

# CLASS TEST

S.No. : 04 SP\_EC\_W+Y\_110819

Network Theory



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# CLASS TEST 2019-2020

## ELECTRONICS ENGINEERING

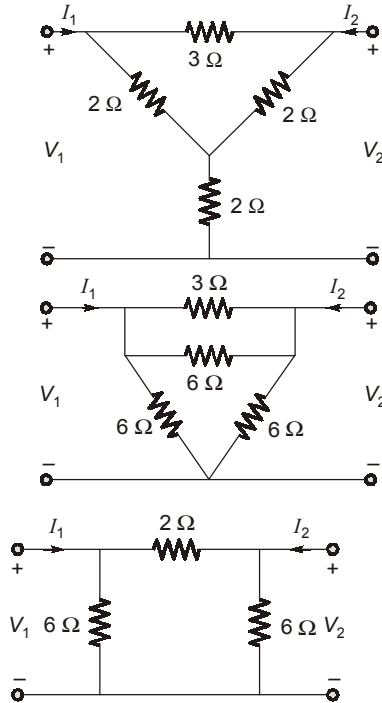
Date of Test : 11/08/2019

### ANSWER KEY > Network Theory

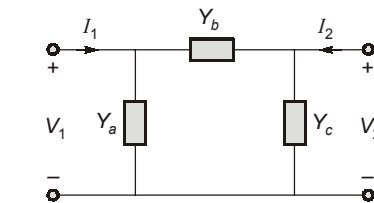
- |        |         |         |         |         |
|--------|---------|---------|---------|---------|
| 1. (b) | 7. (d)  | 13. (a) | 19. (b) | 25. (d) |
| 2. (c) | 8. (c)  | 14. (c) | 20. (b) | 26. (a) |
| 3. (b) | 9. (a)  | 15. (b) | 21. (b) | 27. (b) |
| 4. (b) | 10. (b) | 16. (c) | 22. (c) | 28. (b) |
| 5. (a) | 11. (c) | 17. (a) | 23. (d) | 29. (b) |
| 6. (b) | 12. (d) | 18. (c) | 24. (b) | 30. (d) |

**Detailed Explanations**

1.(b)



for II-network



$$\therefore [Y] = \begin{bmatrix} Y_a + Y_b & -Y_b \\ -Y_b & Y_b + Y_c \end{bmatrix}$$

for the given problem,  $Y_a = \frac{1}{6} \text{ } \Omega^{-1}$

$$Y_b = \frac{1}{2} \text{ } \Omega^{-1}$$

$$Y_c = \frac{1}{6} \text{ } \Omega^{-1}$$

$$\therefore [Y] = \begin{bmatrix} \frac{2}{3} \text{ } \Omega^{-1} & -\frac{1}{2} \text{ } \Omega^{-1} \\ -\frac{1}{2} \text{ } \Omega^{-1} & \frac{2}{3} \text{ } \Omega^{-1} \end{bmatrix}$$

2.(c)

$$\text{Time period } (T) = \frac{2\pi}{\omega}$$

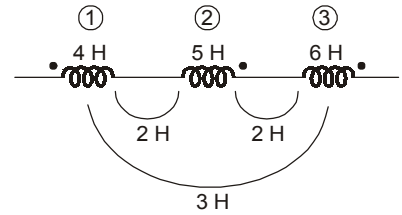
where  $\omega = \frac{1}{\sqrt{LC}}$

thus,

$$T = 2\pi\sqrt{LC}$$

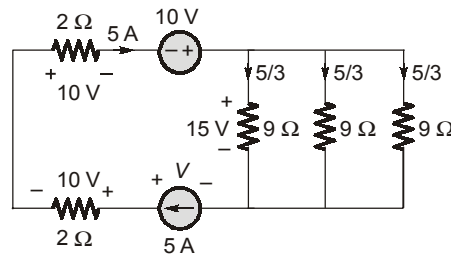
In figure

$$\begin{aligned} L_{eq} &= L_1 + L_2 + L_3 - 2M_{12} + 2M_{23} - 2M_{13} \\ &= 4 + 5 + 6 - 2(2) + 2(2) - 2(3) = 9 \text{ H} \\ C &= 1 \text{ F} \\ T &= 2\pi\sqrt{9} = 6\pi \text{ sec} \end{aligned}$$



**3.(b)**

The circuit can be redrawn by short circuiting inductor and open circuiting capacitor as DC sources are used.



Applying KVL

$$\begin{aligned} V - 10 - 10 + 10 - 15 &= 0 \\ V &= 25 \text{ V} \end{aligned}$$

**4.(b)**

$$\frac{R}{2}\sqrt{\frac{L}{C}} \Rightarrow \frac{R}{2}\sqrt{\frac{L\omega}{C\omega}} = \frac{R}{2}\sqrt{X_L X_C}$$

Unit of  $R_1$  is  $\Omega$

Unit of ' $X_L$ ' and ' $X_C$ ' is  $\Omega$

$$\text{Unit of } \frac{R}{2}\sqrt{\frac{L}{C}} \text{ is } \Omega \times \sqrt{\Omega \times \Omega} \Rightarrow (\Omega)^2$$

**5.(a)**

The average value of periodic signal can be calculated by considering one time period

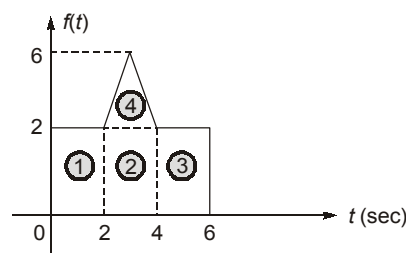
$$= \frac{\text{Total area under the graph for one period}}{T_0}$$

Total area under the graph for one period = Area 1 + Area 2 + Area 3 + Area 4

here Area 1 = Area 2 = Area 3 = Area 4 = 4

and  $T_0 = 8 \text{ sec}$

$$\text{Average value} = \frac{4 + 4 + 4 + 4}{8} = \frac{16}{8} = 2$$

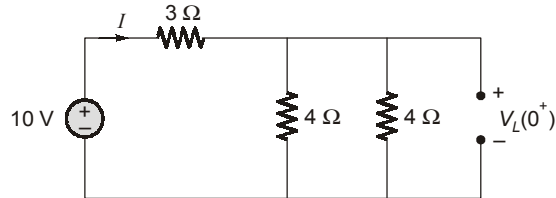


6.(b)

Before closing the switch, the circuit was not energized, therefore, current through inductor and voltage across capacitor are zero.

After closing the switch, at  $t = 0^+$  inductor acts as open-circuit and capacitor acts as short-circuit.

Equivalent circuit at  $t = 0^+$



$$I = \frac{10}{3 + 4 \parallel 4} = 2 \text{ A}$$

$$V_L(0^+) = I \times (4 \parallel 4)$$

$$= 2 \times 2 = 4 \text{ V}$$

7.(d)

Applying KVL in both the loops we get

$$V_1 = (j\omega L_1)I_1 + (j\omega M)I_2$$

$$V_2 = (j\omega L_2)I_2 + (j\omega M)I_1$$

$$\frac{V_2}{V_1} = \frac{L_2 I_2 + M I_1}{L_1 I_1 + M I_2}$$

also,

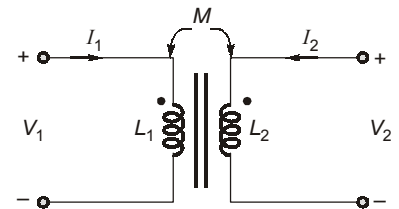
$$M = K\sqrt{L_1 L_2} = \sqrt{L_1 L_2};$$

$\therefore K = 1$  for ideal transformer

$$\frac{V_2}{V_1} = \frac{L_2 I_2 + \sqrt{L_1 L_2} I_1}{L_1 I_1 + \sqrt{L_1 L_2} I_2}$$

$$\therefore \frac{V_2}{V_1} = \frac{\sqrt{L_2}}{\sqrt{L_1}}$$

$$\Rightarrow \frac{V_2}{V_1} = \sqrt{\frac{5}{25}} = \frac{1}{\sqrt{5}}$$



8.(c)

As we know,

$$\text{Real power} = V_{\text{rms}} \cdot I_{\text{rms}} \cos \phi$$

...(i)

$$\text{Reactive power} = V_{\text{rms}} \cdot I_{\text{rms}} \sin \phi$$

...(ii)

$$\text{Apparent power} = V_{\text{rms}} \cdot I_{\text{rms}}$$

...(iii)

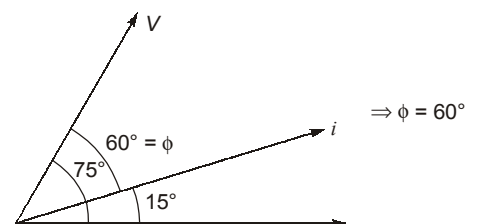
Given that  $v(t) = 10\cos(2t + 75^\circ)$

$i(t) = 2\cos(2t + 15^\circ)$

from equation (i)

$$\text{Real power} = \frac{10}{\sqrt{2}} \times \frac{2}{\sqrt{2}} \times \frac{1}{2} = 5 \text{ Watts}$$

from equation (ii)



$$\text{Reactive power} = \frac{10}{\sqrt{2}} \times \frac{2}{\sqrt{2}} \times \frac{\sqrt{3}}{2} = 5\sqrt{3} = 8.66 \text{ VAR}$$

from equation (iii)

$$\text{Apparent power} = \frac{10}{\sqrt{2}} \times \frac{2}{\sqrt{2}} = 10 \text{ VA}$$

9.(a)

$$Y = Y_1 + Y_2$$

$$Y = \frac{1}{R + jX_L} + \frac{1}{R - jX_C}$$

$$Y = \frac{R - jX_L}{(R^2 + X_L^2)} + \frac{(R + jX_C)}{(R^2 + X_C^2)}$$

$$\text{Im}(Y) = \frac{-X_L(R^2 + X_C^2) + X_C(R^2 + X_L^2)}{(R^2 + X_L^2)(R^2 + X_C^2)}$$

For 'Z' to purely resistive

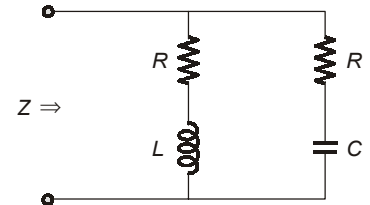
also  $\text{Im}(Y) = 0$

$$\Rightarrow X_L(R^2 + X_C^2) = X_C(R^2 + X_L^2)$$

$$R^2 X_L + X_C^2 X_L = R^2 X_C + X_L^2 X_C$$

$$R^2(X_L - X_C) = X_L X_C(X_L - X_C)$$

$$R^2 = X_L X_C = \omega L \times \frac{1}{\omega C} = \frac{L}{C}$$



10.(b)

For parallel resonant circuit

$$Q_0 = R \sqrt{\frac{C}{L}}$$

$$Q_0 = 2000 \sqrt{\frac{54 \times 10^{-6}}{240 \times 10^{-3}}}$$

$$Q_0 = 2000 \sqrt{\frac{9}{4} \times 10^{-4}}$$

$$Q_0 = \frac{2000}{100} \times \frac{3}{2}$$

$$Q_0 = 30$$

11.(c)

$$L_{\text{eq}} = (L + L - 2M) \parallel L$$

Also

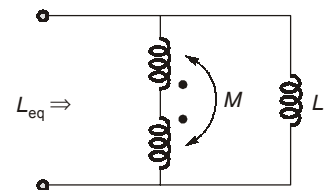
$$M = k\sqrt{L_1 L_2} = M = k\sqrt{L^2} = kL$$

$$L_{\text{eq}} = (L + L - 2kL) \parallel L$$

$$\frac{L}{3} = \frac{(2L - 2kL) \times L}{2L - 2kL + L}$$

on solving, we get

$$k = 0.75$$



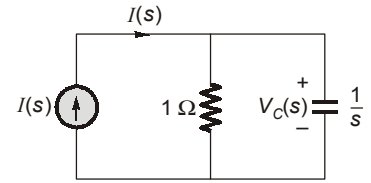
12.(d)

$$V_C(s) = I(s) \times \frac{1}{1 + \frac{1}{s}} \times \frac{1}{s} = I(s) \times \frac{1}{s+1}$$

$$I(s) = \frac{2(s+1)}{(s+1)^2 + 1}$$

$$V_C(s) = \frac{2(s+1)}{(s+1)^2 + 1} \times \frac{1}{1+s} = \frac{2}{(s+1)^2 + 1}$$

$$v_C(t) = 2e^{-t} \sin t u(t) \text{ V}$$



13.(a)

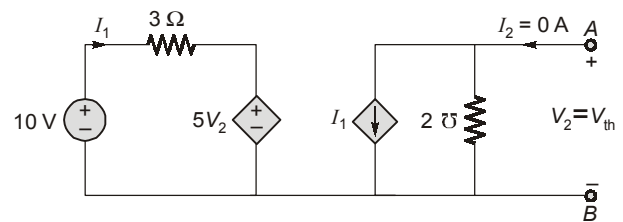
to determine  $V_{th}$ :

$$I_1 = \frac{10 - 5V_2}{3} = \frac{10 - 5V_{th}}{3}$$

$$V_{th} = -\frac{I_1}{2} = \frac{5V_{th} - 10}{6}$$

$$6V_{th} = 5V_{th} - 10$$

$$V_{th} = -10 \text{ V}$$



to determine  $R_{th}$ :

$$I_2 = 2V_2 + I_1$$

$$1 \text{ A} = 2V_2 + I_1$$

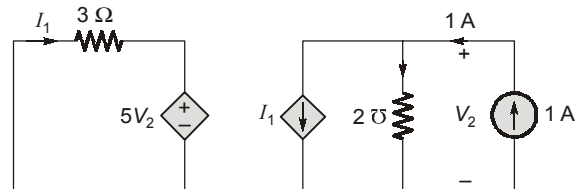
$$0 = 3I_1 + 5V_2$$

$$I_1 = -\frac{5}{3}V_2$$

$$1 \text{ A} = 2V_2 - \frac{5}{3}V_2$$

$$V_2 = 3 \text{ V}$$

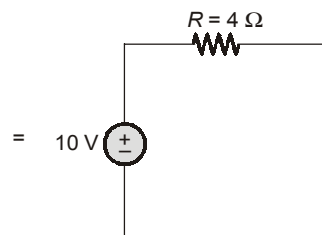
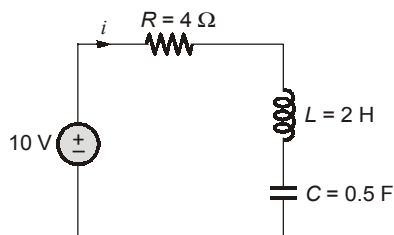
$$R_{th} = \frac{V_2}{1 \text{ A}} = 3 \Omega$$



$$P_{Lmax} = \frac{V_{th}^2}{4R_{th}} = \frac{100}{12} \text{ W} = 8.33 \text{ W}$$

14.(c)

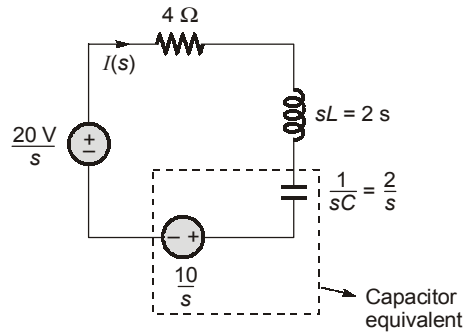
At  $(t = 0^-)$



$$V_C(0^-) = 10 \text{ V}$$

$$i_L(0^-) = 0 \text{ A}$$

At  $(t = 0^+)$



$$I(s) = \frac{10/s}{4 + 2s + \frac{2}{s}} = \frac{10/s}{\frac{4s + 2s^2 + 2}{s}} = \frac{10}{2(s^2 + 2s + 1)}$$

$$I(s) = \frac{5}{(s+1)^2}$$

$$i(t) = 5te^{-t} u(t) \text{ A}$$

15.(b)

For a series  $RL$  circuit with DC excitation,

$$i(t) = \frac{V_s}{R} \left( 1 - e^{-\frac{Rt}{L}} \right) u(t) \text{ A}$$

$$v(t) = V_s \left( e^{-\frac{Rt}{L}} \right) u(t) \text{ A}$$

$$A = \frac{V_s}{R}$$

$$B = \frac{R}{L}$$

$$C = V_s$$

$$\frac{AB}{C} = \frac{\frac{V_s}{R} \times \frac{R}{L}}{V_s} = \frac{1}{L}$$

$$= \frac{1}{5 \times 10^{-3}} = 200$$

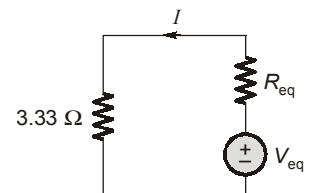
16.(c)

Applying Millman's Theorem

$$\frac{1}{R_{eq}} = \frac{1}{10} + \frac{1}{30} + \frac{1}{90} + \dots$$

$$\frac{1}{R_{eq}} = \frac{1}{10} \left( 1 + \frac{1}{3} + \frac{1}{9} + \dots \right)$$

$$\frac{1}{R_{eq}} = \frac{1}{10} \left( \frac{1}{1 - \frac{1}{3}} \right) = \frac{3}{10 \times 2} = \frac{3}{20}$$



$$R_{eq} = \frac{20}{3} \Omega$$

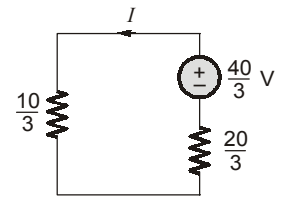
$$V_{eq} = \frac{\frac{E_1}{R_1} + \frac{E_2}{R_2} + \frac{E_3}{R_3} + \dots}{\frac{1}{R_{eq}}}$$

$$= \frac{1 + \frac{1}{2} + \frac{1}{4} + \frac{1}{8} + \dots}{\frac{3}{20}} = \frac{1 - \frac{1}{2}}{\frac{3}{20}} = \frac{20}{3} \times 2$$

$$V_{eq} = \frac{40}{3} \text{ V}$$

$$I = \frac{\frac{40}{3}}{\frac{20}{3} + \frac{10}{3}} = \frac{40}{30} = \frac{4}{3} \text{ A}$$

$$I = 1.33 \text{ A}$$



17.(a)

At  $(t = 0^-)$ , both the switches are opened.

$L$  is initially uncharged  $i_L(0^-) = 0$

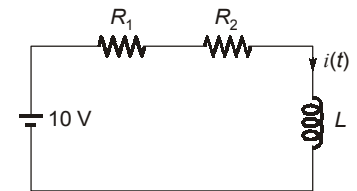
At  $(t = 0^+)$

$$i(t) = i(\infty) + (i(0^+) - i(\infty)) e^{-\frac{R_{eq}t}{L_{eq}}}$$

$$R_{eq} = 5 \Omega$$

$$L_{eq} = 1 \text{ H}$$

$$i(0^+) = 0 \text{ A}$$



$$i(t) = 2 + (0 - 2)e^{-\frac{5t}{1}} \text{ A ; for } t > 0$$

At  $(t = 2^-)$

$$i(2^-) = 2 - 2e^{-\frac{10}{1}} \text{ A}$$

$$i(2^-) \approx 2 \text{ A}$$

At  $(t = 2^+)$

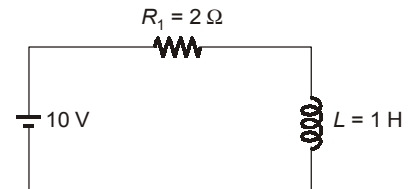
$$i(2^-) = i(2^+) = 2 \text{ A}$$

for  $t > 2$  sec

$$i(t) = i(\infty) + (i(2^+) - i(\infty)) e^{-\frac{R_1(t-2)}{L}} ; \text{ for } t > 2$$

$$i(t)|_{t=3s} = 5 + (2 - 5)e^{-\frac{2}{1}(3-2)} = 5 - 3e^{-2}$$

$$= 4.594 \text{ A}$$

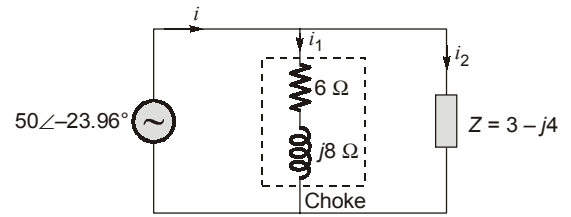




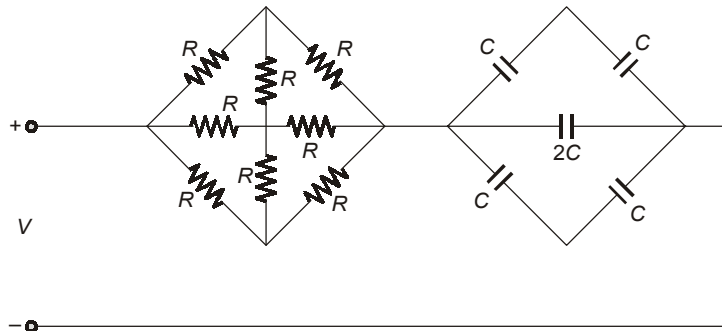
18.(c)

Here, applying KCL

$$\begin{aligned}
 i(t) &= i_1(t) + i_2(t) \\
 &= \frac{V_1}{6 + j8} + \frac{V_1}{3 - j4} \\
 \Rightarrow V_1 \left( \frac{3 - j4 + 6 + j8}{2(3 + j4)(3 - j4)} \right) &= \frac{V_1(9 + j4)}{2 \times 25} \\
 \Rightarrow \frac{50 \angle -23.96^\circ \times \sqrt{97} \angle 23.96^\circ}{50} \text{ A} &= \sqrt{97} \text{ A} \\
 &= 9.85 \text{ A}
 \end{aligned}$$

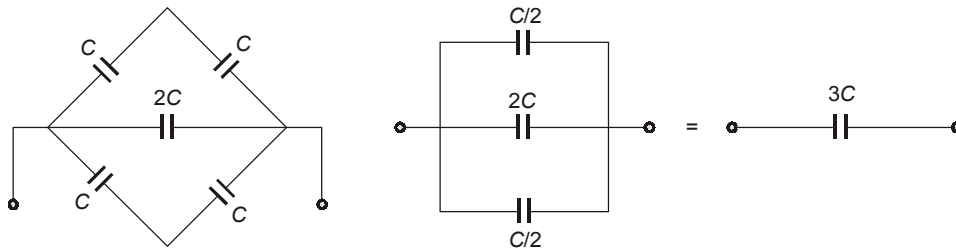


19.(b)

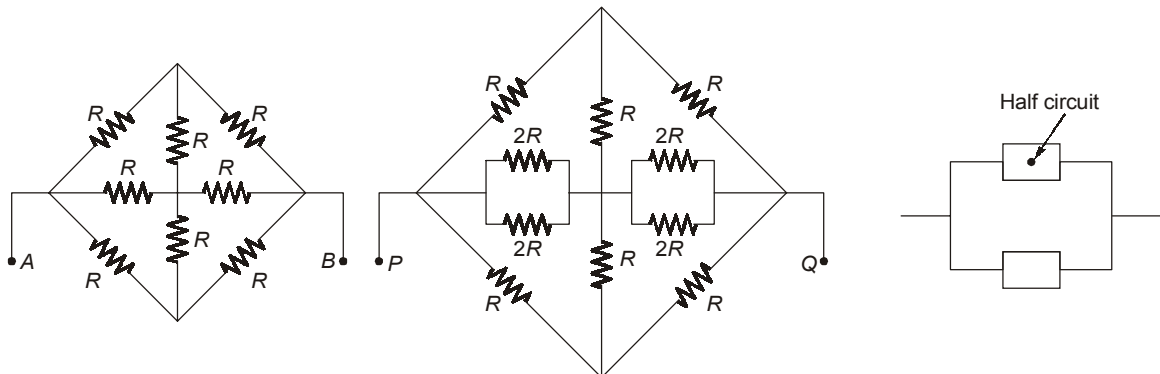


The time constant of an RC circuit is  $\tau = R_{eq}C_{eq}$

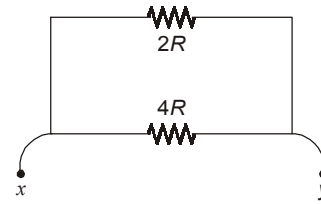
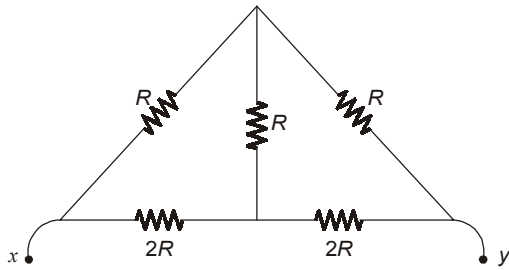
Calculation of  $C_{eq}$



Calculation of  $R_{eq}$



It is Wheatstone bridge.



$$(R_{eq})_{\text{Half circuit}} = \frac{4R \times 2R}{6R} = \frac{4R}{3}$$

$$R_{eq} = (R_{eq})_{\text{Half circuit}} \parallel (R_{eq})_{\text{Half circuit}}$$

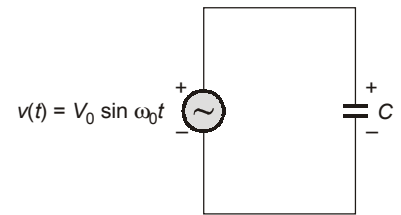
$$= \left(\frac{4R}{3}\right) \parallel \left(\frac{4R}{3}\right) = \left(\frac{2R}{3}\right)$$

∴

$$\begin{aligned} \tau &= R_{eq} C_{eq} \\ &= \frac{2R}{3} \times 3C = 2RC \end{aligned}$$

20.(b)

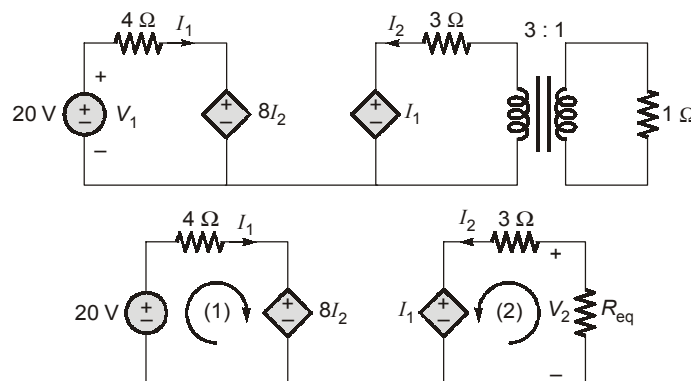
$$\begin{aligned} W(t) &= \frac{1}{2} C v^2(t) \\ &= \frac{1}{2} C V_0^2 \sin^2 \omega_0 t \\ &= \frac{1}{4} C V_0^2 (1 - \cos 2\omega_0 t) \end{aligned}$$



thus only option (b) satisfies this condition.

21.(b)

Network 'N' can be replaced as



$$R_{eq} = (1) \times \left(\frac{3}{1}\right)^2 = 9 \Omega$$

Applying KVL at loop (1)

$$20 = 4I_1 + 8I_2$$

also

$$V_2 = -9I_2 = I_1 + 3I_2$$

$$\begin{aligned} \Rightarrow I_1 &= -12I_2 \\ 20 &= 4(-12I_2) + 8I_2 \\ 20 &= -40I_2 \\ \Rightarrow I_2 &= -0.5 \text{ A} \\ I_1 &= -12(-0.5) = 6 \text{ A} \end{aligned}$$

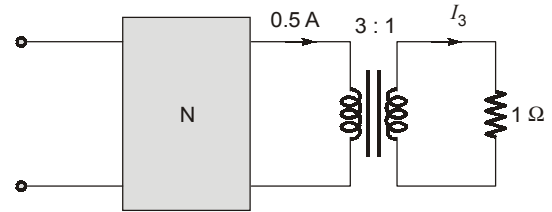
Now,

$$\frac{I_{\text{primary}}}{I_{\text{secondary}}} = \frac{1}{3}$$

$$\begin{aligned} \Rightarrow 3I_{\text{primary}} &= I_{\text{secondary}} \\ 3(-I_2) &= I_3 \end{aligned}$$

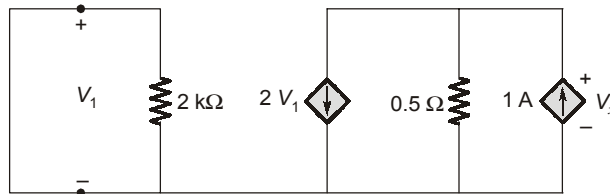
$$\Rightarrow I_3 = 1.5 \text{ A}$$

$$\begin{aligned} \text{Power delivered to } 1 \Omega &= (I_3)^2 \times R_L = (1.5)^2 \times 1 \\ &= 2.25 \text{ Watts} \end{aligned}$$



22.(c)

Calculating  $R_{th}$



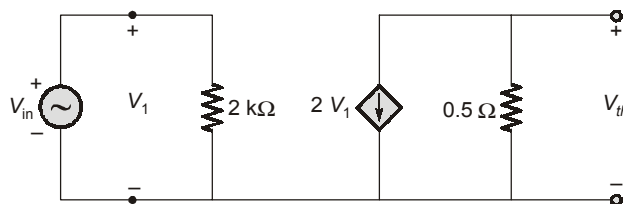
$$1 \text{ A} = \frac{V_x}{0.5} + 2V_1$$

$\therefore V_1 = 0$   
independent voltage source is short circuited

(As

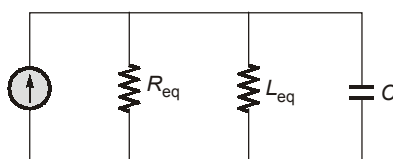
$$\begin{aligned} \Rightarrow V_x &= 0.5 \text{ V} \\ R_{th} &= 0.5 \Omega = 500 \text{ m}\Omega \end{aligned}$$

Calculating  $V_{th}$



$$\begin{aligned} V_{th} &= -2V_1 \times 0.5 \\ V_1 &= V_{in} = 5 \angle 0^\circ \text{ V} \\ V_{th} &= -2 \times 5 \times 0.5 = -5 \angle 0^\circ \\ &= 5 \angle 180^\circ \text{ V} \end{aligned}$$

23.(d)

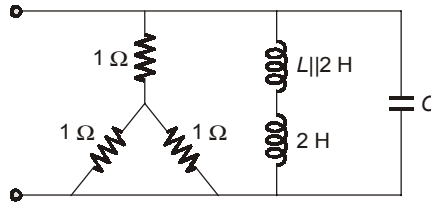


For a parallel resonant circuit

the damping ratio  $\xi = \frac{1}{2Q} = \frac{1}{2R_{eq}} \sqrt{\frac{L_{eq}}{C}} = \frac{1}{\sqrt{2}}$

...(i)

given



$$R_{eq} = 1 \parallel 1 + 1 = 1.5 \Omega$$

$$L_{eq} = 2 + \frac{2L}{L+2}$$

$$C_{eq} = \frac{4}{9}$$

From equation (i)

$$\frac{L_{eq}}{C} = \frac{4R_{eq}^2}{2} = 2R_{eq}^2$$

or

$$L_{eq} = 2R_{eq}^2 C = 2 \times (1.5)^2 \times \frac{4}{9} = 2 \text{ H}$$

∴

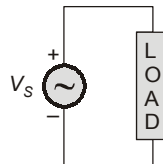
$$L_{eq} = 2H + \frac{2L}{L+2}$$

$$2 + \frac{2L}{L+2} = 2 \text{ H}$$

⇒

$$L = 0 \text{ H}$$

24.(b)



$pf = 0.50$  lagging

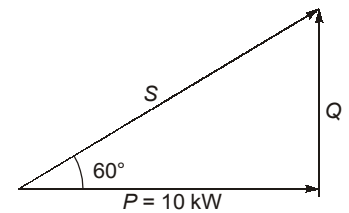
$\cos\phi = 0.5$

$\phi = 60^\circ$

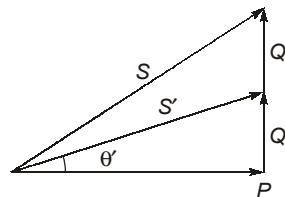
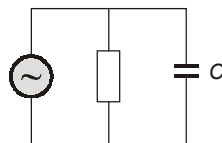
⇒

$$Q = P \tan 60^\circ = 10\sqrt{3} \text{ kVAR}$$

$$S = \sqrt{P^2 + Q^2} = 20 \text{ kVA}$$



Now,



As per question

$$S' = 14.14 = 10\sqrt{2} \text{ kVA}$$

$$P' = 10 \text{ kW}$$

$$\cos \theta' = \frac{\rho}{s} = \frac{10}{10\sqrt{2}} = \frac{1}{\sqrt{2}}$$

$$\theta' = 45^\circ$$

$$Q' = 10 \text{ kVAR}$$

Reduction in reactive power

$$\begin{aligned} &= (10\sqrt{3} - 10) \text{ kVAR} \\ &= 10(\sqrt{3} - 1) = 10(0.732) \\ &= 7.32 \text{ kVAR} \end{aligned}$$

25. (d)

Let  $v_{in}(t)$  be the input voltage while  $v_{out}(t)$  be the output voltage

$$h(t) = (e^{-2t} + e^{-3t}) u(t) \text{ V}$$

$$\therefore H(s) = \frac{V_{out}(s)}{V_{in}(s)} = \frac{1}{s+2} + \frac{1}{s+3} = \frac{2s+5}{(s+2)(s+3)}$$

if  $v_{out}(t) = te^{-2t} u(t) \text{ V}$

$$V_{out}(s) = \frac{1}{(s+2)^2}$$

However,

$$H(s) = \frac{2s+5}{(s+2)(s+3)} = \frac{V_{out}(s)}{V_{in}(s)}$$

$$\therefore V_{in}(s) = \frac{V_{out}(s)}{H(s)}$$

or

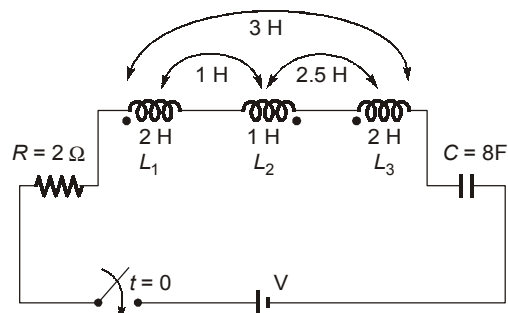
$$V_{in}(s) = \frac{1}{(s+2)^2} \times \frac{(s+2)(s+3)}{(2s+5)} = \frac{1}{2} \left[ \frac{2}{s+2} - \frac{1}{s+2.5} \right]$$

Taking inverse of Laplace

$$v_{in}(t) = \left( e^{-2t} - \frac{1}{2} e^{-2.5t} \right) u(t) \text{ volts}$$

26.(a)

For the circuit



$$L_{eq} = L_1 + L_2 + L_3 - 2M_{12} - 2M_{23} + 2M_{13}$$

$$L_1 = 2 \text{ H}$$

$$L_2 = 1 \text{ H}$$

$$L_3 = 2 \text{ H}$$

$$M_{12} = 1 \text{ H}$$

$$M_{23} = 2.5 \text{ H}$$

$$\begin{aligned}
 M_{13} &= 3 \text{ H} \\
 L_{eq} &= 2 + 1 + 2 - 2 - 5 + 6 \\
 &= 11 - 7 \\
 &= 4 \text{ H} \\
 C &= 8 \text{ F} \\
 R &= 2 \Omega
 \end{aligned}$$

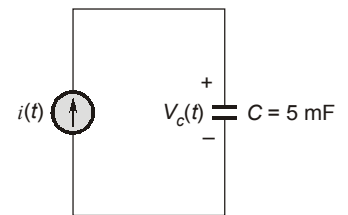
Note :  $M_{12}, M_{23}$  is negative, because both  $L_1, L_2$  and  $L_2, L_3$  opposes the flux of respective loops.

$$\xi = \frac{R}{2} \sqrt{\frac{C}{L}} = \frac{2}{2} \sqrt{\frac{8}{4}} = \sqrt{2} = 1.414$$

Thus, the given circuit is overdamped.

27.(b)

$$\begin{aligned}
 i &= C \frac{dv}{dt} \\
 V_C &= \frac{1}{C} \int_{-\infty}^t i dt
 \end{aligned}$$

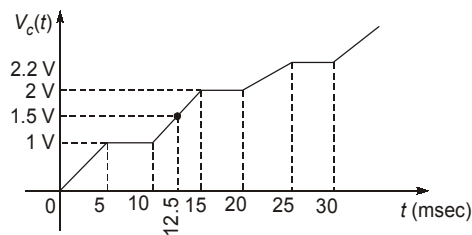


For  $0 < t < 5$  ; Unit step current is applied ; voltage will increase linearly.

For  $5 < t < 10$  ; No current is applied, hence open circuit, the capacitor will hold the charge.

For  $10 < t < 15$  ; again capacitor's voltage increases linearly.

From above analysis,



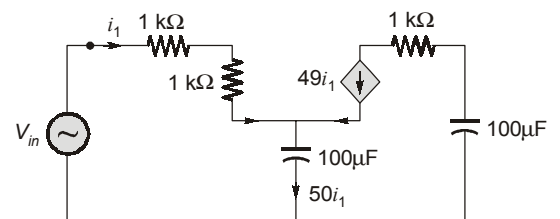
$$V_C(t)|_{12.5 \text{ msec}} = 1.5 \text{ V}$$

28.(b)

Applying KVL,

$$\begin{aligned}
 V_{in} - i_1(1 + 1) - 50 i_1(-jX_C) &= 0 \\
 \Rightarrow V_{in} &= i_1[2 - j50 X_C]
 \end{aligned}$$

$$\text{Input impedance} = \frac{V_{in}}{i_1} = 2 - j50 X_C$$

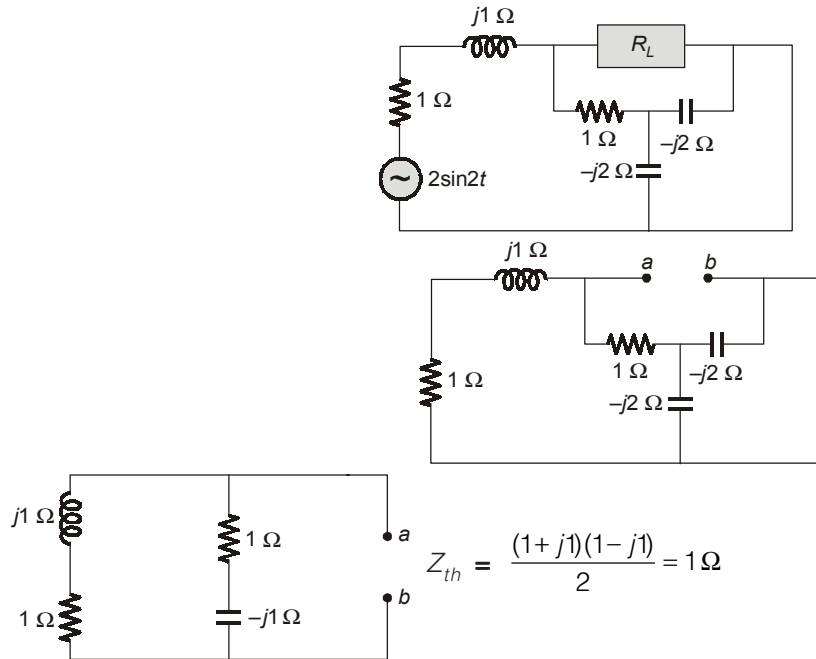


As imaginary part is negative, input impedance has equivalent capacitive reactance  $X_{Ceq}$ .

$$\begin{aligned}
 X_{Ceq} &= 50 X_C \\
 \frac{1}{\omega C_{eq}} &= \frac{50}{\omega C} \\
 C_{eq} &= \frac{C}{50} = \frac{100}{50} \mu\text{F} \\
 C_{eq} &= 2 \mu\text{F}
 \end{aligned}$$

29.(b)

Since  $(\omega) = 2$  rad/sec, the network is drawn as



$\therefore R_L = |Z_{th}| = 1 \Omega$

Hence, for maximum power to  $R_L$ , it should be  $1 \Omega$ .

30. (d)

$$v(t) = 2\cos(500t + 60^\circ) \text{ V}$$

$$= 2\angle 60^\circ \text{ V}$$

Using AC phasor

$$i(t) = -\sqrt{3}\cos(500t + 30^\circ) \text{ A}$$

$$v(t) = 2\angle 60^\circ \text{ V}$$

$$i_1(t) = -i(t) = \sqrt{3}\angle 30^\circ \text{ A}$$

$$\sqrt{3}\angle 30^\circ = \frac{2\angle 60^\circ}{1 + jX_L} + \frac{2\angle 60^\circ}{2}$$

$$\sqrt{3}\angle 30^\circ = \frac{2\left(\frac{1}{2} + j\frac{\sqrt{3}}{2}\right)}{(1 + jX_L)} + \frac{1}{2} + j\frac{\sqrt{3}}{2}$$

By equating real parts on both sides,

$$\frac{3}{2} = \frac{1}{1 + X_L^2} + \frac{1}{2} + \frac{X_L\sqrt{3}}{1 + X_L^2}$$

$$= \frac{1 + X_L\sqrt{3}}{1 + X_L^2} = \frac{3}{2} - \frac{1}{2} = 1 = 1 + X_L\sqrt{3} = 1 + X_L^2$$

$$X_L = \omega L = \sqrt{3} \Omega$$

or  $L = \frac{\sqrt{3}}{500} = 3.46 \text{ mH}$

