



Saket Centre

**MADE EASY**

India's Best Institute for IES, GATE & PSUs

## ESE 2024 : Mains Test Series

UPSC ENGINEERING SERVICES EXAMINATION

### Electrical Engineering

Test-1 : Electrical Circuits [All Topics]

Control Systems [All Topics]

Name :

Roll No

#### Test Centres

Delhi  Bhopal  Jaipur   
Pune  Kolkata  Hyderabad

#### Student's Signature

#### Instructions for Candidates

1. Do furnish the appropriate details in the answer sheet (viz. Name & Roll No).
2. There are Eight questions divided in TWO sections.
3. Candidate has to attempt FIVE questions in all in English only.
4. Question no. 1 and 5 are compulsory and out of the remaining THREE are to be attempted choosing at least ONE question from each section.
5. Use only black/blue pen.
6. The space limit for every part of the question is specified in this Question Cum Answer Booklet. Candidate should write the answer in the space provided.
7. Any page or portion of the page left blank in the Question Cum Answer Booklet must be clearly struck off.
8. There are few rough work sheets at the end of this booklet. Strike off these pages after completion of the examination.

#### FOR OFFICE USE

Question No.	Marks Obtained
Section-A	
Q.1	41
Q.2	56
Q.3	
Q.4	45
Section-B	
Q.5	53
Q.6	
Q.7	51
Q.8	
<b>Total Marks Obtained</b>	<b>246</b>

Signature of Evaluator

Cross Checked by

Nihal

- Approach and accuracy is very good.  
→ Do not forget to mention units  
→ Do not write outside borders marked.

## IMPORTANT INSTRUCTIONS

**CANDIDATES SHOULD READ THE UNDERMENTIONED INSTRUCTIONS CAREFULLY. VIOLATION OF ANY OF THE INSTRUCTIONS MAY LEAD TO PENALTY.**

### DONT'S

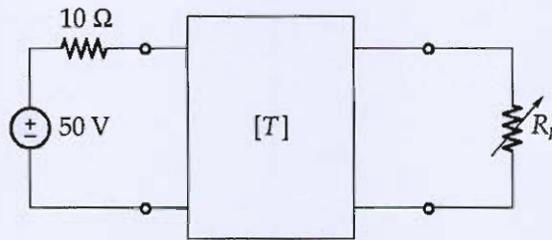
1. Do not **write your name or registration** number anywhere inside this **Question-cum-Answer Booklet (QCAB)**.
2. Do not write anything other than the actual answers to the questions anywhere inside your QCAB.
3. Do not tear off any leaves from your QCAB, if you find any page missing do not fail to notify the supervisor/**invigilator**.
4. Do not leave behind your QCAB on your table unattended, it should be handed over to the invigilator after conclusion of the exam.

### DO'S

1. Read the Instructions on the cover page and strictly follow them.
2. Write your registration number and other particulars, in the space provided on the cover of QCAB.
3. Write legibly and neatly.
4. For rough notes or calculation, the last two blank pages of this booklet should be used. The rough notes should be crossed through afterwards.
5. If you wish to cancel any work, draw your pen **through** it or write "Cancelled" across it, otherwise it may be evaluated.
6. Handover your QCAB **personally** to the invigilator before **leaving the examination** hall.

## Section A : Electrical Circuits

- 2.1 (a) The ABCD parameter of the two-port network in figure are  $\begin{bmatrix} 4 & 20\Omega \\ 0.1S & 2 \end{bmatrix}$ .



The output port is connected to a variable load for maximum power transfer. Find  $R_L$  and the maximum power transferred.

[12 marks]

ABCD parameter of the network =  $\begin{bmatrix} 4 & 20 \\ 0.1 & 2 \end{bmatrix}$

$$V_1 = 4V_2 + 20I_2$$

$$I_1 = 0.1V_2 + 2I_2$$

from figure

$$V_1 = 50 - 10I_1$$

for finding equivalent resistance

~~$I_2 = 0$~~

50V voltage source deactivated

$$V_1 = -10I_1$$

$$-10I_1 = 4V_2 + 20I_2$$

$$-10I_1 = -V_2 + 20I_2$$

$$\begin{array}{r} + \\ + \end{array}$$

$$0 = 5V_2 + 40I_2$$

$$\frac{+40I_2}{8} = \frac{-5V_2}{8}$$

$$V_2 = +8I_2$$

$$\frac{V_2}{I_2} = 8\Omega$$

$$R_{th} = 8\Omega$$

for finding  $V_{oc}$

$$I_2 = 0$$

$$V_1 = 50 - 10I_1$$

$$V_1 = 4V_2 - 20(I_2)$$

$$I_2 = 0 \quad V_1 = 4V_2 \quad \text{--- (i)}$$

$$I_1 = 0.1V_2 - 2I_2 \quad \text{--- (ii)}$$

$$I_1 = 0.1V_2 \quad \text{--- (ii)}$$

by solving equation (i) & (ii)

$$I_1 = 0.1 \left( \frac{V_1}{4} \right) = \frac{0.1}{4} (50 - 10I_1)$$

$$4I_1 = 5 - I_1$$

$$5I_1 = 5$$

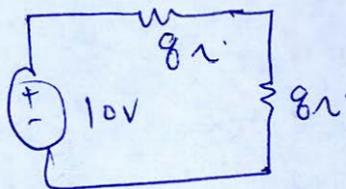
$$\boxed{I_1 = 1}$$

from equation (ii)

$$V_2 = 10 \text{ volt} = V_{th} \quad \checkmark$$

$$R_L = R_{th} = 8\Omega \quad \checkmark$$

(by max MPT)



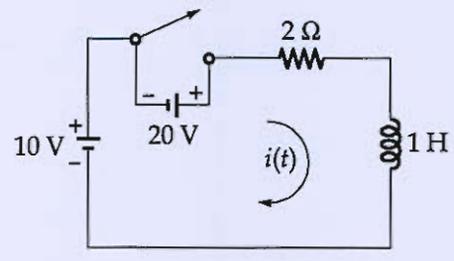
12  
"Good"

$$P_{max} = \left( \frac{V_{th}^2}{4R_{th}} \right) \quad \checkmark$$

$$= \left( \frac{100}{4 \times 8} \right)$$

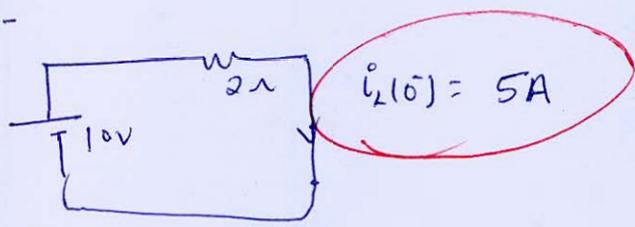
$$= \frac{25}{8} \text{ watt} = 3.125 \text{ watt} \quad \checkmark$$

2.1 (b) Determine the current  $i(t)$  in the circuit shown in figure at an instant  $t$ , after opening the switch at  $t = 0$ , if a current of 1 A had been passed through the circuit at the instant of opening.

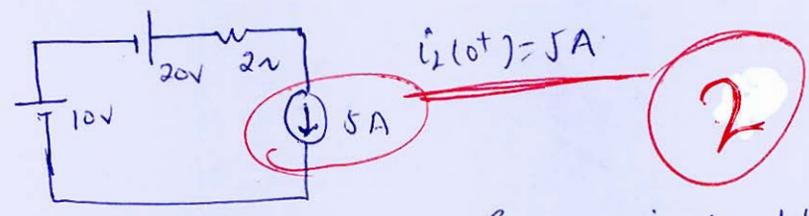


[12 marks]

before  $t=0$   
at  $t=0^-$

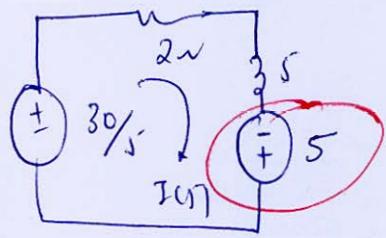


after  $t=0^+$



current in inductor does not change instantly

convert ckt. time domain to sdomain



$$I(s) = \left( \frac{30/s + 5}{s+2} \right)$$

$$I(s) = \frac{5(s+6)}{s(s+2)}$$

$$I(s) = \frac{A}{s} + \frac{B}{(s+2)}$$

$$A = 15, B = -10$$

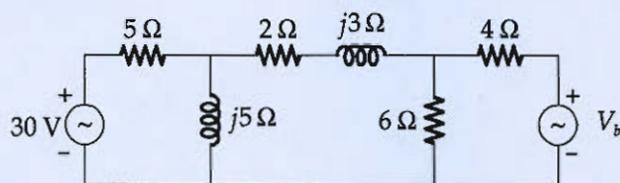
$$I(s) = 15/s - 10/s+2$$

apply Laplace inverse transform

$$I(t) = (15 - 10e^{-2t})u(t)$$

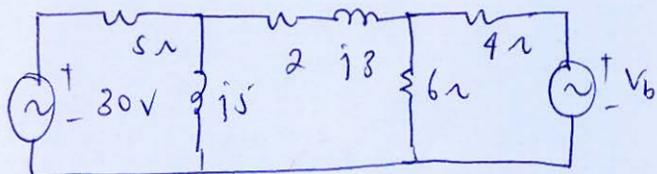
3x+2

Q.1 (c) For the circuit shown below:



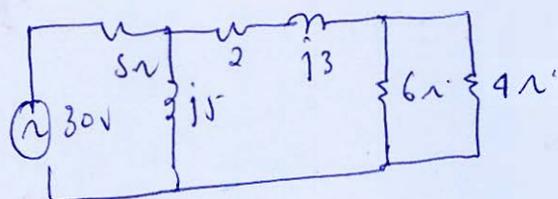
Determine the voltage  $V_b$  which results in a zero current through the  $(2 + j3)\Omega$  impedance branch. Using superposition theorem.

[12 marks]

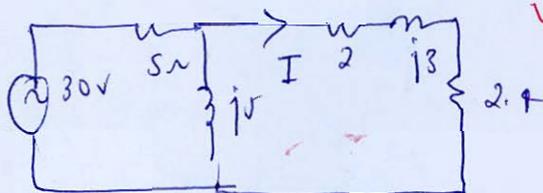


by using superposition theorem

① Consider 30V voltage source and deactivate  $V_b$  voltage source and replace by internal resistance.



6 & 4 ohm are in parallel  
 $= \frac{6 \times 4}{10} = 2.4$



$$Z_{eq} = 5 + \frac{j5 \times (4 + j3)}{j5 + 4 + j3}$$

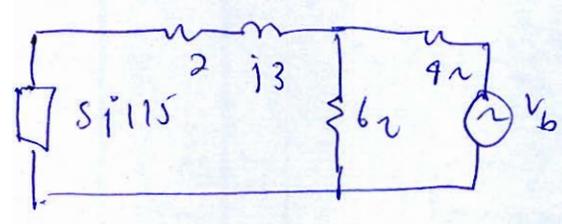
$$I_s = \left( \frac{30}{Z_{eq}} \right) = 4.389 \angle -22.369^\circ$$

$$I = \left( \frac{I_s \times j5}{j5 + 4 + j3} \right)$$

$$= \frac{4.389 \angle -22.369^\circ \times (5j)}{(4.4 + j8)}$$

$$I_1 = 2.404 \angle 6.44^\circ$$

by considering  $V_b$  voltage source, deactivate 30V voltage source by replacing its internal resistance.



$$Z_{eq} = 6 \parallel (4.5 + j5.5) + 4$$

$$= 7.999 \angle 10.912^\circ$$

$$I_{S2} = \left( \frac{V_b}{Z_{eq}} \right) = 0.134 V_b \angle -10.912^\circ$$

$$I_1' = \left( \frac{I_{S2} \times 4}{6 + 4.5 + j5.5} \right)$$

$$I_1' = 0.068 V_b \angle -38.558^\circ$$

to make zero current through  $(2 + j3)$   
 then  $I_1 = I_1'$  —  $I_1 = -I_1'$

~~$$7.999 \angle 10.912^\circ = 0.068 V_b \angle -38.558^\circ$$~~

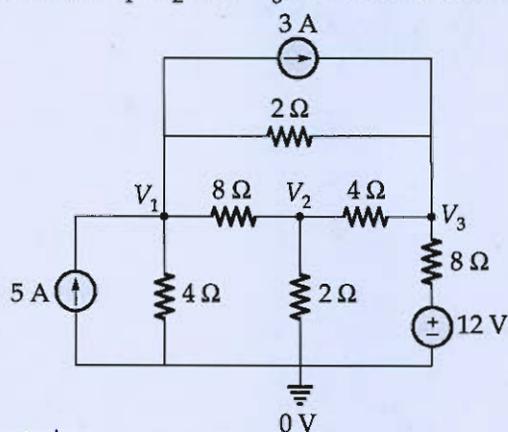
$$V_b = 71.14 + j83.208$$

$$V_b = 109.474 \angle 49.47^\circ$$

not write here

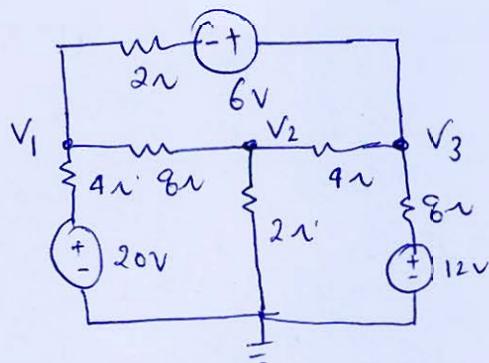
5

Q.1 (d) Use nodal analysis to find  $V_1$ ,  $V_2$  and  $V_3$  in the circuit of figure.



simply fig ckt.

[12 marks]



by using nodal analysis  
apply nodal at 1 node

$$\frac{V_1 - 20}{4} + \frac{V_1 - V_2}{8} + \frac{V_1 + 6 - V_3}{2} = 0$$

$$2V_1 - 40 + V_1 - V_2 + 4V_1 + 24 - 4V_3 = 0$$

$$7V_1 - V_2 - 4V_3 = 16 \quad \text{--- (1)}$$

apply nodal at 2 node:

$$\frac{V_2}{2} + \frac{V_2 - V_3}{4} + \frac{V_2 - V_1}{8} = 0$$

$$4V_2 + 2V_2 - 2V_3 + V_2 - V_1 = 0$$

$$-V_1 + 7V_2 - 2V_3 = 0 \quad \text{--- (2)}$$

apply nodal at node 3.

$$\frac{V_3 - 12}{8} + \frac{V_3 - V_2}{4} + \frac{V_3 - 6 - V_1}{2} = 0$$

$$V_3 - 12 + 2V_3 - 2V_2 + 4V_3 - 24 - 4V_1 = 0$$

$$-4V_1 + 2V_2 + 7V_3 = 36 \quad \text{--- (11)}$$

by solving equation (i) (ii) & (iii)

$$\begin{bmatrix} 7 & -1 & -4 \\ -1 & 7 & -2 \\ -4 & -2 & 7 \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \\ V_3 \end{bmatrix} = \begin{bmatrix} 16 \\ 0 \\ 36 \end{bmatrix}$$

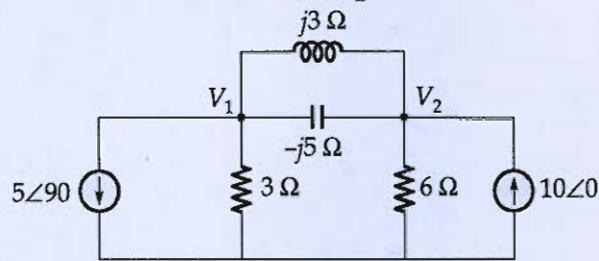
(11)

$$V_1 = 10 \text{ Volt}$$

$$V_2 = 4.933 \text{ Volt}$$

$$V_3 = 12.267 \text{ Volt}$$

Q.1 (e) Use nodal analysis on the circuit to find  $V_2$ .



[12 marks]

apply nodal at  $V_1$

$$5\angle 90^\circ + \frac{V_1}{3} + \frac{V_1 - V_2}{-j5} + \frac{V_1 - V_2}{j3} = 0$$

$$V_1 \left( \frac{1}{3} - \frac{1}{j5} + \frac{1}{j3} \right) + V_2 \left( \frac{1}{j5} - \frac{1}{j3} \right) = -5\angle 90^\circ$$

$$V_1 (0.333 - 0.133j) + V_2 (0.133j) = -5j \quad \text{--- (I)}$$

apply nodal at  $V_2$

$$\frac{V_2}{6} + \frac{V_2 - V_1}{-j5} + \frac{V_2 - V_1}{j3} = 10\angle 0^\circ$$

$$V_2 \left( \frac{1}{6} - \frac{1}{j5} + \frac{1}{j3} \right) + V_1 \left( \frac{1}{j5} - \frac{1}{j3} \right) = 10 \quad \text{--- (II)}$$

$$V_2 (0.166 - 0.133j) + V_1 (-0.133j) = 10 \quad \text{--- (II)}$$

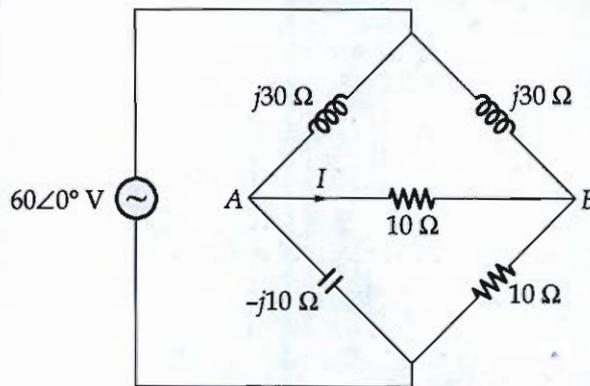
$$\begin{bmatrix} 0.333 - 0.133j & 0.133j \\ 0.133j & 0.166 - 0.133j \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} -5\angle 90^\circ \\ 10 \end{bmatrix}$$

$$V_2 = \frac{\Delta_2}{\Delta} = \frac{2.665 - 1.33j}{0.0556 - 0.0665} = 34.35\angle 23.57^\circ \quad \text{vald}$$

$$\Delta_2 = \begin{bmatrix} 0.333 - 0.133j & -5j \\ 0.133j & 10 \end{bmatrix} = 2.665 - 1.33j$$

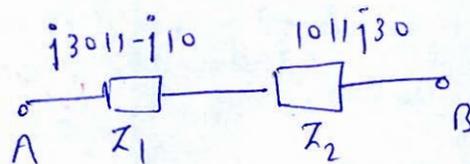
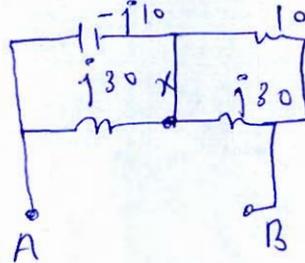
$$\Delta = \begin{bmatrix} 0.333 - 0.133j & 0.133j \\ 0.133j & 0.166 - 0.133j \end{bmatrix} = 0.0556 - 0.0665j$$

Q.2 (a) Determine the current  $I$  through the terminal AB of the network shown below:



[20 marks]

find equivalent resistance across AB  
deactivate 60 volt voltage source and replace  
by its internal resistance.



$$Z_1 = \frac{j30 \times -j10}{j30 - j10}$$

$$= \frac{j30 \times -j10}{j20}$$

$$= -j15$$

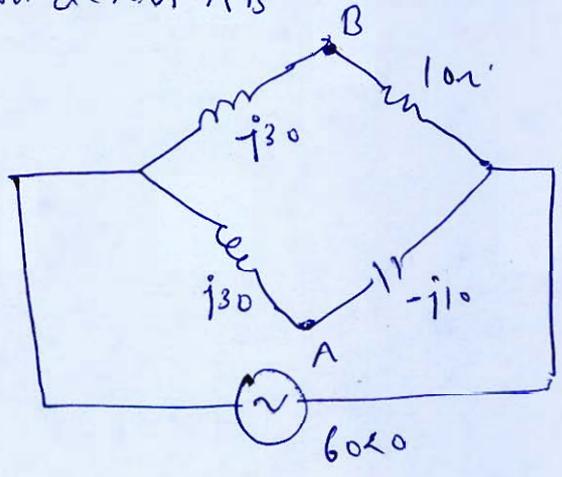
$$Z_2 = \frac{10 \times j30}{10 + j30}$$

$$= \left( \frac{j30}{j3+1} \right)$$

$$= 9 + j3 \Omega$$

$$Z_{eq} = -j15 + 9 + j3 \Omega = \underline{9 - j12 \Omega}$$

find  $V_{th}$  across AB

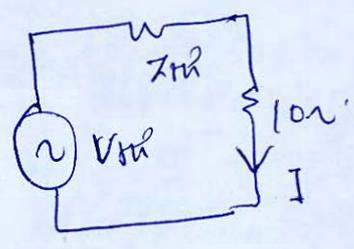


$$V_A = \frac{60 \times -j10}{j30 - j10} = \frac{60 \times \cancel{-j10}}{\cancel{j20}} = -30 \angle 180^\circ$$

$$V_B = \frac{60 \times 1 \angle 0}{10 + j30} = \left( \frac{60}{j3+1} \right) = 6 \angle -18^\circ$$

$$\begin{aligned} V_{AB} = V_{th} &= V_A - V_B \\ &= -30 \angle (6 - 18^\circ) \\ &= -36 + 18i = 40.25 \angle 153.734^\circ \end{aligned}$$

40.25 153.734° Volt



$$I = \left( \frac{V_{th}}{Z_{th} + 10} \right)$$

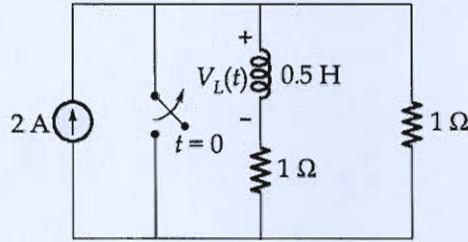
$$= \frac{40.25 \angle 153.734^\circ}{(9 - 12i + 10)}$$

$$I = 1.79 \angle -174.28^\circ \text{ (A)}$$

18

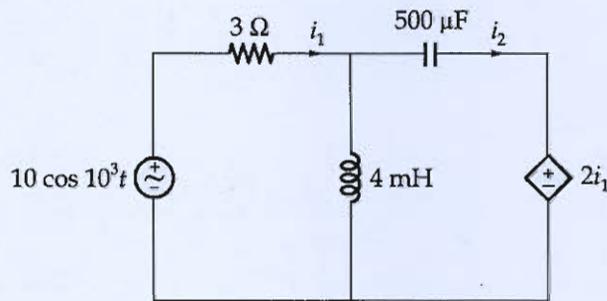


- Q.2 (b) (i) For the network shown in figure below, the switch is closed for a long time and at  $t = 0$ , the switch is opened.



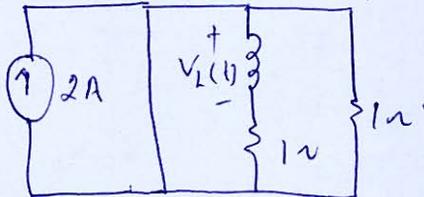
Determine the voltage across inductor for  $t > 0$ .

- (ii) Obtain expressions for the time domain currents  $i_1$  and  $i_2$  in the circuit given as figure.



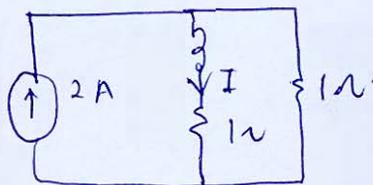
[10 + 10 marks]

at  $t = 0^-$  (before opening)

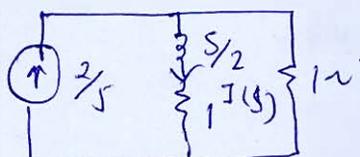


$i_L(0^-) = 0A$   
switch closed.

after  $t = 0^+$   
switch open



convert ckt into s domain



$$I(s) = \frac{2/s \times 1}{1 + (1 + s/2)}$$

2b(i)  
do not write here



$$I(s) = \frac{2}{s(4+s)} = \frac{4}{s(s+4)}$$

$$V_2(s) = I(s) \times \frac{s}{2}$$

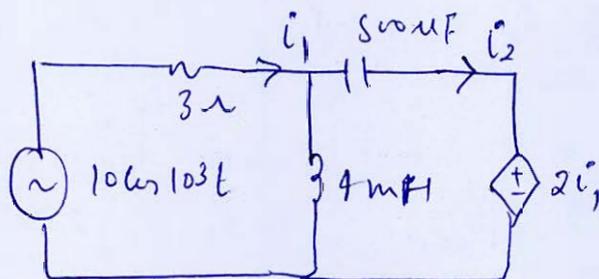
$$= \frac{4}{s(s+4)} \times \frac{s}{2}$$

$$V_2(s) = \left( \frac{2}{s+4} \right)$$

9 apply Laplace inverse transform

$$V_2(t) = 2e^{-4t}u(t)$$

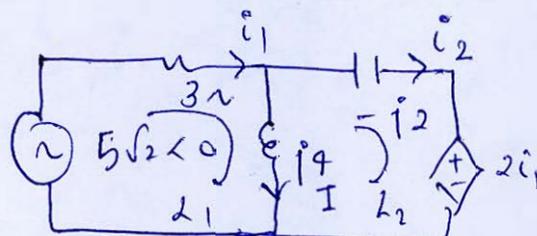
2b(ii)



$$(\omega = 1000 \text{ rad/sec})$$

$$jX_L = j(\omega L) = 1000 \times 4 \times 10^{-3} j = 4j$$

$$X_C = \left( \frac{1}{\omega C} \right) = \left( \frac{1 \times 2}{2 \times 10^3 \times 10^{-6}} \right) = 2$$



$$i_1 = I + i_2$$

$$I = i_1 - i_2$$

apply KVL at loop 1.

$$5\sqrt{2} - 3i_1 - 4j(I_1 - I_2) = 0$$

$$i_1(3 + 4j) - 4jI_2 = 5\sqrt{2} \quad \text{--- (1)}$$

apply KCL at loop 2.

$$-(-i2)i_2 - 2i_1 - j4(I_2 - i_1) = 0$$

$$j2i_2 - j4i_2 - 2i_1 + j4i_1 = 0$$

$$j2i_2 = (-2 + j4)i_1$$

$$i_2 = i_1 \left( \frac{-1 + j2}{1} \right) = i_1 (2 + j) \quad \text{--- (11)}$$

put value of  $i_2$  to (1) eqn.

$$i_1(3 + j) - 4i_1(2 + j)i_1 = 5\sqrt{2}$$

$$i_1 = \left( \frac{5\sqrt{2}}{7 - 4j} \right)$$

$$i_{1 \text{ rms}} = 0.877 \angle 29.744^\circ$$

$$i_1 = 1.24 \cos(1000t + 29.744^\circ) \text{ A}$$

$$i_2 = (2 + j)i_1$$

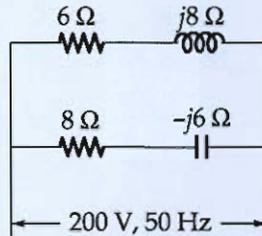
$$= 1.96 \angle 56.301^\circ$$

$$i_2 = 2.77 \cos(1000t + 56.301^\circ) \text{ A}$$



Q.2 (c) For the circuit shown below, calculate,

- (i) Total admittance, total conductance and total susceptance.
- (ii) Total current and total power factor (pf).
- (iii) The value of pure capacitance to be connected in parallel with the above combination to make the total power factor (pf) unity.



[20 marks]

①

$$\text{admittance } Y = \left( \frac{1}{6 + j8} + \frac{1}{8 - j6} \right)$$

$$Y = \left( \frac{6 - j8}{6^2 + 8^2} + \frac{8 + j6}{8^2 + 6^2} \right)$$

$$Y = \frac{1}{100} (6 - j8 + 8 + j6)$$

$$Y = \left( \frac{14 - j2}{100} \right) = 0.1414 \angle -8.13^\circ$$

$$\boxed{\text{total admittance} = 0.1414 \angle -8.13^\circ}$$

$$\boxed{\text{total conductance} = \left( \frac{14}{100} \right)}$$

$$= 0.14 \angle 0^\circ$$

$$\boxed{\text{total susceptance} = \frac{2}{100} = 0.02 \angle -90^\circ}$$

②

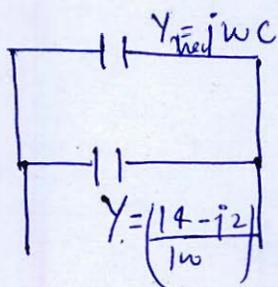
$$\text{total current} = (YV)$$

$$= 0.1414 \times 200$$

$$\boxed{I = 28.28 \angle -8.13^\circ}$$

$$\text{power factor} = \cos(8.13)$$

$$\boxed{\text{Pf} = 0.989 \text{ lagg.}}$$



$$Y_{\text{net}} = Y + Y_{\text{new}}$$

$$Y_{\text{net}} = \frac{14}{100} - \frac{j2}{100} + jwC$$

to make power factor unity  
 $\text{Im}(Y_{\text{net}})$  must be zero.

$$Y_{\text{net}}(\text{imag}) = 0$$

$$\left(-\frac{2}{100} + wC\right) = 0$$

$$wC = \frac{2}{100}$$

$$C_{\text{new}} = \frac{2}{100w}$$

$$C_{\text{new}} = \frac{2}{100 \times 2\pi \times 50}$$

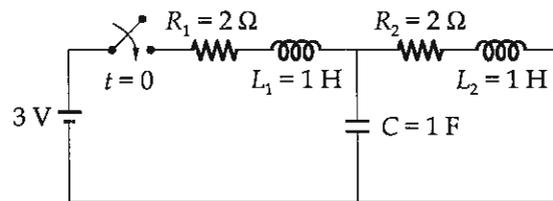
$$C_{\text{new}} = \frac{2 \times 10^6}{100 \times 100\pi \times 10^6}$$

$$C_{\text{new}} = \frac{200}{\pi} \mu\text{F}$$

$$\boxed{C_{\text{new}} = 63.662 \mu\text{F}}$$

19

- Q.3 (a) In the network shown in figure the switch is closed at time  $t = 0$ . Assuming all the initial currents and voltages as zero, find the current through the inductor  $L_2$  by the use of Norton's theorem.



[20 marks]



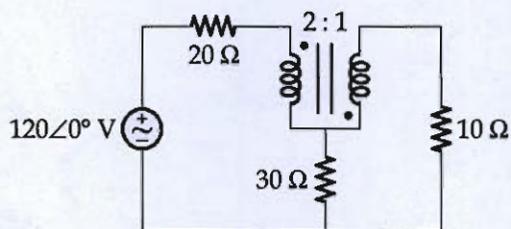


- Q.3 (b) Show that the resonant frequency  $\omega_0$  of a series  $R$ - $L$ - $C$  circuit is geometric mean of  $\omega_1$  and  $\omega_2$ , i.e., the upper and lower half power frequencies respectively.

[20 marks]



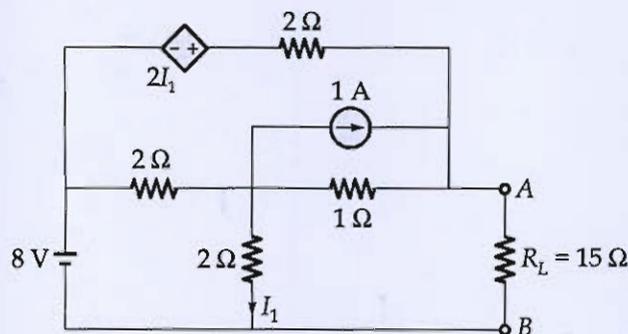
- 2.3 (c) Calculate the power supplied to the  $10\ \Omega$  resistor in the ideal transformer circuit given in the figure below.



[20 marks]



2.4 (a) Determine the current through the load resistance  $R_L = 15 \Omega$  across the terminal A-B of the circuit shown in figure below, using Thevenin's theorem. Also find the maximum power that can be transferred to the load resistance  $R_L$ .

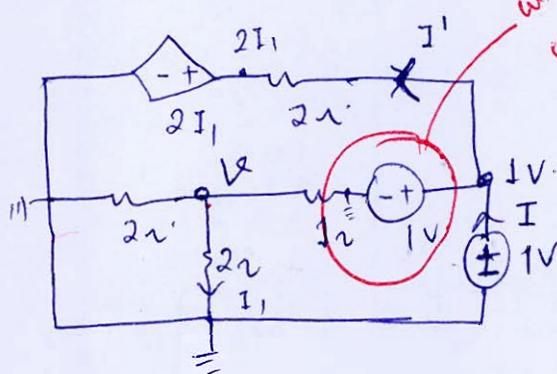


[20 marks]

apply Thevenin theorem

find  $R_{th}$ .

deactivate the all the sources and replace by internal resistances.



*will be SC while calculating Rin.*

$$\frac{1 - 2I_1}{2} = I'$$

$$\frac{V}{2} + \frac{V}{2} + \frac{V}{2} = 0$$

$$V = 0$$

$$I_1 = \frac{V}{2} = 0$$

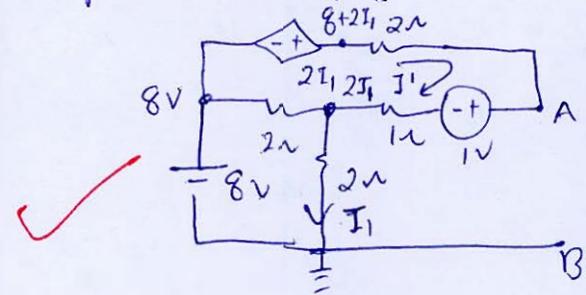
$$I = I' = \frac{1 - 2 \times 0}{2}$$

$$I = \frac{1}{2}$$

$$R_{th} = \left( \frac{1}{1} \right) = \left( \frac{1}{1/2} \right)$$

$$R_{th} = 2 \Omega$$

find  $V_{th}$ . open across A+B



$$8 + 2I_1 - 2I_1 - 1 - I_1 = 2I_1$$

$$7 = 3I_1$$

$$I_1 = \frac{7}{3}$$

$$V_A - 1 - \frac{7}{3} = 2I_1 = V_B$$

$$I' = \left( \frac{2I_1 - 8}{2} \right) + I_1 = \frac{4I_1 - 8}{2} = \frac{7}{3}$$

$$12I_1 - 24 = 14$$

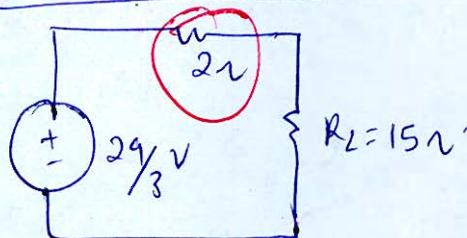
$$6I_1 - 12 = 7$$

$$I_1 = \left( \frac{19}{6} \right)$$

$$V_{AB} = 1 + \frac{7}{3} + 2 \left( \frac{19}{6} \right)$$

$$= 1 + \frac{26}{3} = \frac{29}{3}$$

$$V_{th} = \frac{29}{3} \text{ Volt}$$



$$I = \left( \frac{\frac{29}{3}}{2 + 15} \right) = \left( \frac{29}{3 \times 17} \right) = \left( \frac{29}{51} \right)$$

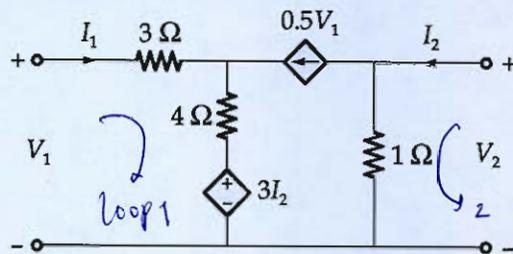
$$P_{\text{max transfer}} = \left( \frac{29}{51} \right)^2 \times 15$$

$$P_{\text{max transfer}} = 4.85 \text{ watt}$$

6



Q.4 (b) Find the  $h$ -parameters for the two-port network shown



[20 marks]

hparameter

$$V_1 = h_{11}I_1 + h_{12}V_2$$

$$I_2 = h_{21}I_1 + h_{22}V_2$$

apply KCL loop 1.

$$V_1 - 3I_1 - 4(I_1 + 0.5V_1) - 3I_2 = 0$$

$$V_1 - 3I_1 - 4I_1 - 2V_1 - 3I_2 = 0$$

$$-V_1 + 3I_2 + 7I_1 = 0 \quad \text{--- (I)}$$

apply KCL loop 2.

$$V_2 - 1(I_2 - 0.5V_1) = 0$$

$$V_2 - I_2 + 0.5V_1 = 0 \quad \text{--- (II)}$$

from equation (I)

$$V_1 = -7I_1 - 3I_2$$

put value of  $V_1$  in eq<sup>n</sup> (II)

$$V_2 - I_2 + 0.5(-7I_1 - 3I_2) = 0$$

$$V_2 - I_2 - 3.5I_1 - 1.5I_2 = 0$$

$$2.5I_2 = V_2 - 3.5I_1$$

$$I_2 = \frac{-3.5}{2.5} I_1 + \frac{V_2}{2.5}$$

put value of  $I_2$  in equation (1)

$$V_2 + \frac{3.5}{2.5} I_1 - \frac{V_2}{2.5} + \frac{V_1}{2} = 0$$

$$\frac{V_1}{2} = -0.6V_2 - 1.4I_1$$

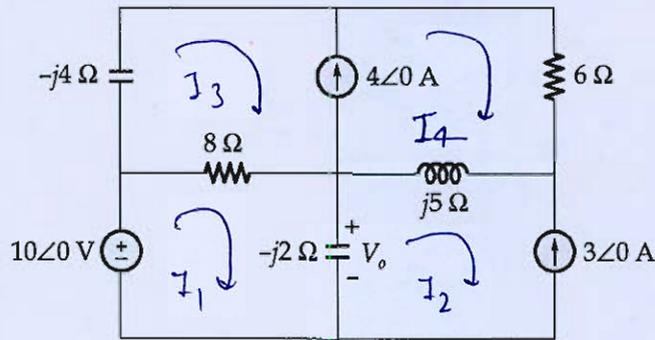
$$V_1 = -2.8I_1 - 1.2V_2$$

$$I_2 = \underline{-1.4I_1} + \underline{0.4V_2}$$

$$h \text{ parameter} = \begin{bmatrix} -2.8 & -1.2 \\ -1.4 & 0.4 \end{bmatrix}$$

19

Q.4 (c) Solve for  $V_o$  in the circuit of figure using mesh analysis.



apply mesh at loop 1.

[20 marks]

$$10 - 8(I_1 - I_3) + j2(I_1 - I_2) = 0$$

$$10 - 8I_1 + 8I_3 + j2I_1 - j2I_2 = 0 \quad \text{--- (i)}$$

$$\begin{cases} I_2 = -3 \text{ A} \quad \text{--- (ii)} \\ I_4 - I_3 = 4 \text{ A} \quad \text{--- (iii)} \end{cases}$$

by applying current in same branch are same.

by apply mesh in (i) (iii) & (iv) at same time

$$10 - (-j4)I_3 - 6I_4 - j5(I_4 - I_2) + j2(I_1 - I_2) = 0$$

$$(10 + j6) = (8 - j2)I_1 - 8I_3 \quad \text{--- (iv)}$$

$$10 + j4I_3 - 6(4 + I_3) - j5(4 + I_3 + 3) + j2(I_1 + 3) = 0$$

$$10 + j4I_3 - 24 - 6I_3 - 35j - j5I_3 + j2I_1 + 6j = 0$$

$$14 + 29j = j2I_1 + I_3(-6 - j) \quad \text{--- (v)}$$

$$\begin{bmatrix} 8 - j2 & -8 \\ j2 & -6 - j \end{bmatrix} \begin{bmatrix} I_1 \\ I_3 \end{bmatrix} = \begin{bmatrix} 10 + j6 \\ 14 + 29j \end{bmatrix}$$

$$\Delta = \begin{bmatrix} 8-j2 & -8 \\ j2 & -6-j \end{bmatrix}$$

$$= -50 + 20i \quad \checkmark$$

$$\Delta_1 = \begin{bmatrix} 10+j6 & -8 \\ 14+29i & -6-j \end{bmatrix}$$

$$= 58 + 186i \quad \checkmark$$

$$I_1 = \frac{\Delta_1}{\Delta} = \left( \frac{58 + 186i}{-50 + 20i} \right)$$

$$I_1 = 3.618 \angle -85.517^\circ \quad \checkmark$$

$$V_0 = (I_1 - I_2) \times -j2$$

$$= (3.618 \angle -85.517^\circ + 3) \times -j2$$

$$V_0 = -7.214 - 6.565i \quad \checkmark$$

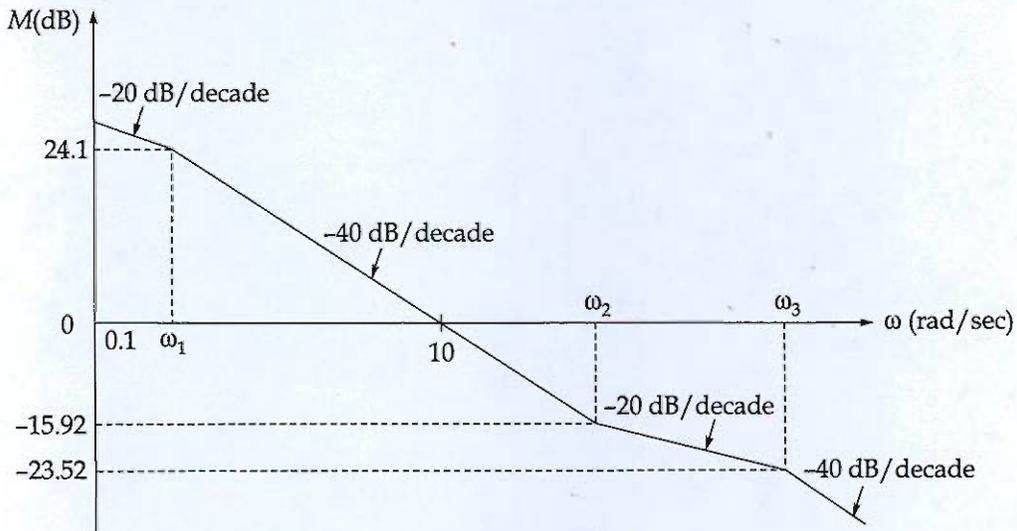
$$V_0 = 9.754 \angle -137.69^\circ \quad \text{Volt} \quad \checkmark$$

20

"Good Approach"

## Section B : Control System

Q.5 (a) Obtain the open loop transfer function for a unity negative feedback system whose bode magnitude plot is shown below:



[12 marks]

- (i) bode plot start  $-20$  dB/decade slope means it has one pole at origin
- (ii) at  $\omega = \omega_1$  slope changes  $-20$  dB/decade to  $-40$  dB/decade means it has one pole at  $\omega = \omega_1$  ✓
- (iii) at  $\omega = \omega_2$  slope changes  $-40$  dB/decade to  $-20$  dB/decade means it has one zero at  $\omega = \omega_2$  ✓
- (iv) at  $\omega = \omega_3$  slope changes  $-20$  dB/dec to  $-40$  dB/dec means it has one pole at  $\omega = \omega_3$  ✓

$$\text{OLTF} = \frac{K \left( \frac{s}{\omega_2} + 1 \right)}{s \left( \frac{s}{\omega_1} + 1 \right) \left( \frac{s}{\omega_3} + 1 \right)}$$

$$-40 = \left( \frac{-15.92 - 0}{\log \omega_2 - \log 10} \right)$$

$$\log \left( \frac{\omega_2}{10} \right) = \left( \frac{15.92}{40} \right)$$

$$\omega_2 = 10 \cdot 10^{\left( \frac{15.92}{40} \right)} = 25 \text{ rad/sec.}$$

$$-20 = \frac{-23.52 + 15.92}{\log \left( \frac{\omega_3}{\omega_2} \right)}$$

$$\omega_3 = \omega_2 \cdot 10^{(0.38)}$$

$$\omega_3 = 25 \times 10^{0.38} = 60 \text{ rad/sec.}$$

$$-40 = \frac{0 - 24.1}{\log \left( \frac{10}{\omega_1} \right)}$$

$$\left( \frac{10}{\omega_1} \right) = 10^{\frac{24.1}{40}} = 4$$

$$\omega_1 = \frac{10}{4} = 2.5 \text{ rad/sec.}$$

OLTF  $0.1 < \omega < \omega_1$

$$OLTF = \frac{k}{s \left( \frac{s}{\omega_1} \right)^1} = \frac{k \omega_1}{s^2}$$

$$24.1 = -40 \log \omega_1 + 20 \log k \omega_1$$

$$k \omega_1 = 100$$

$$k = \frac{100}{2.5} = 40$$

$$OLTF = \frac{40 \left( \frac{s}{25} + 1 \right)}{s \left( \frac{s}{2.5} + 1 \right) \left( \frac{s}{60} + 1 \right)}$$

$$OLTF = \frac{240(s+25)}{s(s+2.5)(s+60)}$$

12

Q.5 (b) A servo mechanism is represented by the equation :

$$\frac{d^2y}{dt^2} + 4.8 \frac{dy}{dt} = 144E$$

where  $E = C - 0.5y$  is the actuating signal. Find the value of damping ratio, damped and undamped frequency of oscillation. Draw the block diagram of the system described by the above equation.

[12 marks]

$$\frac{d^2y}{dt^2} + 4.8 \left( \frac{dy}{dt} \right) = 144 (C - 0.5y)$$

$$\frac{d^2y}{dt^2} + 4.8 \frac{dy}{dt} + 72y = 144C$$

taking Laplace inverse transform:

$$s^2Y(s) + 4.8sY(s) + 72Y(s) = 144(C/s)$$

$$\frac{Y(s)}{C(s)} = \left( \frac{144}{s^2 + 4.8s + 72} \right)$$

$$CE = s^2 + 4.8s + 72.$$

by comparing standard  $CE = s^2 + 2\zeta\omega_n s + \omega_n^2 = 0$

$$\omega_n = \sqrt{72} \text{ rad/sec.} = 8.485 \text{ rad/sec}$$

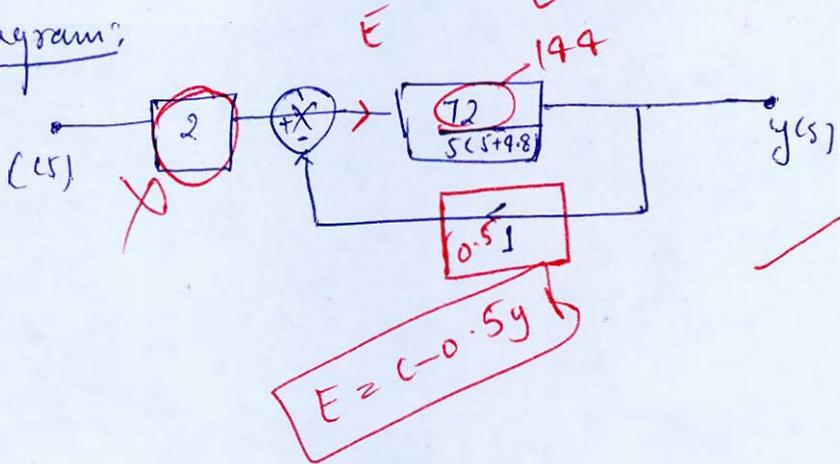
$$2\zeta\omega_n = 4.8$$

$$\zeta = \frac{2.4}{\sqrt{72}} = 0.2828$$

$$\text{damped freq} = \omega_n \sqrt{1-\zeta^2} = \sqrt{72} \sqrt{1-0.2828^2} = 8.138 \text{ rad/sec}$$

7

block diagram:



Q.5 (c) Closed loop system with unity feedback has the forward loop transfer function as :

$$G(s) = \frac{28.8}{s(1+0.2s)}$$

Modify the design using cascaded compensation to satisfy the optimum performance criterion, so that the transient response to unit step input reaches its final steady state value in minimum time without having any overshoot. Take gain of proportional controller equal to 5.

[12 marks]

modify by using PD controller ✓

$$\text{gain} = (K_p + K_D s) \quad \checkmark$$

$$\text{given } (K_p = 5)$$

$$\text{gain} = (5 + K_D s)$$

$$\text{Now } G_{\text{overall}}(s) = \frac{28.8(5 + K_D s)}{s(1 + 0.2s)}$$

CE (characteristic equation)

$$s^2 + 0.2s^2 + 28.8K_D s + 28.8 \times 5 = 0$$

by comparing standard CE,  $s^2 + 5s + 28.8K_D s + 28.8 \times 5 = 0$

$$\omega_n^2 = 28.8 \times 5 \quad \checkmark$$

$$\omega_n = 12.5 \text{ rad/sec.}$$

$$= 26.8328 \text{ rad/sec.}$$

$$2\zeta\omega_n = 28.8K_D + 5$$

no overshoot means

$$\zeta = 1 \quad \checkmark$$

$$2\zeta\omega_n = 5 + 28.8K_D$$

$$2 \times 12.5 \times 1 - 5 = 28.8K_D$$

$$K_D = \left( \frac{2 \times 1 \times 12 \sqrt{5} - 5}{28.98 \times 5} \right)$$

$$= 0.3379$$

$$PD \text{ Controller} = (5 + 0.338 S)$$

$$\text{Overall } G(s) = \frac{144(5 + 0.338 S)}{5(S + 5)}$$

12

Q.5 (d) A unity negative feedback system has open loop transfer function,  $G(s) = \frac{K}{s(1+sT)}$ , where

$K$  and  $T$  are positive constants. Determine the factor by which the amplifier gain  $K$  be reduced so that peak overshoot of the unit step response is reduced from 80% to 50%?

[12 marks]

$$G(s) = \frac{K}{s(1+sT)}$$

$$LE = s^2T + s + K$$

$$CE = s^2 + \frac{1}{T}s + \frac{K}{T}$$

$$\omega_n = \sqrt{\frac{K}{T}}, \quad 2\zeta\sqrt{\frac{K}{T}} = \frac{1}{T}$$

$$\zeta = \left( \frac{1}{2\sqrt{KT}} \right), \quad \zeta \propto \frac{1}{\sqrt{K}}$$

Let  $\zeta_1$

when overshoot is 80%.

$$\text{overshoot} = \frac{e^{-\zeta\pi/\sqrt{1-\zeta^2}}}{\sqrt{1-\zeta^2}}$$

$$e^{-\zeta_1\pi/\sqrt{1-\zeta_1^2}} = 0.8$$

$$\zeta_1 = \sqrt{\frac{\ln(0.8)^2}{\pi^2 + \ln(0.8)^2}}$$

$$\zeta_1 = 0.07$$

when overshoot is 50%.

$$e^{-\zeta_2\pi/\sqrt{1-\zeta_2^2}} = 0.5$$

$$\zeta_2 = \sqrt{\frac{\ln(0.5)^2}{\pi^2 + \ln(0.5)^2}}$$

$$\zeta_2 = 0.2159$$

$$\xi \propto \frac{1}{\sqrt{K}}$$

$$\xi^2 \propto \frac{1}{K}$$

$$K \propto \frac{1}{\xi^2}$$

Now:

$$\frac{K_1}{K_2} = \left( \frac{\xi_2^2}{\xi_1^2} \right)$$

$$\frac{K_1 \text{ at } (80\%)}{K_2 \text{ at } (50\%) \text{ overshoot}} = \left( \frac{0.2154^2}{0.07^2} \right)$$

$$\frac{K_1}{K_2} = 9.4735$$

$$K_2 = 0.1055 K_1$$

gain K reduced by 89.44%



Q.5 (e) The open loop transfer function of a unity negative feedback system is given as,

$$G(s) = \frac{K}{2s(1+0.1s)(1+s)}$$

system is 14 dB.

[12 marks]

$$G(s) = \frac{10K}{2s(s+1)(s+10)}$$

$$\angle G(j\omega) = -90^\circ - \tan^{-1}\left(\frac{\omega}{1}\right) - \tan^{-1}\left(\frac{\omega}{10}\right)$$

for finding phase cross over frequency

$$\angle G(j\omega) = -180^\circ$$

$$\angle G(j\omega) = -180^\circ$$

$$-90^\circ - \tan^{-1}(\omega) - \tan^{-1}\left(\frac{\omega}{10}\right) = -180^\circ$$

$$\tan^{-1}\left(\frac{\omega + \frac{\omega}{10}}{1 - \frac{\omega^2}{10}}\right) = 90^\circ$$

$$\left(\frac{\frac{11\omega}{10}}{1 - \frac{\omega^2}{10}}\right) = \frac{1}{0}$$

$$1 - \frac{\omega^2}{10} = 0$$

$$\omega = \sqrt{10} \text{ rad/sec}$$

$$\text{gain at } \omega_{pc} = \left(\frac{10K}{2\sqrt{10}\sqrt{1+10}\sqrt{100+10}}\right)$$

$$= \frac{10K}{2\sqrt{10} \times \sqrt{11} \times \sqrt{110}}$$

$$= \frac{10K}{11 \times 2 \times 10} = \frac{K}{22}$$

$$\text{Gain Margin} = 14 \text{ dB}$$

$$20 \log(GM) = 14$$

$$GM = 10^{0.7}$$

$$\text{gain} = \left( \frac{1}{10^{0.7}} \right) = \left( \frac{1}{5} \right)$$

$$\text{gain at wpc} = \left( \frac{k}{1/22} \right) = \frac{1}{5} \quad \checkmark$$

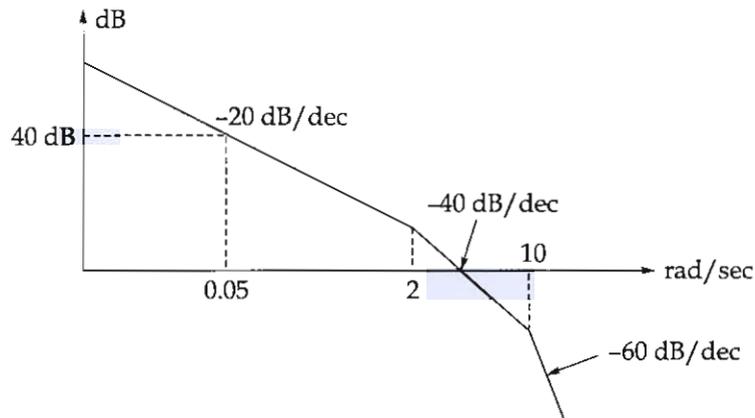
$$k = \frac{22}{5}$$

$$k = 4.4$$

at  $k = 4.4$  GM is 14 dB



- Q.6 (a) The open loop transfer function of a unity feedback system is given by  $G(s)H(s) = e^{-Ts}G_1(s)$ , where  $G_1(s)$  is minimum phase system. The approximate bode magnitude plot of the open loop transfer function is shown in the figure below. If the phase margin of the system is  $-18.19^\circ$ , determine the transportation lag  $T$ .

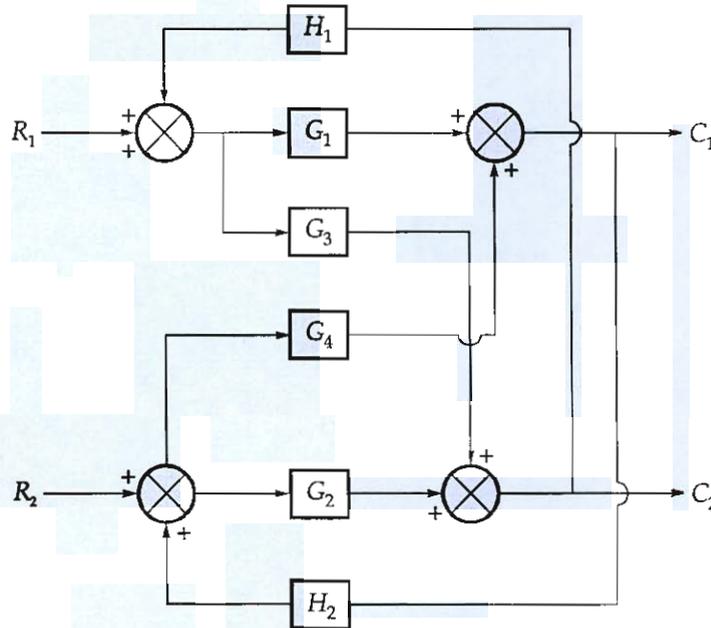


[20 marks]

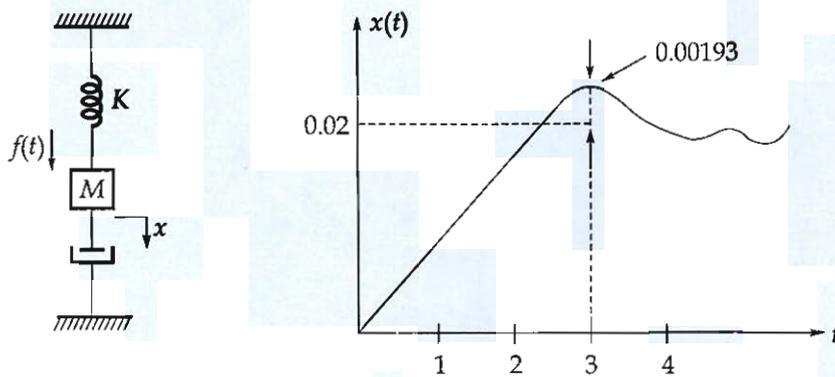




- Q.6 (b) (i) Evaluate  $\frac{C_2}{R_1}$  for the system whose block diagram representation is shown in figure below. (Use block diagram reduction technique to solve).



- (ii) Figure below shows a mechanical system and the response when 10 N of force is applied to the system. Determine the values of M, F, K. The dimension 'x' is in meter.



[10 + 10 marks]







Q.6 (c) Derive the expression for the transfer function of an ac servomotor and obtain the same in respect of a servomotor having following data :

- (i) Starting torque = 0.166 N-m
  - (ii) Moment of inertia,  $J = 1 \times 10^{-5} \text{ kgm}^2$
  - (iii) Supply voltage = 115 Volts
  - (iv) No load speed = 2904 rpm
- (Assume friction to be zero)

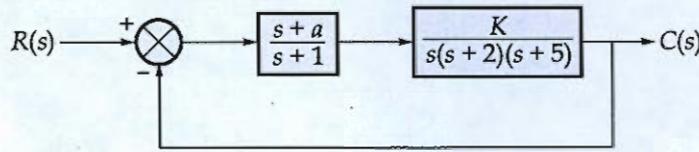
[15 + 5 = 20 marks]







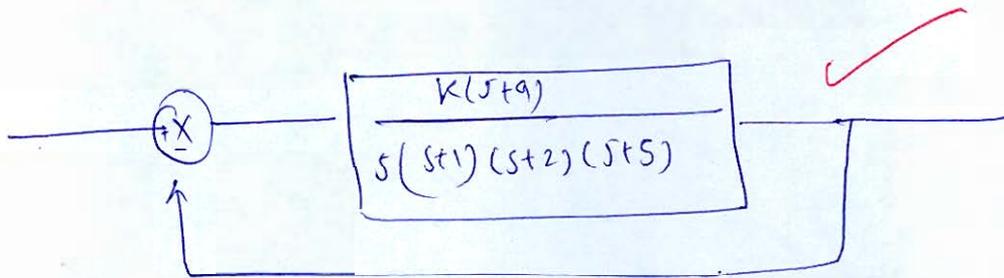
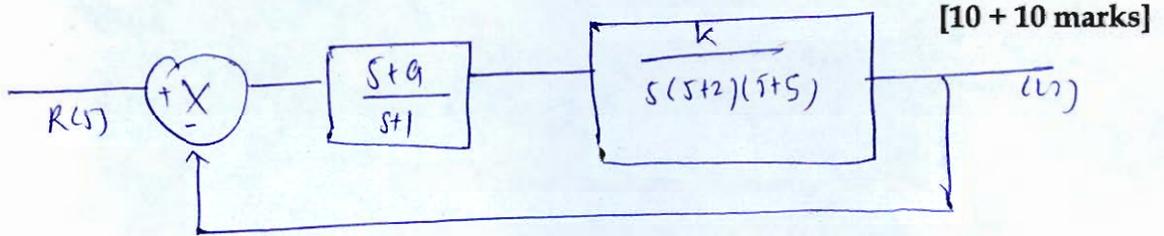
Q.7 (a) (i) A position control system is shown in figure below :



$K$  and  $a$  are the parameters of the system. Determine the range of  $K$  and  $a$  for which system is stable.

(ii) Sketch the root-locus of  $G(s) = \frac{K(s+1)}{s^2(s+2)}$ .

7001)



$$1 + G(s)H(s) = 1 + \frac{K(s+1)}{s(s+1)(s+2)(s+5)} = 0$$

$$s(s+1)(s+2)(s+5) + K(s+1) = 0$$

$$s^4 + 8s^3 + 17s^2 + (10+K)s + Ka = 0$$

apply hurwitz criteria to find stability.

$$s^4 \quad 1 \quad 17 \quad Ka$$

$$s^3 \quad 8 \quad 10+K$$

$$s^2 \quad \frac{136-10-K}{8} \quad Ka$$

$$s^1 \quad \frac{\left(\frac{126-K}{8}\right)(10+K) - 8Ka}{\left(\frac{126-K}{8}\right)}$$

$$s^0 \quad Ka$$

for stable system  
first column of all element should have same  
sign

$$\frac{126-k}{8} > 0$$

$$k < 126$$

$$ka > 0$$

$$a > 0$$

$$k > 0$$

$$\left(\frac{126-k}{8}\right)(10+k) - 6ka > 0$$

$$(126-k)(10+k) - 64ka > 0$$

$$64ka < (126-k)(10+k)$$

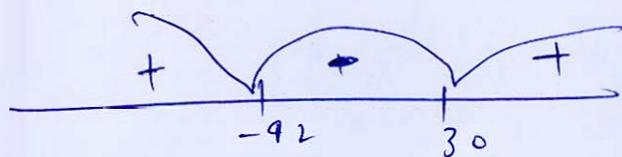
let  $a=2$ .

$$128k < 1260 - 10k + 126k - k^2$$

$$k^2 - 116k + 128k - 1260 < 0$$

$$k^2 + 12k - 1260 < 0$$

$$(k-30)(k+42) < 0$$



$$k \in (-42, 30)$$

$$0 < k < 30$$

$$\vee k < 126$$

when  $a=2$ ,  $k$  lies between  $(0, 30)$

$$\boxed{0 < k < 30} \quad a=2$$

10

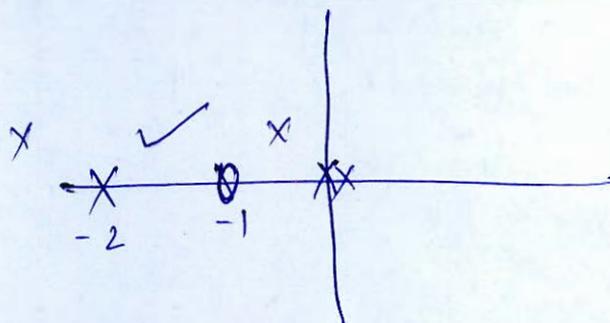
7a (ii)

$$G(s) = \frac{k(s+1)}{s^2(s+2)}$$

(i)

no of poles = 3

no of zero = 1.



(ii)

$$\text{Centroid} = \frac{-0-2-(-1)}{2}$$

$$= \frac{-2+1}{2} = -0.5$$

(iii)

$$\text{asymptotes} = \frac{(2k+1)\pi}{P-Z}$$

$$= \frac{(2k+1)\pi}{2}$$

$$= 90^\circ, 270^\circ$$

(iv)

$$k = - \frac{(s^3 + 2s^2)}{s+1}$$

$$\frac{dk}{ds} = - \left[ \frac{(s+1)(3s^2+4s) - (s^3+2s^2)(1)}{(s+1)^2} \right]$$

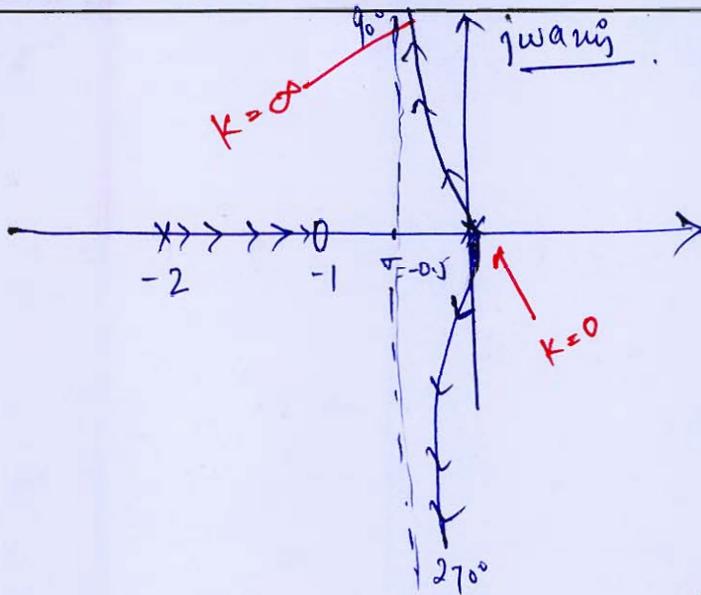
$$0 = 3s^3 + 3s^2 + 4s^2 + 4s - s^3 - 2s^2$$

$$2s^3 + 5s^2 + 4s = 0$$

$$s(2s^2 + 5s + 4) = 0$$

$$s = 0, -1.25 \pm 0.66143j$$

only one break point lies at  $s=0$  on  
real axis.



Ans.

$$s^3 + 2s^2 + ks + k = 0$$

$$1 \quad k$$

$$2 \quad k$$

$$\frac{2k - k}{2} = \frac{k}{2}$$

$$k$$

when  $k > 0$

always stable.

08

- Q.7 (b) Sketch the polar plot of the transfer function given below. Determine whether the plot crosses the real axis. If so, determine the frequency at which the plot cross the real axis and the corresponding magnitude  $|G(j\omega)|$ .

$$G(s) = \frac{1}{s^2(1+s)(1+2s)}$$

[20 marks]

Polar plot

$$s = j\omega$$

$$G(j\omega) = \frac{1}{(j\omega)^2 (j\omega+1) (2j\omega+1)}$$

$$= \frac{1 (-j\omega+1) (-2j\omega+1)}{-\omega^2 (\omega^2+1) (4\omega^2+1)}$$

$$G(j\omega) = \frac{(-2\omega^2+1)}{-\omega^2(\omega^2+1)(4\omega^2+1)} + \frac{j3}{\omega(\omega^2+1)(4\omega^2+1)}$$

at  $\omega=0$ 

$$G(j\omega) = -\infty + j\infty$$

at  $\omega \rightarrow \infty$ 

$$G(j\omega) = 0 + j0$$

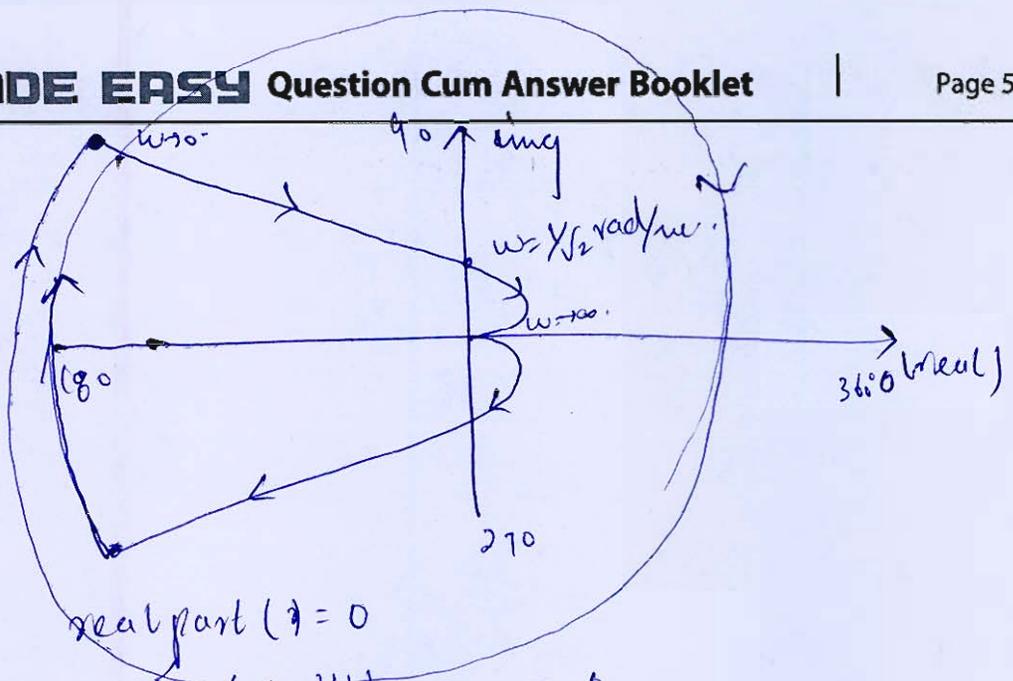
$$\angle G(j\omega) = -180^\circ - \tan^{-1}(\omega) - \tan^{-1}(2\omega)$$

at  $\omega=0$ 

$$\angle G(j\omega) = -180^\circ$$

at  $\omega=\infty$ 

$$\angle G(j\omega) = -360^\circ$$



$$\frac{-2w^2 + 1}{-w^2(w^2 + 1)(4w^2 + 1)} = 0$$

$$w = \frac{1}{\sqrt{2}} \text{ rad/sec.}$$

at  $w = \frac{1}{\sqrt{2}} \text{ rad/sec}$  polar plot cut at imaginary axis.

when  $R \rightarrow \infty$

$$G(Re^{i\theta}) = \frac{1}{R^2 e^{i2\theta} (Re^{i\theta}) (2Re^{i\theta})}$$

$$= \frac{1}{2R^4} e^{-i4\theta}$$

18

2 semicircle from.

# Polar plot does not intersect at real axis hence w. is undefined

Q.7 (c) Construct the state model for a system characterised by the differential equation :

$$\frac{d^3y}{dt^3} + \frac{6d^2y}{dt^2} + \frac{11dy}{dt} + 6y = u$$

Give the block diagram representation of the state model.

[15 + 5 = 20 marks]

$$\frac{d^3y}{dt^3} + \frac{6d^2y}{dt^2} + \frac{11dy}{dt} + 6y = u$$

Let  $y = x_1$  ——— (i)

$$\frac{dy}{dt} = \dot{x}_1 = x_2$$
 ——— (ii)

$$\frac{d^2y}{dt^2} = \dot{x}_2 = x_3$$
 ——— (iii)

$$\frac{d^3y}{dt^3} = \dot{x}_3$$

$$\dot{x}_3 + 6x_3 + 11x_2 + 6x_1 = u$$

State model:

$$\dot{x}_3 = -6x_1 - 11x_2 - 6x_3 + u$$
 ——— (iv)

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -6 & -11 & -6 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} u$$

$$y = [1 \ 0 \ 0]$$

$$A = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -6 & -11 & -6 \end{bmatrix} \quad B = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$$

$$C = [1 \ 0 \ 0] \quad , \quad D = [0]$$

$$\frac{d^3y}{dt^3} + 6\frac{d^2y}{dt^2} + 11\frac{dy}{dt} + 6y = u$$

by applying inverse laplace transform

$$Y(s) (s^3 + 6s^2 + 11s + 6) = U(s)$$

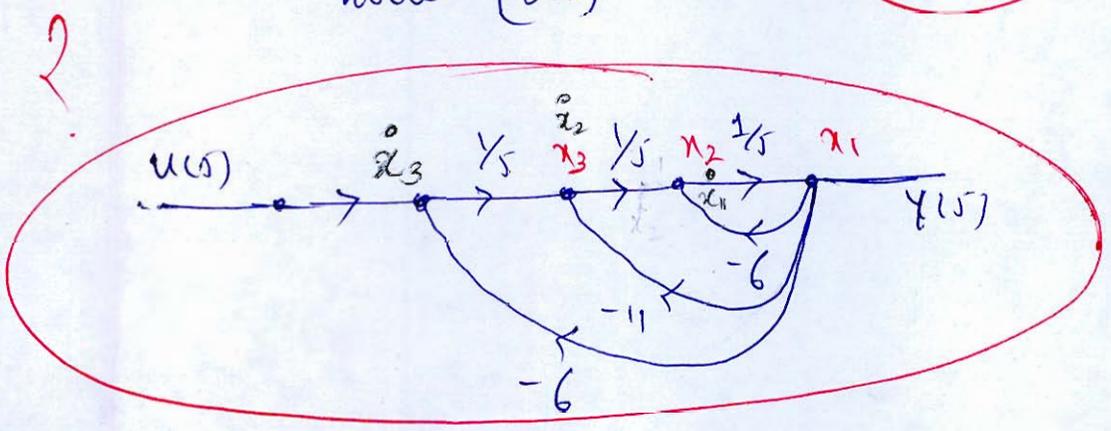
$$\frac{Y(s)}{U(s)} = TIF = \left( \frac{1}{s^3 + 6s^2 + 11s + 6} \right)$$

$$TIF = \left( \frac{1/s^3}{1 + 6/s + 11/s^2 + 6/s^3} \right)$$

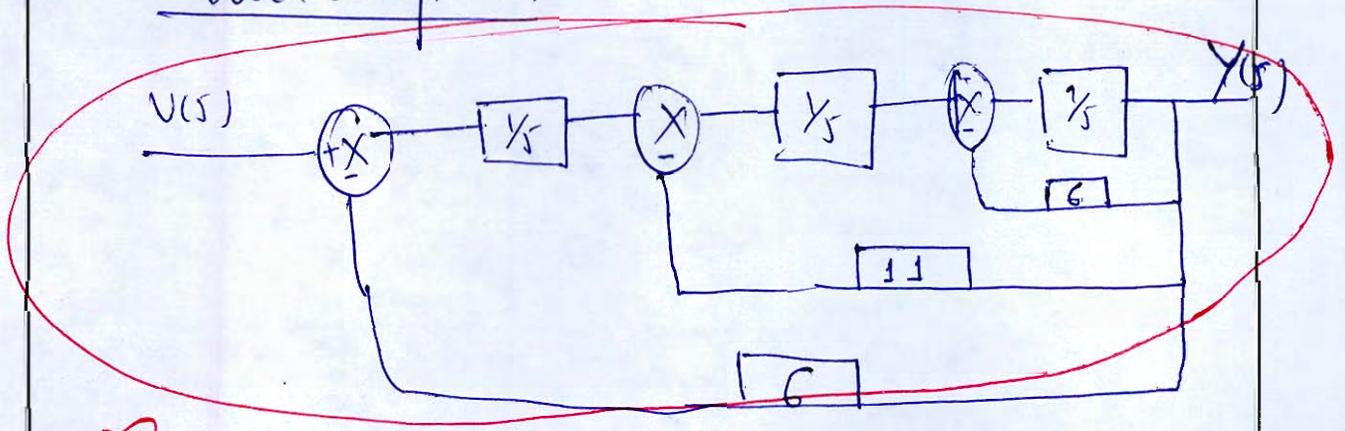
order of transfer fn = 3

node = (3+2) = 5

15



block diagram:



Q.8 (a) The open-loop transfer function of a unity feedback control system is given below:

$$G(s) = \frac{K}{s(s+2)(s^2+2s+2)}$$

Plot the root locus and determine the value of  $K$  at the breakaway point.

[20 marks]





2.8 (b) The open loop transfer function of a feedback control system is

$$G(s)H(s) = \frac{K(1+2s)}{s(1+s)(1+s+s^2)}$$

Find the restriction on  $K$  for stability. Find the value of  $K$  for the system to have a gain margin of 3 dB. With this value of  $K$ , find the gain cross over frequency and phase margin. Use Nyquist Approach.

[20 marks]







- 2.8 (c) The state space model of a second order system given below is designed using feedback control system.

$$\dot{x} = \begin{bmatrix} 2 & 3 \\ -1 & 4 \end{bmatrix} x + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u$$

$$y = [1 \ 0] x$$

- (i) What are the conditions for the desired response? Also check whether desired response is possible or not.
- (ii) Design an observer system such that the above system has settling time of 0.5 sec and damping frequency of 6 rad/sec.

[8 + 12 marks]





**Space for Rough Work**

---



**Space for Rough Work**

---

**Space for Rough Work**

---



$$\frac{999}{5(5+5)}$$

$$\frac{199(5+5)}{5^2+5}$$

$$5^2+5 + 199 \times 5 + 199 \times 5$$

$$w_4 = 12\sqrt{5}$$

$$\frac{212\sqrt{5}-5}{179}$$