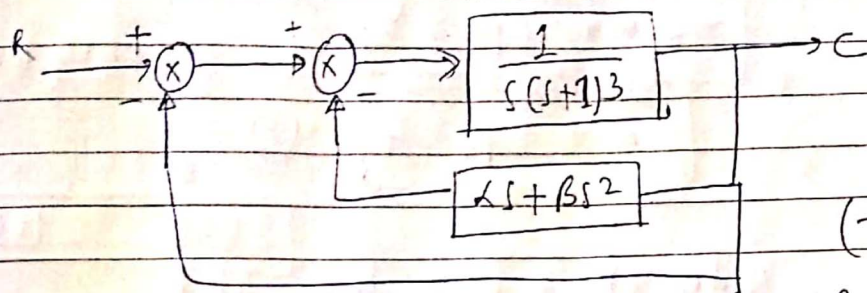


NAME +

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$(-1 \pm 2j)$
↓
Poles of system

$$\frac{C}{R} = \frac{1}{s(s+1)^3}$$

$$1 - \left[\frac{\alpha s + \beta s^2}{s(s+1)^3} - \frac{1}{s(s+1)^3} \right]$$

$$= \frac{1}{s(s+1)^3} \Rightarrow \frac{1}{s(s+1)^3 + \alpha s + \beta s^2 + 1}$$

$$q(s) \Rightarrow s(s+1)^3 + \alpha s + \beta s^2 + 1 = 0$$

Put $s = -1 + 2j$ & $-1 - 2j$ in $q(s)$.

$$(-1+2j)(-1+2j+1)^3 + \alpha(-1+2j) + \beta(-1+2j)^2 + 1 = 0$$

$$-8j(-1+2j) + (-\alpha) + 2\alpha j + \beta(1-4-4j) + 1 = 0$$

$$8j + 16 + 2\alpha j - \alpha - 3\beta - 4\beta j + 1 = 0$$

$$(17 - \alpha - 3\beta) + j(8 + 2\alpha - 4\beta) = 0$$

$$17 - \alpha - 3\beta = 0$$

$$\boxed{\alpha + 3\beta = 17} \quad \text{--- (1)}$$

DATE

$$8 + 2\alpha - 4\beta = 0$$

$$\boxed{-2\alpha + 4\beta = 8}$$

$$\boxed{-\alpha + 2\beta = 4} \quad \text{--- (2)}$$

Add (1) x (2), we get

$$5\beta = 21 \Rightarrow \boxed{\beta = 21/5}$$

put $\Rightarrow \beta = 21/5$ in (2), we get

$$-\alpha + 2\left(\frac{21}{5}\right) = 4$$

$$-\alpha = 4 - \frac{42}{5} = \frac{20 - 42}{5} = -\frac{22}{5}$$

$$\boxed{\alpha = \frac{22}{5}}$$



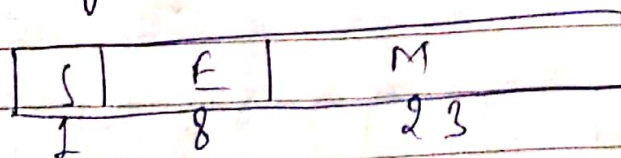
12

(b)

solⁿ:-

$$(-307.1875)_{10} \Rightarrow -(\quad)$$

single precision format (32 bit)
format



↓
excess 127 format

$$S = 1 \text{ (since negative)}$$



2	307	
2	153	1
2	76	1
2	38	0
2	19	0
2	9	1
2	4	1
2	2	0
2	1	0
	0	1

$$0.1875 \times 2 = 0.375$$

$$0.375 \times 2 = 0.75$$

$$0.75 \times 2 = 1.5$$

$$0.5 \times 2 = 1.0$$

$$(307.1875)_{10} = (100110011.0011)_2$$

$$= 1 \cdot [001100110011] \times 2^8$$

Mantissa (1. normalized form)

$$\text{exponent} = 8 + 127 = 135 = (10000111)_2$$

2	135	
2	67	1
2	33	1
2	16	1
2	8	0
2	4	0
2	2	0
2	1	0
	0	1

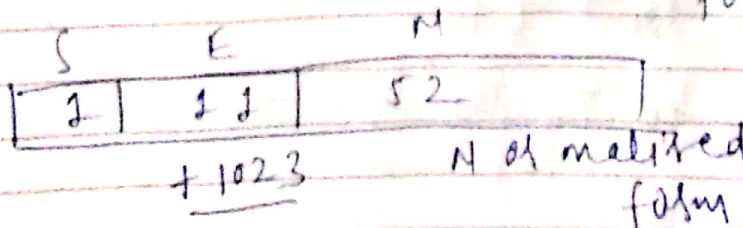
$$1 \quad 10000111 \quad 001100110011$$

11 times

$$(C3999800)_H$$



double precision format (64 bit) format



$$-(307.1875)_{10} =$$

$S = 1$ ($\because$ negative number)

$$(307.1875)_{10} = 1 \cdot \underbrace{001100110011}_{\text{mantissa}} \times 2^8$$

$$E = 8 + 1023 = 1031 = (10000000111)_2$$

2	1031	
2	515	1
2	257	1
2	128	1
2	64	0
2	32	0
2	16	0
2	8	0
2	4	0
2	2	0
2	1	0
	0	1

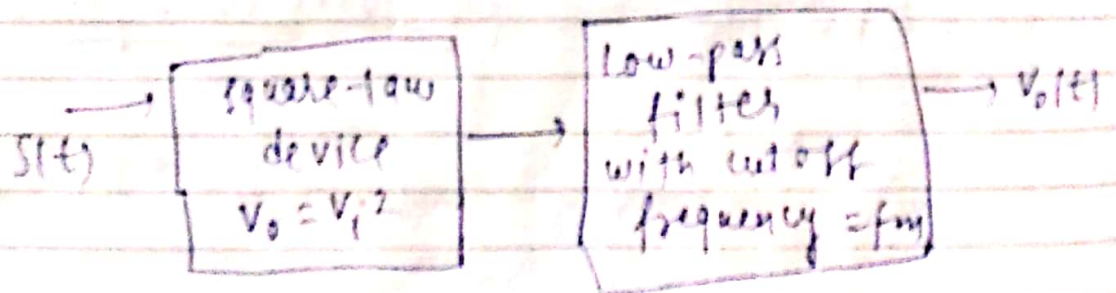
S	E
1	10000000111
	001100110011

(C 0733300000000000)



11

③
 $\sin^2 \omega_c t$



$$S(t) = [1 + m(t)] \cos \omega_c t$$

$$S \& L \& D \& O/P = V_o = V_i^2 = [(1 + m(t)) \cos \omega_c t]^2$$

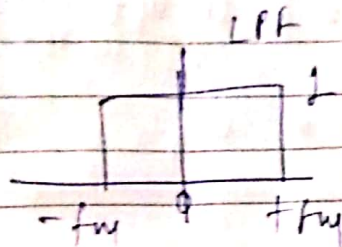
$$V_o \rightarrow [1 + m^2(t) + 2m(t)] \cos^2 \omega_c t$$

$$V_o \rightarrow \cos^2 \omega_c t + m^2(t) \cos^2 \omega_c t + 2m(t) \cos^2 \omega_c t$$

$$\cos^2 \omega_c t = \frac{1 + \cos 2\omega_c t}{2}$$

DATE

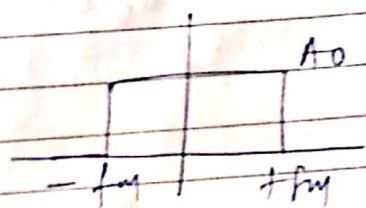
$$S.D. \text{ of } r = \frac{1}{2} + \frac{1}{2} \cos 2\omega_c t + m^2(t) \left[\frac{1 + \cos 2\omega_c t}{2} \right] + 2m(t) \left[\frac{1 + \cos 2\omega_c t}{2} \right]$$



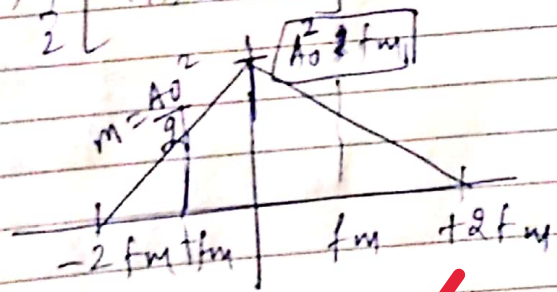
$$V_o(t) = \frac{1}{2} + \frac{m^2(t)}{2} + m(t)$$

(o/p of filter)

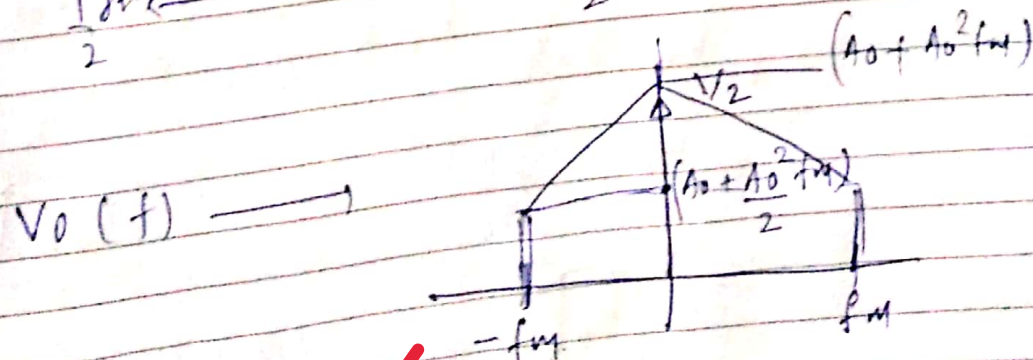
$$m(t) \xrightarrow{F.T} M(f)$$



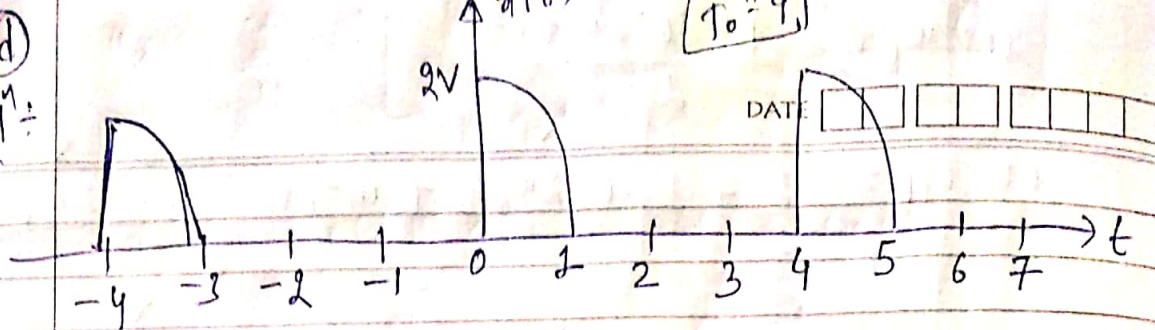
$$\frac{1}{2} m(t) \cdot m(t) \rightarrow \frac{1}{2} [M(f) \otimes M(f)]$$



$$\frac{1}{2} \delta(f) \rightarrow \frac{1}{2} \delta(f)$$



(d) Soln:



$T_0 = 4$

$$x(t) = \begin{cases} 2 \cos \omega_0 t & 0 \leq t \leq 1 \\ 0 & 1 \leq t \leq 4 \end{cases}$$

$$x(t) = a_0 + \sum_{n=1}^{\infty} a_n \cos n\omega_0 t + b_n \sin n\omega_0 t$$

$$a_0 = \frac{1}{T_0} \int_0^{T_0} x(t) dt$$

$$a_0 = \frac{1}{4} \int_0^4 x(t) dt$$

$$= \frac{1}{4} \left[\int_0^1 2 \cos \omega_0 t dt + \int_1^4 0 dt \right]$$

$$\Rightarrow \frac{2}{4} \left[\frac{\sin \omega_0 t}{\omega_0} \right]_0^1$$

$$\Rightarrow \frac{1}{2} \left[\frac{\sin \omega_0 - \sin 0}{\omega_0} \right]$$

$$\omega_0 = \frac{2\pi}{T_0} = \frac{2\pi}{4} = \boxed{\pi/2}$$

$$\boxed{a_0} = \frac{1}{2} \frac{\sin(\pi/2)}{\frac{\pi}{2}} \Rightarrow \boxed{\frac{1}{\pi}}$$

$$a_n = \frac{2}{T_0} \int_0^{T_0} x(t) \cos(n\omega_0 t) dt$$

DATE

$$a_n = \frac{2}{24T_0} \int_0^{24} x(t) \cos(n\omega_0 t) dt$$

$$a_n = \frac{1}{2} \left[\int_0^1 \cos \omega_0 t \cos(n\omega_0 t) dt \right]$$

$$a_n = \frac{1}{2} \int_0^1 [\cos(n+1)\omega_0 t + \cos(n-1)\omega_0 t] dt$$

$$2 \cos A \cos B = \cos(A+B) + \cos(A-B)$$

$$a_n = \frac{1}{2} \left[\left. \frac{\sin(n+1)\omega_0 t}{(n+1)\omega_0} \right|_0^1 + \left. \frac{\sin(n-1)\omega_0 t}{(n-1)\omega_0} \right|_0^1 \right]$$

$$a_n = \frac{1}{2} \left[\frac{\sin(n+1)\omega_0 \times 1}{(n+1)\omega_0} - \frac{\sin(n+1)\omega_0 \times 0}{(n+1)\omega_0} \right.$$

$$\left. + \frac{\sin(n-1)\omega_0 \times 1}{(n-1)\omega_0} - \frac{\sin(n-1)\omega_0 \times 0}{(n-1)\omega_0} \right]$$

$$a_n = \frac{1}{2\omega_0} \left[\frac{\sin(n+1)\omega_0}{n+1} + \frac{\sin(n-1)\omega_0}{n-1} \right]$$

$$a_n = \frac{1}{2 \times \frac{\pi}{24}} \left[\frac{\sin\left(\frac{\pi}{2} + \frac{\pi n}{2}\right)}{n+1} + \frac{\sin\left(\frac{\pi n}{2} - \frac{\pi}{2}\right)}{n-1} \right]$$

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$$a_n = \frac{1}{\pi} \left[\frac{\cos\left(\frac{\pi}{2}n\right)}{n+1} - \frac{\cos\left(\frac{\pi}{2}n\right)}{n-1} \right]$$

$$a_n = \frac{\cos\left(\frac{\pi}{2}n\right)}{\pi} \left[\frac{n-1 - (n+1)}{n^2-1} \right]$$

$$a_n = \frac{-2}{\pi} \frac{\cos\left(\frac{\pi}{2}n\right)}{n^2-1} \quad \checkmark$$

$$b_n = \frac{2}{T_0} \int_0^{T_0} x(t) \sin(n\omega_0 t) dt$$

$$b_n = \frac{2}{4} \int_0^4 x(t) \sin(n\omega_0 t) dt$$

$$b_n = \frac{1}{2} \left[\int_0^1 2 \cos\omega_0 t \sin(n\omega_0 t) dt \right]$$

$$b_n = \frac{1}{2} \left[\int_0^1 [\sin(n+1)\omega_0 t + \sin(n-1)\omega_0 t] dt \right]$$

$$2 \sin A \cos B = \sin(A+B) + \sin(A-B)$$

$$b_n = \frac{1}{2} \left[\left[\frac{-\cos(n+1)\omega_0 t}{(n+1)\omega_0} \right]_0^1 + \left[\frac{-\cos(n-1)\omega_0 t}{(n-1)\omega_0} \right]_0^1 \right]$$

$$b_n = \frac{1}{2} \left[\frac{-\cos(n+1)\omega_0 \times 1 + \cos 0}{(n+1)\omega_0} \right.$$

$$\left. - \frac{\cos(n-1)\omega_0 \times 1 + \cos 0}{(n-1)\omega_0} \right]$$

classmate

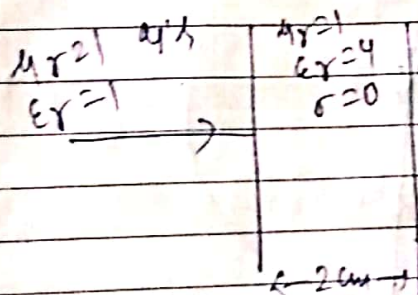
$$b_n = \frac{1}{2w_0} \left[\frac{\sin(n\pi/2)}{(n+1)} + \frac{1}{n+1} - \frac{\sin(n\pi/2)}{n-1} + \frac{1}{n-1} \right]$$

$$b_n = \frac{1}{\pi} \left[\sin(n\pi/2) \left[\frac{1}{n+1} - \frac{1}{n-1} \right] + \frac{1}{n+1} + \frac{1}{n-1} \right]$$

$$b_n = \frac{1}{\pi} \left[\frac{-2 \sin(n\pi/2)}{n^2-1} + \frac{2n}{n^2-1} \right]$$

$$b_n = \frac{2}{\pi(n^2-1)} [n - \sin(n\pi/2)]$$

Q
101



$$f = 3 \text{ GHz}$$

$$\Gamma = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} = \frac{\sqrt{\frac{\mu_r \mu_0}{\epsilon_r \epsilon_0}} - \sqrt{\frac{\mu_0}{\epsilon_0}}}{\sqrt{\frac{\mu_r \mu_0}{\epsilon_r \epsilon_0}} + \sqrt{\frac{\mu_0}{\epsilon_0}}}$$

$$= \frac{\frac{1}{\sqrt{4}} - 1}{\frac{1}{\sqrt{4}} + 1} = \frac{\frac{1}{2} - 1}{\frac{1}{2} + 1} = \frac{-\frac{1}{2}}{\frac{3}{2}}$$

$$\Gamma = -\frac{1}{3} = \frac{E_r}{E_i}$$

$$E_r = -\frac{300}{3} = -100 \text{ mV/m}$$

$$|E_r| = 100 \text{ mV/m}$$

classmate

PAGE

$$Z = 1 + \Gamma$$

$$Z = 1 - 1/3 = 2/3$$

$$= \frac{E_f}{K_f}$$

$$E_f = \frac{2}{3} \times 300 = 200$$

$$200 \text{ mV/m}$$

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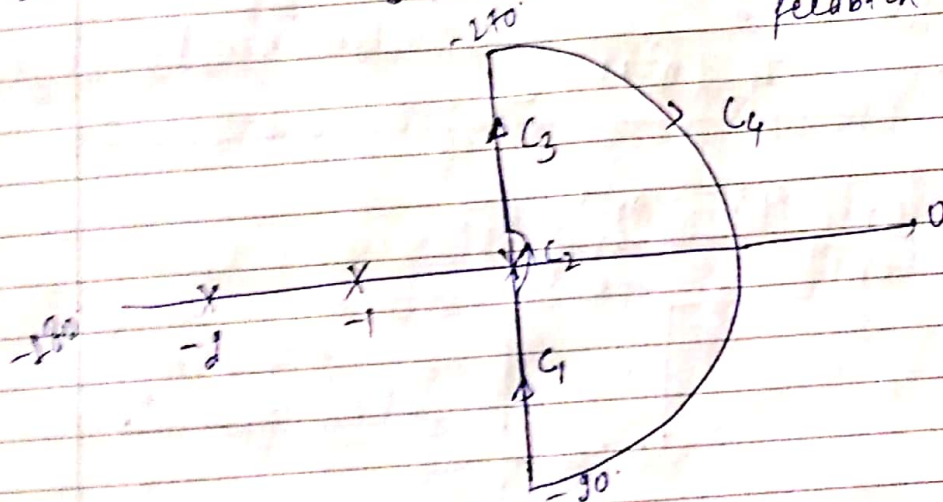
Q2.

(3)

Q2.

$$G(s) = \frac{K}{s(s+1)(s+2)}$$

(open loop transfer function of unity feedback control system)



Nyquist contour

$$C_1: s = -j\omega, \quad \theta = -90^\circ + 0 + 90^\circ, \quad \omega = \infty \rightarrow 0$$

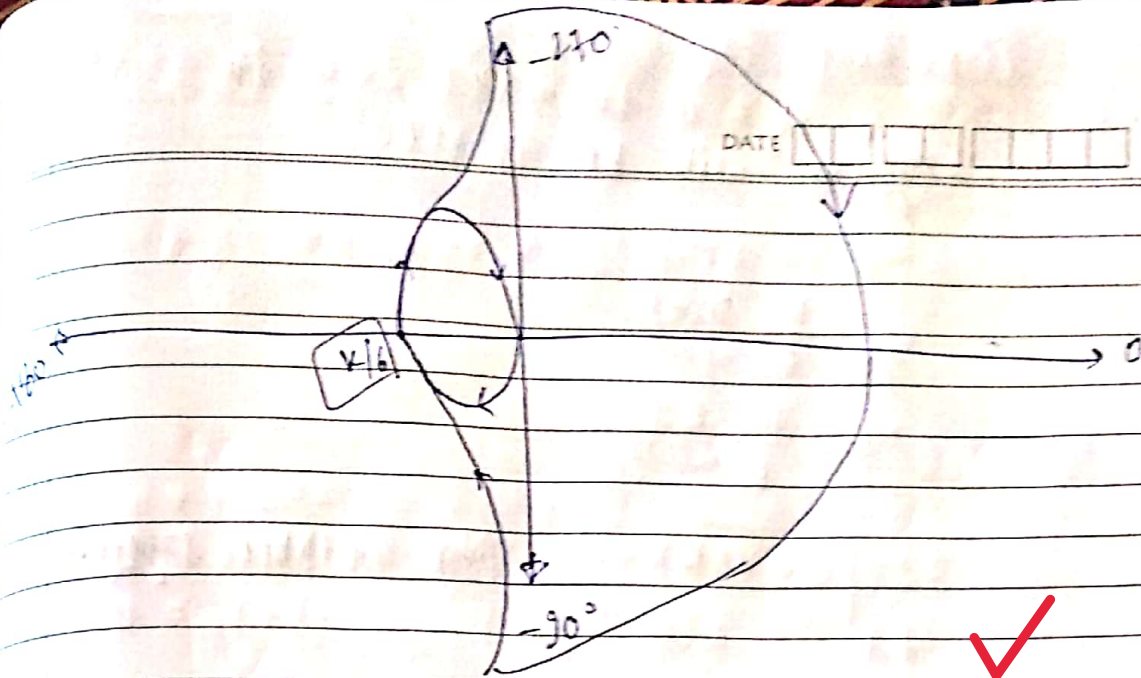
$$C_2: s = \omega e^{j\theta}, \quad \theta = -90^\circ + 0 + 90^\circ$$

$$C_3: s = +j\omega, \quad \omega = 0 \rightarrow \infty$$

$$C_4: s = \omega e^{j\theta}, \quad \theta = +90^\circ - 90^\circ$$

ω	M	$\phi = -90^\circ - \tan^{-1}(\omega) - \tan^{-1}(\omega/2)$
0	∞	-90°
∞	0	-270°

PAGE



$$-180^\circ = -90^\circ - \tan^{-1}(\omega) - \tan^{-1}(\omega/2)$$

$$90^\circ = \tan^{-1}(\omega) + \tan^{-1}(\omega/2) \quad \left| \begin{array}{l} \tan^{-1}x + \tan^{-1}y \\ = \tan^{-1}\left(\frac{x+y}{1-xy}\right) \end{array} \right.$$

$$90^\circ = \tan^{-1}\left(\frac{\omega + \omega/2}{1 - \omega^2/2}\right)$$

$$\frac{1 - \omega^2}{2} = 0 \quad \left| \omega_{pc} = \sqrt{2} \text{ rad/sec} \right.$$

$$M = \frac{K}{\omega \sqrt{1+\omega^2} \sqrt{4+\omega^2}} = \frac{K}{\sqrt{2} \sqrt{3} \sqrt{6}}$$

$$M = \frac{K}{6}$$

(ii)

$$N = P - Z$$

N = No of encirclement

P = open-loop poles on the right side

Z = closed-loop poles on the right side

PAGE

$$P = 0,$$

$\therefore N = 0$ to make $Z = 0$, so
system will be stable.

DATE

$$\frac{-K}{6} > -1$$

$$\frac{K}{6} < 1$$

$0 < K < 6$: for stable system.

(iii)
for $\pm$

$$G_M \Rightarrow M \Big|_{\omega = \omega_{pc}} = \frac{K}{6}$$

$$20 \log_{10} \frac{1}{K/6} = 20 \log_{10} \frac{6}{K} = 0$$

$$\frac{6}{K} = 1$$

$$K = 6$$

18

(b)

80171

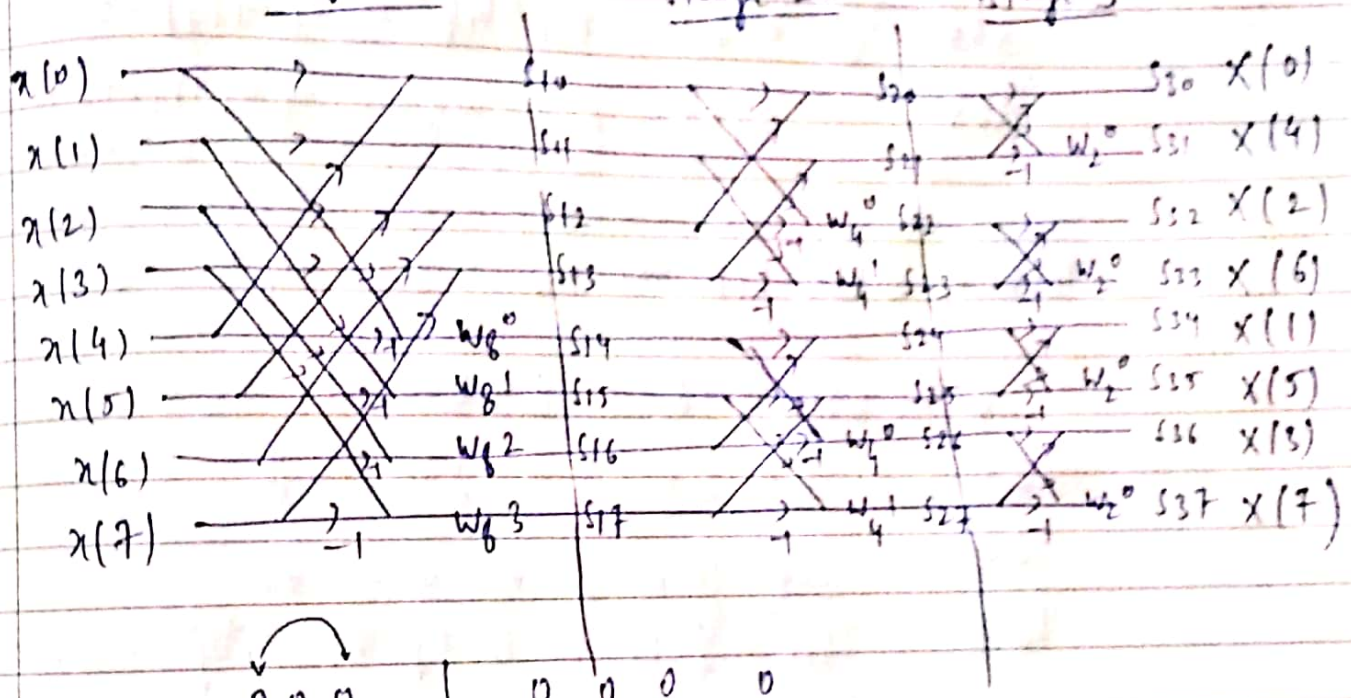
Decimation in frequency FFT algorithm

$$x(n) = \{ \overset{0}{1}, \overset{1}{-1}, \overset{2}{-1}, \overset{3}{-1}, \overset{4}{1}, \overset{5}{1}, \overset{6}{1}, \overset{7}{-1} \}$$

Stage-1

Stage-2

Stage-3



0 0 0	0	0	0	0
0 0 1	1	0	0	-4
0 1 0	0	1	0	2
0 1 1	1	1	0	6
1 0 0	0	0	1	1
1 0 1	1	0	1	5
1 1 0	0	1	1	3
1 1 1	1	1	1	7



classmate

First stage of P

$$W_8^0 = 1$$

DATE

$$W_8^1 = \frac{0-j}{\sqrt{2}}$$

$$W_8^2 = -1$$

$$W_8^3 = \frac{-(1+j)}{\sqrt{2}}$$

$$S_{10} = 2(0) + 2(4)W_8^0 = 1+1 = 2$$

$$S_{11} = 2(1) + 2(5) = -1+1 = 0$$

$$S_{12} = 2(2) + 2(6) = -1+1 = 0$$

$$S_{13} = 2(3) + 2(7) = -1-1 = -2$$

$$S_{14} = 2(0) - 2(4)W_8^0 = 1-1 = 0$$

$$S_{15} = 2(1) - 2(5)W_8^1 = (-1) - 1\left(\frac{1-j}{\sqrt{2}}\right) = \frac{-(1+j)}{\sqrt{2}}$$

$$S_{16} = 2(2) - 2(6)W_8^0 = -1-1(j) = -1-j$$

$$S_{17} = 2(3) - 2(7)W_8^1 = -1 - (1)\left(\frac{1-j}{\sqrt{2}}\right) = \frac{-(1+j)}{\sqrt{2}}$$

Second stage of P $W_4^0 = 1$ $W_4^1 = j$

$$S_{20} = S_{10} + S_{12} = 2+0 = 2$$

$$S_{21} = S_{11} + S_{13} = 0-2 = -2$$

$$S_{22} = S_{10} - S_{12} = 2-0 = 2$$

$$S_{23} = S_{11} - jS_{13} = 0 - j(-2) = 2j$$

$$S_{24} = S_{14} + S_{16} = 0 + (-1-j) = -1-j$$

$$S_{25} = S_{15} + S_{17} = -\left(\frac{2+j}{\sqrt{2}}\right)$$

$$S_{26} = S_{14} - S_{16} = 1-j$$

$$S_{27} = S_{15} - jS_{17} = \frac{-(1+j)(1-j) + j-1}{\sqrt{2}}$$

$$\rightarrow -\left(\frac{(2+j)}{\sqrt{2}}\right) + j\left(\frac{(2+j)}{\sqrt{2}}\right)$$

classmate

PAGE

Third stage output $w_2'' \Rightarrow$

$$S_{30} = S_{20} + S_{21} = 2 - 2 \Rightarrow 0 \quad \boxed{} \boxed{} \boxed{} \boxed{} \boxed{} \boxed{} \boxed{} \boxed{}$$

$$S_{31} = S_{20} - S_{21} = 4$$

$$S_{32} = S_{22} + S_{23} = 2 + 2j$$

$$S_{33} = S_{22} - S_{23} = 2 - 2j$$

$$S_{34} = S_{24} + S_{25} = -(3 + \sqrt{2}) + j$$

$$S_{35} = S_{24} - S_{25} = (1 + \sqrt{2}) + j$$

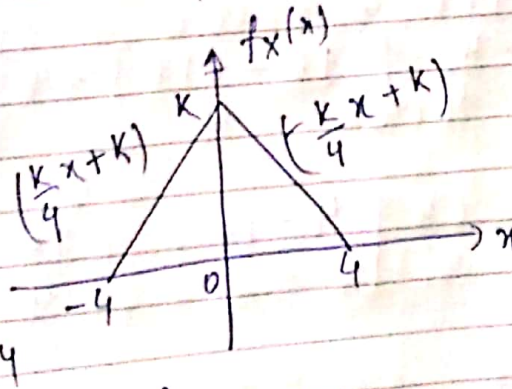
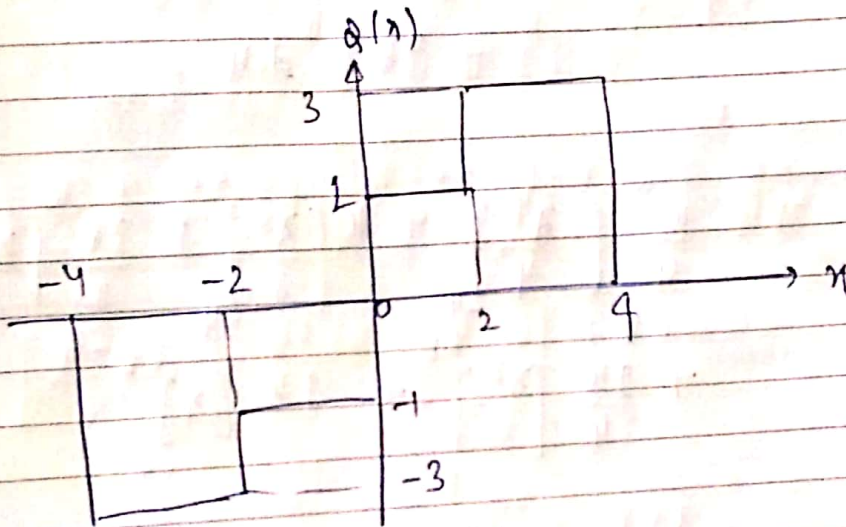
$$S_{36} = S_{26} + S_{27} = -\sqrt{2} + \sqrt{2}j$$

$$S_{37} = S_{26} - S_{27} = (2 + \sqrt{2}) - j(2 + \sqrt{2})$$

14

(c)

5017-



$$Q = \int_{-4}^4 (x-x_0)^2 f(x) dx$$

$$Q = \int_{-4}^{-2} (x+3)^2 \left(\frac{k}{4}\right) dx + \int_{-2}^0 (x+3)^2 \left(\frac{k}{4}\right) dx + \int_0^2 (x-1)^2 \left(\frac{k}{4}\right) dx + \int_2^4 (x-1)^2 \left(\frac{k}{4}\right) dx$$

$$+ \int_{-4}^{-2} (x+3)^2 \left(\frac{k}{4}\right) dx + \int_{-2}^0 (x+3)^2 \left(\frac{k}{4}\right) dx + \int_0^2 (x-1)^2 \left(\frac{k}{4}\right) dx + \int_2^4 (x-1)^2 \left(\frac{k}{4}\right) dx$$

classmate

$$q_1 = \int_{-4}^{-2} (x+3)^2 \left(\frac{k}{4}x + k \right) dx$$

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$$= \int_{-4}^{-2} (x^2 + 6x + 9) \left(\frac{k}{4}x + k \right) dx$$

$$\Rightarrow \int_{-4}^{-2} \left[\frac{k}{4}x^3 + kx^2 + \frac{6k}{4}x^2 + 6kx + \frac{9k}{4}x + 9k \right] dx$$

$$\Rightarrow \frac{k}{4} \left[\frac{x^4}{4} \right]_{-4}^{-2} + k \left[\frac{x^3}{3} \right]_{-4}^{-2} + \frac{3k}{2} \left[\frac{x^3}{3} \right]_{-4}^{-2} + 6k \left[\frac{x^2}{2} \right]_{-4}^{-2} + \frac{9k}{4} \left[\frac{x^2}{2} \right]_{-4}^{-2} + 9k \left[x \right]_{-4}^{-2}$$

$$\Rightarrow \frac{k}{4} \left[\frac{16 - 256}{4} \right] + k \left[\frac{-8 + 64}{3} \right] + \frac{3k}{2} \left[\frac{-8 + 64}{3} \right]$$

$$+ 6k \left[\frac{4 - 16}{2} \right] + \frac{9k}{4} \left[\frac{4 - 16}{2} \right] + 9k \left[-2 + 4 \right]$$

$$\Rightarrow -15k + 18.66k + 28k - 36k - 13.5k + 18k$$

$$\Rightarrow \boxed{18k - 17.84k} = 0.16k$$



$$q_2 = \int_{-2}^0 (x^2 + 2x + 1) \left(\frac{k}{4}x + k \right) dx$$

$$\Rightarrow \int_{-2}^0 \left(\frac{k}{4}x^3 + x^2k + \frac{2k}{4}x^2 + 2kx + \frac{k}{4}x + k \right) dx$$

classmate

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$$1 \left[\frac{k}{4} \left[\frac{x^4}{4} \right]_0^2 + k \left[\frac{x^3}{3} \right]_0^2 + \frac{k}{2} \left[\frac{x^2}{2} \right]_0^2 + 2k \left[\frac{x}{2} \right]_0^2 + \frac{k}{4} \left[x \right]_0^2 + k \left[1 \right]_0^2 \right]$$

$$= \frac{k}{4} \left[\frac{0-16}{4} \right] + k \left[\frac{0-(-8)}{3} \right] + \frac{k}{2} \left[\frac{0-(-2)}{2} \right] + 2k \left[\frac{0-1}{2} \right] + \frac{k}{4} \left[\frac{0-4}{2} \right] + k \left[0+2 \right]$$

$$2) -k + \frac{8}{3}k + \frac{8k}{6} - 4k - \frac{k}{2} + 2k$$

$$2) \left[\frac{k}{2} \right]$$

$$[q_3] = \int_0^2 (x+1)^2 \left(-\frac{k}{4}x + k \right) dx$$

$$q_3 = \int_0^2 (x^2 - 2x + 1) \left(-\frac{k}{4}x + k \right) dx$$

$$q_3 = \int_0^2 \left(-\frac{k}{4}x^3 + kx^2 + \frac{2k}{4}x^2 - 2kx - \frac{k}{4}x + k \right) dx$$

$$q_3 = -\frac{k}{4} \left[\frac{x^4}{4} \right]_0^2 + k \left[\frac{x^3}{3} \right]_0^2 + \frac{k}{2} \left[\frac{x^3}{3} \right]_0^2 - 2k \left[\frac{x^2}{2} \right]_0^2 - \frac{k}{4} \left[\frac{x^2}{2} \right]_0^2 + k \left[x \right]_0^2$$

$$\Rightarrow -\frac{k}{4} \left[\frac{16-0}{4} \right] + k \left[\frac{8-0}{3} \right] + \frac{k}{2} \left[\frac{8}{3} \right] - 2k \times \frac{4}{2}$$

$$-\frac{k}{4} \times \frac{4}{1} + k(2)$$

$$2) -k + \frac{8k}{3} + \frac{8k}{6} - 4k - \frac{k}{2} + 2k$$

classmate

$$2) (k/2)$$

$$q_4 = \int_2^4 (x^2 - 6x + 9) \left(-\frac{k}{4}x + k \right) dx$$

DATE

$$\Rightarrow \int_2^4 \left[-\frac{k}{4}x^3 + kx^2 + \frac{6k}{4}x^2 - 6kx - \frac{9k}{4}x + 9k \right] dx$$

$$\Rightarrow -\frac{k}{4} \left[\frac{x^4}{4} \right]_2^4 + k \left[\frac{x^3}{3} \right]_2^4 + \frac{3k}{2} \left[\frac{x^3}{3} \right]_2^4 - 6k \left[\frac{x^2}{2} \right]_2^4 - \frac{9k}{4} \left[\frac{x^2}{2} \right]_2^4 + 9k \left[x \right]_2^4$$

$$\Rightarrow -\frac{k}{4} \left[\frac{256-16}{4} \right] + k \left[\frac{64-8}{3} \right] + \frac{3k}{2} \left[\frac{64-8}{3} \right]$$

$$- 6k \left[\frac{16-4}{2} \right] - \frac{9k}{4} \left[\frac{16-4}{2} \right] + 9k [2]$$

$$\Rightarrow -15k + 18.66k + 28k - 36k - 13.5k + 18k$$

$$= 0.16k$$

$$Q = q_1 + q_2 + q_3 + q_4$$

$$= 0.16k + 0.16k + k/2 + k/2$$

$$\Rightarrow k + 0.32k = \boxed{k(1.32)}$$

total Area under pdf = 1

$$\int_2^4 \left(\frac{1}{8} \right) (k) = 1$$

$$\boxed{k = 1/4}$$

$$Q = \frac{1}{4} (1.32) = 0.33$$

DATE

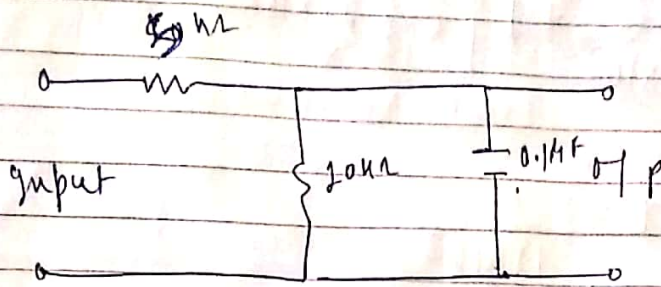


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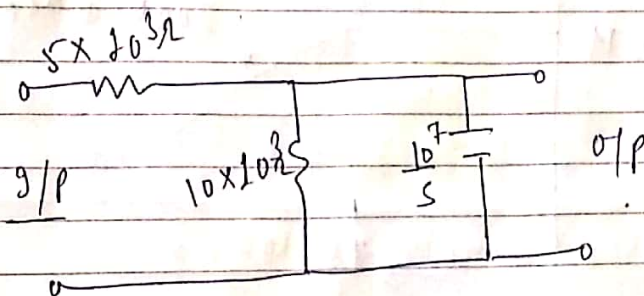
Q4.

(a)

Soln



Laplace domain



$$H(s) = \frac{o/p}{g/p} = \frac{(10^7/s) \parallel (10^4)}{(10^7/s \parallel 10^4) + 5 \times 10^3}$$

$$\frac{10^7}{s} \parallel 10^4 = \frac{10^{11}}{s + 10^4}$$

$$\frac{10^{11-4}}{10^3 + s} = \frac{10^7}{10^3 + s}$$

$$H(s) = \frac{10^7}{s + 10^3}$$

$$\frac{10^7}{s + 10^3} + \frac{5 \times 10^3 (s + 10^3)}{(s + 10^3)} = \frac{10^7 + 5 \times 10^3 s + 5 \times 10^6}{s + 10^3}$$

$$\frac{10^4}{5s + 10^4 + 5 \times 10^3} = \left[\frac{10^4}{5s + 15 \times 10^3} \right]$$

PAGE

classmate

$$H(s) =$$

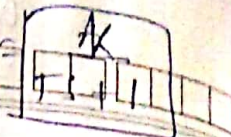
$$\frac{10^4}{15 \times 10^3}$$

$$\frac{2}{3}$$

2)

$$\frac{s}{s+1}$$

DATE



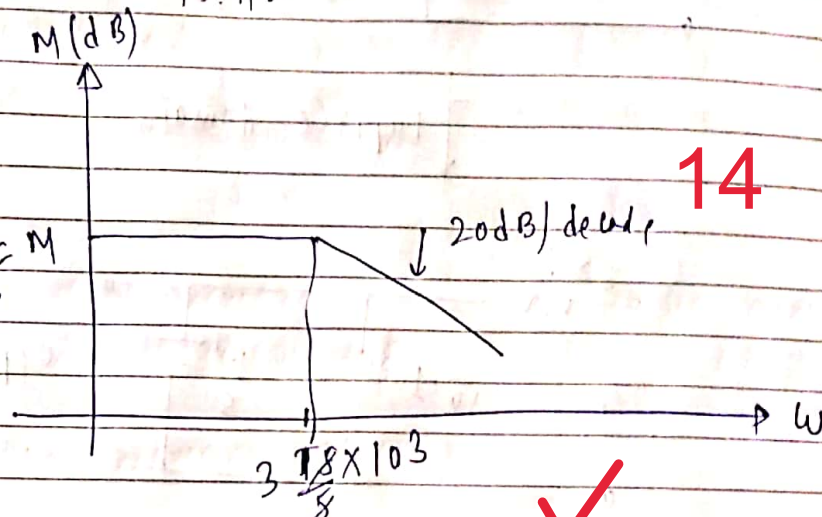
$$\left[\frac{s}{15 \times 10^3} + 1 \right]$$

$$3 \frac{15 \times 10^3}{s}$$

$$T = \frac{1}{15 \times 10^3}$$

M(dB)

-3.52 dB



$$M = 20 \log_{10} \frac{2}{3}$$

(b)

$$(1) y[n] = n y[n-1] + 4[n]$$

$$y_1[n] \rightarrow 9 y[n]$$

$$y_1[n] = n y[n-1] + 9 y[n] \quad \# \text{ not equal}$$

$$y_2[n] = 9 y[n] = 9 n y[n-1] + 9 y[n]$$

∴ Non-linear

$$y_1[n] = y[n-n_0]$$

$$y_1[n] = n y[n-1] + 4[n-n_0] \quad \textcircled{1} \quad \#$$

$$y[n-n_0] = n y[n-n_0-1] + 4[n-n_0] \quad \textcircled{2}$$

classmate

Time - variant

PAGE

(ii) $y(n) = 4[2(n+1)]$

$y_1(n) = a y(n) =$ DATE

$y_1(n) = a [4[2(n+1)]] = a y(n)$

$y_2(n) = b y(n) = b y(n)$

$y_2(n) = b 4[2(n+1)] = b y(n)$

$y_1(n) + y_2(n) = (a+b) y(n)$

$y_3(n) = (a+b) 4[2(n+1)] = a+b (y(n))$
 $= a y(n) + b y(n)$
 $= y_1(n) + y_2(n)$

1. Linear

$y_1(n) = y(n-n_0)$

$y_1(n) = 4[2(n-n_0-1)] = 4[2(n-n_0+1)]$

$y_2(n) = 4[2(n-n_0-1)]$

Time-invariant

(iii) $y(n) = \sum_{k=0}^n y(k)$

Since Σ operator is linear
therefore $y(n)$ is linear

$y_1(k) = u(k-k_0)$

$y_1(n) = \sum_{k=0}^n y[k-k_0] \quad \text{--- (1)}$

$k-k_0 = t$

$k = t + k_0$

PAGE $\sum_{-k_0}^{n-k_0} y(t)$

$$y_2(n) = \sum_{k=0}^n 4(k)$$

$y_1(n)$ & $y_2(n)$

time variant ✓

12

(C)

801st

66 block & { 16 blocks
16 blocks
16 blocks
16 16

main memory
4096 blocks

cache memory
64 blocks

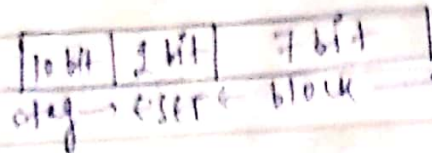
i) main memory address

$$4096 \times 128$$

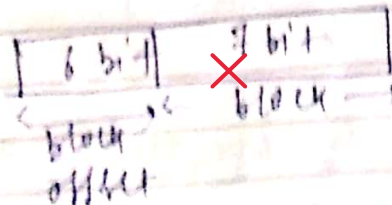
$$2^{12} \times 2^7 \Rightarrow 2^{19} = 19 \text{ bits} \quad \checkmark$$

ii)

m.m



c.m



9

iii)

$$\text{size of cache memory} = 2^{13} \times 16$$

$$16384 \text{ B}$$

$$16 \text{ KB} \quad \checkmark$$

Section-B

DATE

Q05-

(b)

Solⁿ:-

$$f = 100 \text{ MHz}, \quad v_p = 2 \times 10^7 \lambda^{2/3} \text{ m/s}$$

$$\lambda = \frac{c}{f} = \frac{3 \times 10^8}{100 \times 10^6} \Rightarrow \boxed{3 \text{ m}}$$

$$v_p = 2 \times 10^7 \times (3)^{2/3} = 4.16 \times 10^7 \text{ m/s}$$

$$\boxed{v_p \cdot v_g = c^2}$$

05

$$v_g = \frac{3 \times 3 \times 10^6}{4.16 \times 10^7} = \boxed{2.16 \times 10^9 \text{ m/s}}$$

(c)

Solⁿ:-

$$\text{Cellular system area} = 2100 \text{ km}^2$$

$$\text{area of single cell} = 6 \text{ km}^2$$

$$\text{No. of cells} \Rightarrow \frac{2100}{6} = \boxed{350}$$

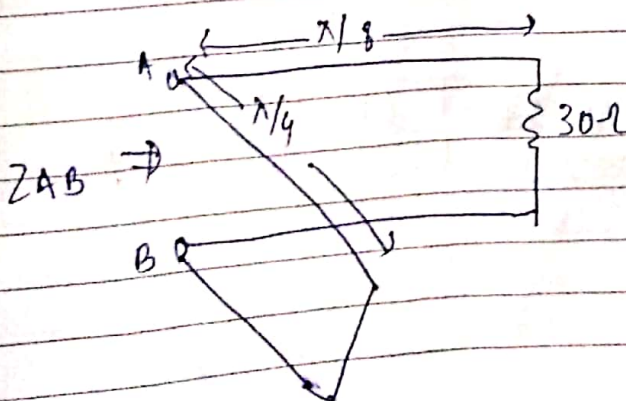
$$\text{Total No. of clusters} = \frac{350}{7} = 50$$

$$\therefore \text{system capacity} = 50 \times 1001 = \boxed{50050}$$

09

(d)

Solⁿ:-



$$\boxed{Z_0 = 60\Omega}$$

PAGE

$$Z_1 \neq Z_0 \left[\frac{Z_L \cos(\beta l) + j Z_0 \sin(\beta l)}{Z_0 \cos(\beta l) + j Z_L \sin(\beta l)} \right]$$

$$\beta l = \frac{2\pi}{\lambda} \times \frac{\lambda}{8} \Rightarrow \boxed{\pi/4}$$

$$Z_1 = 60 \left[\frac{30 \times \cancel{\sqrt{2}} + j 60 \times \cancel{\sqrt{2}}}{60 \times \cancel{\sqrt{2}} + j 30 \times \cancel{\sqrt{2}}} \right]$$

$$\Rightarrow 60 \times \frac{30}{30} \left[\frac{1 + 2j}{2 + j} \right]$$

$$Z_1 \Rightarrow 60 \angle 36.86^\circ$$

$$Z_2 = Z_0 \left[\frac{Z_L \cos(\beta l) + j Z_0 \sin(\beta l)}{Z_0 \cos(\beta l) + j Z_L \sin(\beta l)} \right]$$

$$\beta l = \frac{2\pi}{\lambda} \times \frac{\lambda}{4} = \pi/2$$

$$Z_2 = \frac{60 \left[\frac{60}{0} \right]}{0} = \boxed{\infty}$$

$$Z_{AB} \Rightarrow Z_1 \parallel Z_2 \Rightarrow 60 \angle 36.86 \parallel \infty$$

$$Z_{AB} \Rightarrow \boxed{60 \angle 36.86^\circ}$$

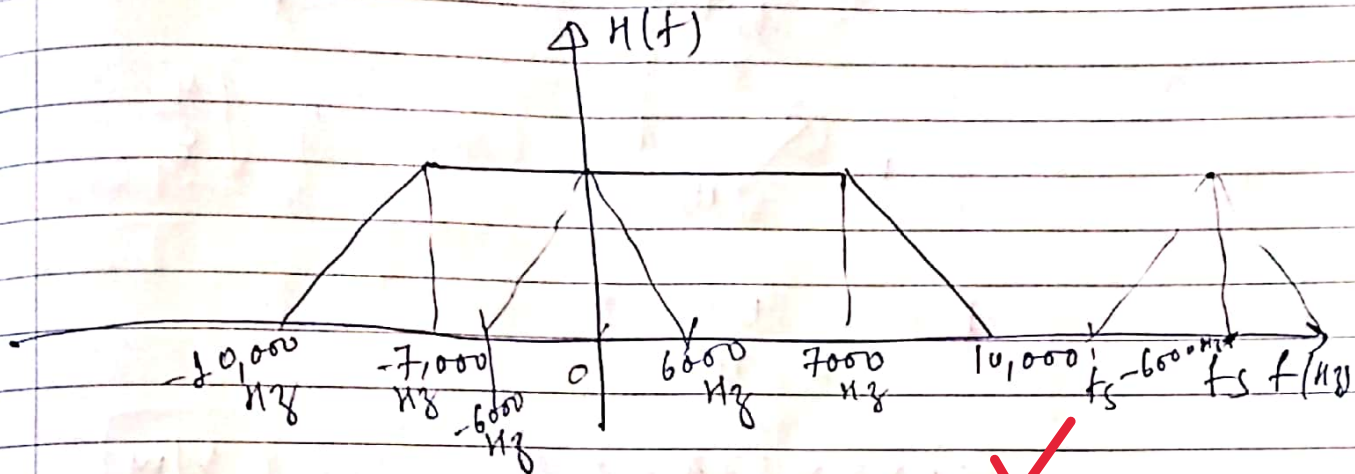
09

(f)

$$f_m = 6000 \text{ Hz}$$

DATE

$$H(f) = \begin{cases} K & |f| \leq 7000 \\ K \left[1 - \frac{|f| - 7000}{3000} \right] & 7000 < |f| < 10,000 \\ 0 & \text{otherwise} \end{cases}$$



$$f_s \geq 20,000 + G.B$$

$$f_s \geq 10,000 + 2000$$

$$f_s \geq 12,000$$

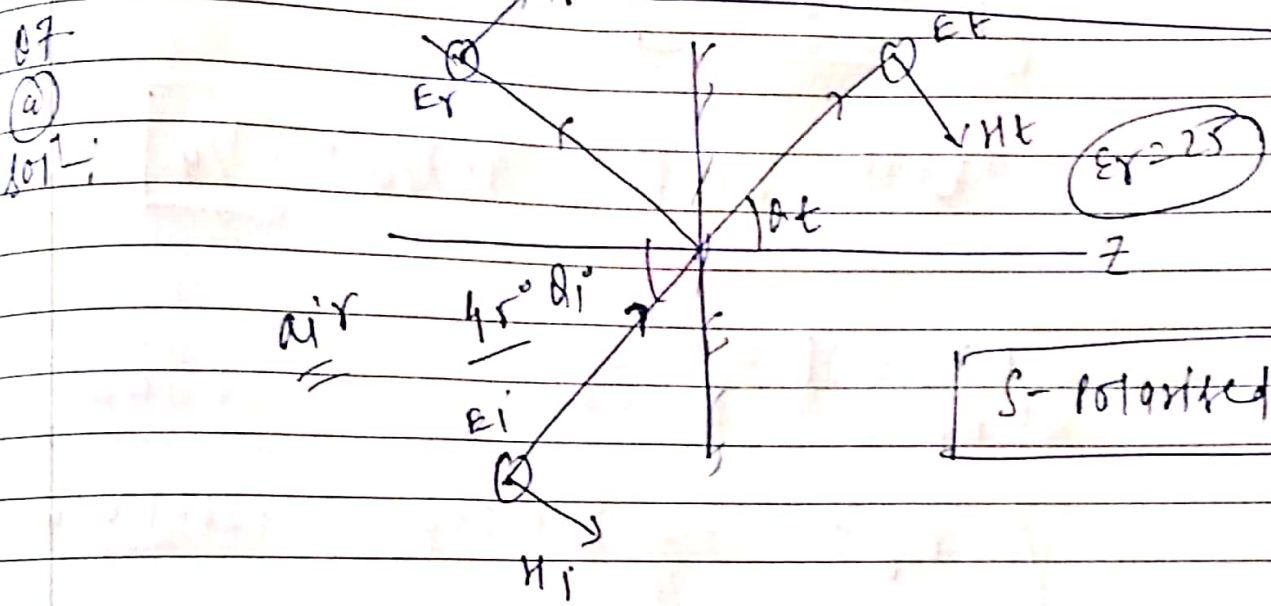
$$f_s \geq 12 \text{ kHz}$$

$$f_s - 6000 \geq 10,000 + 2000 \quad (G.B)$$

$$f_s \geq 16,000$$

$$f_s \geq 18 \text{ kHz}$$

03



$$P = \frac{1}{2} \frac{E^2}{\eta}$$

$$E = \sqrt{2 P \eta} = \sqrt{2 \times 20 \times 377}$$

$$E_i = 122.8 \text{ V/m} \quad (19)$$

By Snell's law

$$n_1 \sin \theta_i = n_2 \sin \theta_t$$

$$1 \sin 45^\circ = 5 \times \sin \theta_t$$

classmate

$$\sin \theta_t = \frac{1}{5\sqrt{2}}$$

DATE

$$\theta_t = 8.13^\circ$$

$$\Gamma = \frac{\eta_2 \sec \theta_t - \eta_1 \sec \theta_i}{\eta_2 \sec \theta_t + \eta_1 \sec \theta_i}$$

$$= \frac{1.01}{5} = 1.414$$

$$\frac{1.01}{5} + 1.414$$

$$\eta_1 = 1.212$$

$$1.616$$

$$\Gamma = -3/4$$

$$\tau = 1 + \Gamma = 1 - 3/4 = 1/4$$

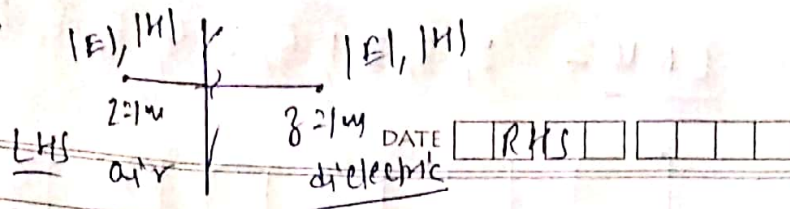
$$\frac{E_r}{E_i} = \Gamma = -3/4$$

$$E_r = \frac{-3}{4} \times 122.8 = -92.1 \text{ V/m}$$

$$\frac{E_t}{E_i} = \tau = 1/4$$

$$E_t = \frac{122.8}{4} = 30.7 \text{ V/m}$$

(ii)



$$E_{LHS} = E_1 + E_2 = (122.8 - 92.1) \hat{y}$$

$$= 30.7 \hat{y} \text{ (V/m)}$$

$$\frac{E_{LHS}}{H_{LHS}} = \eta =$$

$$\frac{30.7}{120\pi} = H_{LHS} = 0.081 \text{ (A/m)}$$

$$E_{RHS} = E_2 = 92.1 \text{ V/m}$$

$$H_{RHS} = \frac{E_{RHS}}{\eta_2} = \frac{92.1}{120\pi} = 1.22 \text{ A/m}$$

14

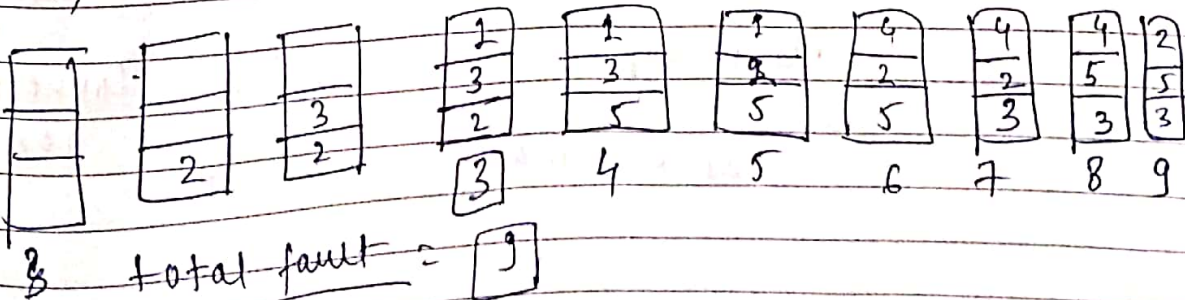
∴ attenuation on both sides is zero.

(b)

2, 3, 2, 1, 5, 2, 4, 5, 3, 2, 5, 2

Solⁿ:

i) FIFO



(ii)

LRU $\div$ 2, 3, 2, 1, 5, 2, 4, 5, 3, 4, 5, 2

DATE

			1	1	4	4	2
		3	3	5	5	5	5
	2	2	2	2	2	3	3
1	1	2	3	4	5	6	7

Total fault = 7

(iii)

optimal $\div$ (2, 3, 2, 1, 5, 2, 4, 5, 3, 2, 5, 2)

			1	5	5	5
		3	3	3	3	3
	2	2	2	2	4	2
1	2	3	4	5	6	

Total fault = 6

14

(c)

80%

LED Tx = 850 nm,

 $P_{in} = -10 \text{ dBm}$ $q = 50 \mu\text{m}$, $\alpha = 2.5 \text{ dB/km}$, $(\Delta T)_{\text{max}} = 3 \text{ m/km}$ $(\Delta T)_{\text{model}} = 1 \text{ ns/km}$, $(P_r) = -42 \text{ dBm}$ BER = 10^{-9} $(t_r)_s = 11 \text{ ns}$, $(t_r)_R = 12 \text{ ns}$ splice loss = 0.5 dB/splice , connector loss = 1 dB

link margin = 6 dB

(i)

$$(P_{in} - P_R) = 2L + (L-1)0.5 + 2 \text{ dB} + 6 \text{ dB}$$

DATE

$$-10 - (-42) = 2.5L + (L-1)0.5 + 8$$

$$32 - 8 = 3L - 0.5$$

$$\frac{24.5}{3} = L = 8.16 \text{ km}$$

(ii)

5011 ÷

$$B \& R = \frac{P_R}{BW}$$

09

$$10^{-9} =$$