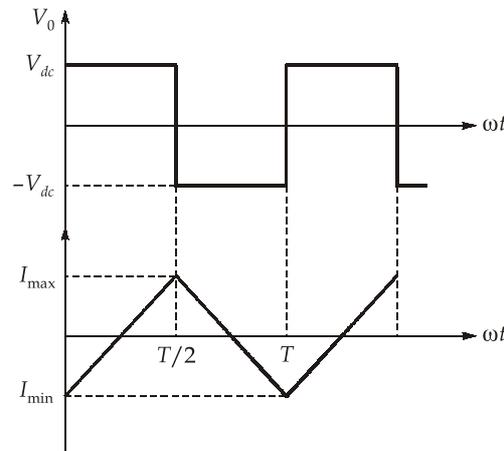
From the graph of  $I_L$ ,



$$I_{L(avg)} = \frac{1}{2} \times b \times h$$

$$= \frac{1}{2} \times \beta T \times I_p = \frac{1}{2} \times 0.6 \times 0.8 = 0.24 \text{ A}$$

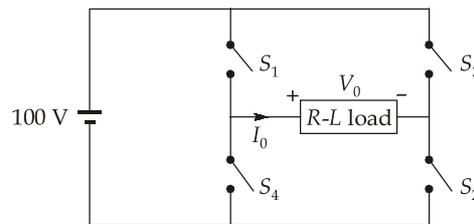
182. (c)



At  $t = 0$ ,

$S_1, S_2$  : switches are closed

$S_3, S_4$  : switches are open



Output power,

$$P = I_{or}^2 R$$

$$V_o = \sum_{n=1,3,5,\dots}^{\infty} \frac{4V_s}{n\pi} \sin n\omega t$$

$$V_{on-rms} = \frac{2\sqrt{2}}{n\pi} V_s$$

$$Z_n = \sqrt{R^2 + (n\omega L)^2}$$

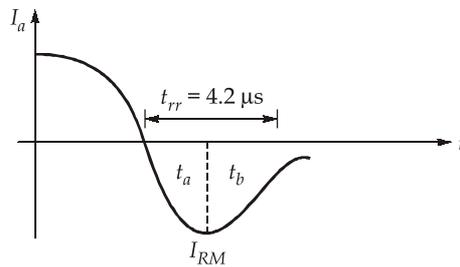
$$I_{on-rms} = \frac{V_{on}}{|Z_n|}$$

$n$	$V_{on-rms}$	$ Z_n $	$I_{on-rms}$
$n = 1$	$V_{o1} = 90 \text{ V}$	$ Z_1  = 12.71\Omega$	$I_{o1} = 7.08 \text{ A}$
$n = 3$	$V_{o3} = 30 \text{ V}$	$ Z_3  = 25.59\Omega$	$I_{o3} = 1.17 \text{ A}$
$n = 5$	$V_{o5} = 18 \text{ V}$	$ Z_5  = 40.5\Omega$	$I_{o5} = 0.44 \text{ A}$
$n = 7$	$V_{o7}$		$I_{o7} = 0.23 \text{ A}$

$$I_{or} = \sqrt{I_{o1}^2 + I_{o3}^2 + I_{o5}^2 + I_{o7}^2} = 7.19 \text{ A}$$

$$P = (7.19)^2 \times 10 = 516.96 \text{ W}$$

183. (b)



Given, softness factor =  $\frac{t_b}{t_a} = 0.25$

$$t_b = 0.25 t_a$$

$$t_{rr} = t_a + t_b = t_a + 0.25 t_a = 1.25 t_a$$

$$\therefore t_a = \frac{4.2}{1.25} = 3.36 \mu\text{s}$$

$$\text{Slope} = \frac{I_{RM}}{t_a} = 40 \text{ A}/\mu\text{s}$$

$$I_{RM} = 40 \times 3.36 = 134.4 \text{ A}$$

Total charge stored in the curve is,

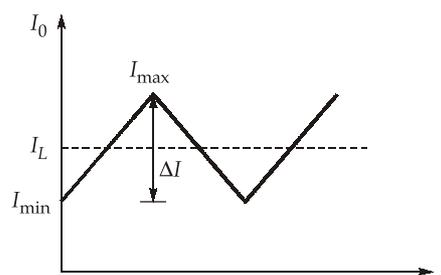
$$= \frac{1}{2} \times b \times h = \frac{1}{2} \times 4.2 \times 10^{-6} \times 134.4$$

$$= 282.24 \mu\text{C}$$

184. (a)

Given chopper is a Buck chopper so,

Output voltage,  $V_0 = \alpha V_s$



$$\Delta I = \frac{V_0 T_{off}}{L}$$

$$= 2I_L = 2I_0$$

This is applicable only at boundary of continuous and discontinuous condition.

$$L_C = \frac{\alpha V_s (1 - \alpha)}{2I_0 f}$$

185. (b)

Reference wave

$$\Rightarrow \begin{aligned} V_r &= 1 \text{ V;} \\ f &= 60 \text{ Hz (Inverter output frequency)} \end{aligned}$$

Carrier wave

$$\Rightarrow \begin{aligned} V_c &= 3 \text{ V;} \\ f_c &= 1.2 \text{ kHz} \end{aligned}$$

Number of pulses per half cycle,

$$N = \frac{f_c}{2f} = \frac{1.2 \times 10^3}{2 \times 60} = 10$$

$$\text{Modulation index} = \frac{V_r}{V_c} = \frac{1}{3}$$

$$\begin{aligned} \text{Total pulse width} &= \pi \left( 1 - \frac{V_r}{V_c} \right) \\ &= \pi \left( 1 - \frac{1}{3} \right) = \frac{2\pi}{3} = 120^\circ \end{aligned}$$

$$\text{Each pulse width} = \frac{2d}{N} = \frac{120^\circ}{10} = 12^\circ$$

186. (b)

Given,  $I_L = 500 \text{ mA} = 0.5 \text{ A}$   
 $t_d = 1.5 \mu\text{s}$   
 $\alpha = 30^\circ$ ,  
 $L = 10 \text{ mH}$   
 and  $R = 10 \Omega$

Instantaneous value of the input voltage  $V_s(t) = V_m \sin \omega t$

Where,  $V_m = \sqrt{2} \times 120 = 169.7 \text{ V}$   
 $V_1 = V_m \sin \omega t$   
 $= 169.7 \times \sin \frac{\pi}{6} = 84.85 \text{ V}$

The rate of rise of anode current  $\frac{di}{dt}$  at the instant of triggering is approximately

$$\frac{di}{dt} = \frac{V_1}{L} = \frac{84.85}{10 \times 10^{-3}} = 8485 \text{ A/s}$$

If  $\frac{di}{dt}$  is assumed constant for a short time after the gate triggering the time  $t_1$  required for the anode current to rise to the level of latching current is calculated from

$$t_1 \times \left(\frac{di}{dt}\right) = I_L$$

or  $t_1 \times 8485 = 0.5$   
 $t_1 = 58.93 \mu\text{s}$

Therefore, the minimum width of the gate pulse is

$$t_G = t_1 + t_d = 58.93 + 1.5$$

$$= 60.43 \mu\text{s}$$

**187. 18.18 (18.00 to 18.50)**

We know that,  $\gamma = \frac{\pi - 2d}{N + 1} + \frac{d}{N}$

$$= \frac{108}{3} + \frac{36}{2} = 54^\circ$$

$$V_0 = \sum_{n=1,3,5}^{\infty} \frac{8V_s}{n\pi} \sin n\gamma \sin \frac{nd}{2} \sin n\omega t$$

Peak value of 7<sup>th</sup> harmonics voltage

$$V_7 = \frac{8V_s}{n\pi} \sin n\gamma \sin \frac{nd}{2}$$

$$= \frac{8 \times 200}{7\pi} \times \sin(7 \times 54^\circ) \times \sin\left(\frac{7 \times 36^\circ}{2}\right)$$

$$= \frac{1600}{7\pi} \times 0.309 \times 0.809 = 18.188 \text{ V}$$

**188. 1.60 (1.40 to 1.80)**

$$V_{0 \text{ rms}} = \frac{V_{ml}}{2\sqrt{\pi}} \left[ \left( \frac{5\pi}{6} - \alpha \right) + \frac{1}{2} \sin \left( 2\alpha + \frac{\pi}{3} \right) \right]^{1/2}$$

$$V_{0 \text{ rms}} = \frac{220 \times \sqrt{2}}{2\sqrt{\pi}} \left[ (2.618 - 0.785) + \frac{1}{2} \sin(90^\circ + 60^\circ) \right]^{1/2}$$

$$= \frac{220 \times \sqrt{2}}{2\sqrt{\pi}} [1.833 + 0.25]^{1/2} = 126.67 \text{ V}$$

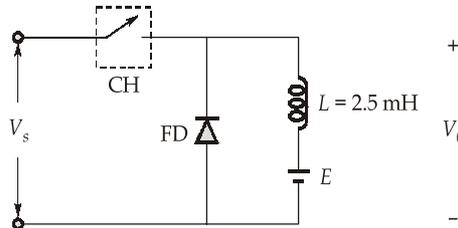
$$I_{0 \text{ rms}} = \frac{V_{0 \text{ rms}}}{R} = \frac{126.67}{10} = 12.667$$

Power consumed by the load,

$$= (12.667)^2 \times 10$$

$$= 1604.55 \text{ W} = 1.604 \text{ kW}$$

189. 1.87 (1.60 to 1.99)



Average output voltage is given by,

$$V_0 = \alpha V_s$$

$$\alpha = \text{duty cycle} = 0.25$$

$$V_0 = 0.25 \times 200 = 50 \text{ V}$$

As the average value of voltage drop across  $L$  is zero

$$V_0 = E = \alpha V_s = 50 \text{ V}$$

During  $T_{ON}$ , the difference in source voltage  $V_s$  and load emf  $E$  appears across inductance  $L$ . Also during  $T_{ON}$ , the current through  $L$  rises from  $I_{min}$  to  $I_{max}$

$$\therefore V_L = L \frac{di}{dt}$$

$$V_s - E = L \frac{\Delta I}{T_{ON}}$$

$$T_{ON} = \frac{L \cdot \Delta I}{V_s - E} = \frac{2.5 \times 8 \times 10^{-3}}{200 - 50}$$

$$= 1.33 \times 10^{-4} \text{ sec} = 133 \mu\text{sec}$$

and, duty cycle,  $\alpha = \frac{T_{ON}}{T}$

$$\frac{\alpha}{T_{ON}} = f$$

Chopping frequency,  $f = \frac{\alpha}{T_{ON}} = \frac{0.25}{133 \times 10^{-6}} = 1.879 \text{ kHz}$

190. (b)

Output voltage Buck-boost converter,

$$V_0 = -\frac{D}{1-D} V_s$$

$$V_0 = \frac{-0.4}{1-0.4} \times 24 = -16 \text{ V}$$

Average current through inductor,

$$I_L = \frac{V_S D}{R(1-D)^2} = \frac{24 \times 0.4}{5(1-0.4)^2}$$

$$I_L = 5.33 \text{ A}$$

$$\Delta i_L = \frac{V_S D T}{L} = \frac{24 \times 0.4}{20 \times 10^{-6} \times 100 \times 10^3} = 4.8 \text{ A}$$

$$I_{L, \max} = I_L + \frac{\Delta i_L}{2} = 5.33 + \frac{4.8}{2} = 7.73 \text{ A}$$

$$I_{L, \min} = I_L - \frac{\Delta i_L}{2} = 5.33 - \frac{4.8}{2} = 2.93 \text{ A}$$

The ratio of inductor currents,

$$\frac{I_{L, \max}}{I_{L, \min}} = \frac{7.73}{2.93} = 2.63$$

191. (d)

Current distortion factor for single phase semi converter,

$$\begin{aligned} \text{CDF} &= \frac{2\sqrt{2} \cos\left(\frac{\alpha}{2}\right)}{\sqrt{\pi(\pi - \alpha)}} \\ &= \frac{2\sqrt{2} \cos\left(\frac{45}{2}\right)}{\sqrt{\pi\left(\pi - \frac{\pi}{4}\right)}} = \frac{2.613}{\sqrt{\frac{3\pi^2}{4}}} = \frac{2.613}{2.721} = 0.960 \end{aligned}$$

192. (c)

From boundary condition of dielectric-dielectric medium.

$$E_{t1} = E_{t2}$$

and

$$D_{n1} = D_{n2}$$

$$\epsilon_{r1} E_{n1} = \epsilon_{r2} E_{n2}$$

or

$$E_{n2} = \frac{\epsilon_{r1}}{\epsilon_{r2}} E_{n1}$$

$$= \frac{2}{8} \times 100 = 25$$

∴

$$\vec{E}_2 = 25\hat{a}_x + 200\hat{a}_y - 50\hat{a}_z$$

193. (c)

The flux in the circuit is,

$$\Psi = \frac{\ddot{o}}{U} = \frac{N_i i_1}{l/\mu S} = \frac{N_1 i_1 \mu S}{2\pi\rho_0}$$

$\ddot{o}$  = magneto motive force

$U$  = reluctance

$l$  = mean length

$S$  = cross-sectional area of magnetic core

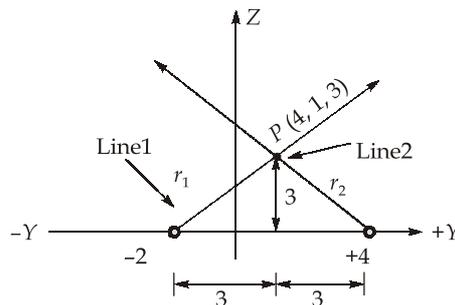
According to Faraday's Law, the emf induced in the second coil is,

$$V_2 = -N_2 \frac{d\Psi}{dt} = -\frac{N_1 N_2 \mu S}{2\pi\rho_0} \frac{di_1}{dt}$$

$$\begin{aligned} V_2 &= -\frac{100 \times 200 \times 500 \times (4\pi \times 10^{-7}) \times 10^{-3} \times 300\pi \cos 100\pi t}{2\pi(10 \times 10^{-2})} \\ &= -6\pi \cos 100\pi t \text{ V} \end{aligned}$$

194. (d)

Let  $r_1$  and  $r_2$  be the directed line segments from the lines 1 and 2 respectively to the point P (in Y-Z plane).



Then,

$$r_1 = 3\hat{a}_y + 3\hat{a}_z$$

and

$$r_2 = -3\hat{a}_y + 3\hat{a}_z$$

$$E_1 = \frac{\lambda}{2\pi\epsilon_0 r_1} \left( \frac{3\hat{a}_y + 3\hat{a}_z}{r_1} \right) \quad \dots (i)$$

where,

$$r_1^2 = 3^2 + 3^2 = r_2^2 \text{ (in magnitude)} = 18 = r^2$$

$$E_2 = \frac{\lambda}{2\pi\epsilon_0 r_2} \left( \frac{-3\hat{a}_y + 3\hat{a}_z}{r_2} \right) \quad \dots (ii)$$

Adding (i) and (ii), we obtain the resultant field

$$\begin{aligned} E &= E_1 + E_2 = \frac{\lambda}{2\pi\epsilon_0 r^2} (2 \times 3\hat{a}_z) \quad \text{(replacing } r_1 \text{ and } r_2 \text{ by } r) \\ &= \frac{5 \times 10^{-9}}{2\pi\epsilon_0 (18)} (6\hat{a}_z) = 30\hat{a}_z \text{ V/m} \end{aligned}$$

195. (180)

From Maxwell's equation,

$$\nabla \cdot \vec{D} = \rho_v = \frac{\partial}{\partial x}(12xyz)\hat{a}_x + \frac{\partial}{\partial y}(6x^2z + 6yz)\hat{a}_y + \frac{\partial}{\partial z}(6x^2y + 3y^2)\hat{a}_z$$

$$\nabla \cdot \vec{D} = 12yz + 6z$$

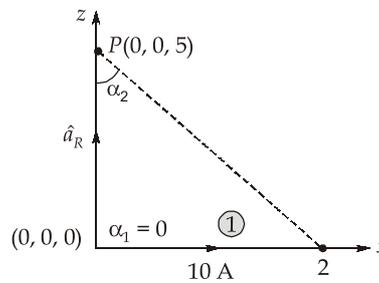
At point (1, 1, 1)  $\nabla \cdot \vec{D} = 12 + 6 = 18 \text{ C/m}^2$

$$Q = \int \rho_v dv$$

$$= 18 \times 10^{-5} = 180 \mu\text{C}$$

196. 118.21 (117.21 to 119.21)

Note here that the point of interest is on z-axis and the side-1 of the triangular is on x-axis. So no need to consider other sides.



The perpendicular from point  $P$  on the current filament is at  $(0, 0, 0)$ . The vector and unit vector in the direction of perpendicular towards point  $P$  is,

$$\vec{R} = 5\hat{a}_z$$

and  $\hat{a}_R = \hat{a}_z$

The angles made by ends of the filament with the perpendicular are (dotted line),

$$\alpha_1 = 0;$$

$$\alpha_2 = \tan^{-1}\left(\frac{2}{5}\right) = 21.80^\circ$$

The current filament along  $x$ -axis gives

$$\hat{a}_l = \hat{a}_x$$

Now the direction of  $H$  is,

$$\hat{a}_\phi = \hat{a}_l \times \hat{a}_R = \hat{a}_x \times \hat{a}_z = -\hat{a}_y$$

The field intensity due to finite filament is,

$$\vec{H} = \frac{I}{4\pi R} [\sin \alpha_2 - \sin \alpha_1] \hat{a}_\phi$$

$$= \frac{20}{4\pi \times (5)} [\sin(21.80^\circ) - \sin(0)] (-\hat{a}_y)$$

$$\vec{H} = -118.21 \hat{a}_y \text{ (mA/m)}$$

197. (a)

$$\mu_1 = 2 \mu_0$$

$$\mu_2 = 5 \mu_0$$

$$B_2 = 10\hat{a}_\rho + 15\hat{a}_\phi - 20\hat{a}_z \text{ mWb/m}^2$$

$$B_{1n} = B_{2n} = 15\hat{a}_\phi$$

$$H_{1t} = H_{2t}$$

$$B_{1t} = \frac{\mu_1}{\mu_2} B_{2t} = \frac{2}{5}(10\hat{a}_\rho - 20\hat{a}_z)$$

$$B_{1t} = (4\hat{a}_\rho - 8\hat{a}_z) \text{ mWb/m}^2$$

$$\begin{aligned} W_{m1} &= \frac{1}{2} B_1 \cdot H_1 = \frac{B_1^2}{2\mu_1} \\ &= \frac{(4^2 + 15^2 + 8^2) \times 10^{-6}}{2 \times 2 \times 4\pi \times 10^{-7}} \\ &= \frac{305}{16\pi} \times 10 = 60.68 \text{ J/m}^3 \end{aligned}$$

198. (b)

$$H = \frac{I}{4\pi\rho} (\cos\alpha_2 - \cos\alpha_1)\hat{a}_\phi$$

$$\rho = \sqrt{5^2 + 5^2} = \sqrt{50} \text{ m}$$

$$= \frac{2}{4\pi\sqrt{5^2 + 5^2}} [\cos\alpha_2 - \cos\alpha_1]\hat{a}_\phi$$

$$\cos\alpha_2 = \frac{10}{\sqrt{50+100}} = \frac{10}{\sqrt{150}}$$

$$\cos\alpha_1 = \cos 90^\circ = 0$$

$$\hat{a}_\phi = \hat{a}_l \times \hat{a}_\rho = \hat{a}_z \times \left( \frac{5\hat{a}_x + 5\hat{a}_y}{5\sqrt{2}} \right) = \left( \frac{-\hat{a}_x + \hat{a}_y}{\sqrt{2}} \right)$$

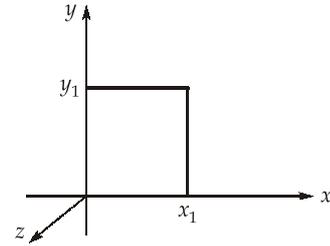
$$\vec{H} = \frac{2}{4\pi \times 5\sqrt{2}} \times \left( \frac{10}{\sqrt{150}} - 0 \right) \times \left( \frac{-\hat{a}_x + \hat{a}_y}{\sqrt{2}} \right)$$

$$= \frac{1}{20\pi} (-\hat{a}_x + \hat{a}_y) \times \frac{10}{5\sqrt{6}}$$

$$= \frac{1}{10\pi\sqrt{6}} (-\hat{a}_x + \hat{a}_y) \text{ A/m}$$

199. (b)

$$\begin{aligned} x_1 &= y_1 = 1 \text{ m} \\ B_0 &= \sin \pi x \sin \pi y T \\ B &= B_0 \cos \omega_0 t \\ E(t) &= \iint_s \frac{-dB}{dt} \cdot \hat{n} dS \\ &= \iint_s \omega_0 B_0 \sin \omega t \hat{n} dS \end{aligned}$$



$$E_{\max} = \omega_0 \int_{y=0}^1 \int_{x=0}^1 \sin \pi x \sin \pi y dx dy$$

$$E_{\max} = 1000 \times 2\pi \times \frac{4}{\pi^2} = \frac{8000}{\pi} \text{ V/turns}$$

$$E_{\text{rms}} = \frac{1}{\sqrt{2}} \times \frac{8000}{\pi} \times 10 = 18 \text{ kV}$$

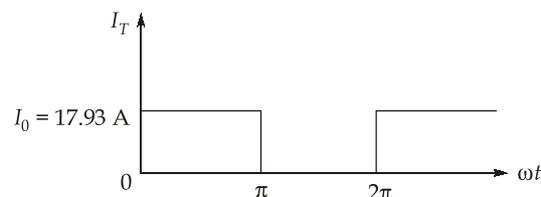
200. (a, b)

Average output voltage of single phase full converter is,

$$V_0 = \frac{2V_m}{\pi} \cos \alpha$$

$$V_0 = \frac{2\sqrt{2} \times 230}{\pi} \cos 30^\circ = 179.33 \text{ V}$$

$$I_0 = \frac{V_0}{R} = \frac{179.33}{10} = 17.93 \text{ A}$$



The average value of thyristor current,

$$I_{T \text{ avg}} = 17.93 \times \frac{\pi}{2\pi} = 8.96 \text{ A}$$

The rms value of thyristor current,

$$I_{T \text{ rms}} = 17.93 \times \sqrt{\frac{\pi}{2\pi}} = 12.67 \text{ A}$$

