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SIGNALS AND SYSTEMS

EC + EE

Date of Test : 24/03/2022

ANSWER KEY >

- | | | | | |
|--------|---------|---------|---------|---------|
| 1. (d) | 7. (d) | 13. (b) | 19. (a) | 25. (b) |
| 2. (b) | 8. (b) | 14. (b) | 20. (d) | 26. (a) |
| 3. (a) | 9. (c) | 15. (b) | 21. (c) | 27. (d) |
| 4. (a) | 10. (b) | 16. (c) | 22. (d) | 28. (c) |
| 5. (a) | 11. (c) | 17. (a) | 23. (d) | 29. (b) |
| 6. (c) | 12. (a) | 18. (a) | 24. (c) | 30. (b) |

DETAILED EXPLANATIONS

1. (d)

$$2^n u[n] \xleftrightarrow{z} \frac{1}{1-2z^{-1}}$$

$$\therefore \text{ROC} \Rightarrow |z| > 2$$

and $(4)^n u[-n-1] \xleftrightarrow{z} \frac{1}{1-4z^{-1}}$

$$\therefore \text{ROC} \Rightarrow |z| < 4$$

\therefore for the signal to converge

$$\text{ROC} \Rightarrow 2 < |z| < 4$$

2. (b)

$$y[n] = x[n] * \delta[n-2] = x[n-2]$$

$$\therefore y[2] = x[0] = 1$$

3. (a)

$$\therefore j \frac{d}{dt} [F(j\omega)] \longleftrightarrow t[f(t)]$$

thus at $t = 0$ the answer will be zero.

4. (a)

$$\begin{aligned} X(e^{j\omega}) &= e^{j\omega} + e^{-j\omega} + 2(e^{2j\omega} - e^{-2j\omega}) + 3(e^{3j\omega} + e^{-3j\omega}) \\ &= 2\cos\omega + 4j\sin(2\omega) + 2\cos3\omega = 2\cos(\pi) + 4j\sin(2\pi) + 6\cos(3\pi) \\ &= -2 + 0 - 6 = -8 \\ |Xe^{j\pi}| &= 8 \end{aligned}$$

5. (a)

$$\text{Re}\{x(t)\} = \frac{x(t) + x^*(t)}{2}$$

\therefore The Fourier coefficient of $x^*(t)$ are

$$b_K = \frac{1}{T} \int_T x^*(t) e^{-jk\frac{2\pi}{T}t} dt$$

Taking conjugate on both sides

$$b_K^* = \frac{1}{T} \int_T x(t) e^{-j(-K)\frac{2\pi}{T}t} dt$$

$$\therefore a_{-K} = b_K^*$$

$$\therefore \text{Fourier series Coefficient of } \text{Re}\{x(t)\} = \frac{a_K + a_{-K}^*}{2}$$

6. (c)

From the given figure of $x(t)$ and $y(t)$, we get conclude that,

$$x(0) = y(5)$$

at $t = 5$ sec, $y(5) = x(-5a + 20) = x(0)$

$$\therefore -5a + 20 = 0$$

$$\therefore a = 4$$

7. (d)

Given, signal $x(t)$ has energy ' E '

for $ax(t) \xrightarrow{E} a^2 E$

$$\therefore ax(bt + c) \xrightarrow{E} \frac{a^2 E}{b}$$

From the given signal $a = 2$, $b = 5$ and hence

\therefore The energy of signal $2x(5t - 6)$ is

$$E = \frac{(2)^2 \times 10}{5} = 8 \text{ J}$$

8. (b)

Given, $x(t) = \frac{\sin(10\pi t)}{\pi t}$

Taking Fourier transform

$$X(j\omega) = \begin{cases} 1 & ; \quad |\omega| \leq 10\pi \\ 0 & ; \quad |\omega| > 10\pi \end{cases}$$

or

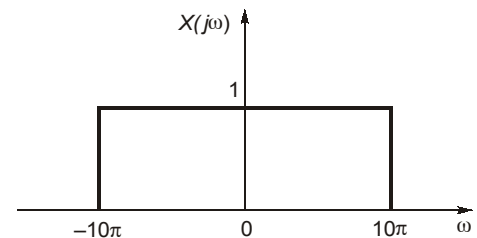
\therefore The maximum frequency ' ω_m ' present in $x(t)$ is $\omega_m = 10\pi$

Hence we require,

$$\frac{2\pi}{T_s} > 2\omega_m$$

$$\frac{2\pi}{T_s} > 20\pi$$

$$\therefore T_s < \frac{1}{10}$$



9. (c)

Given, $H(z) = \frac{z}{z - 0.2} = \frac{1}{1 - 0.2z^{-1}}$ ROC: $|z| > 0.2$

Since the ROC: $|z| > 0.2$, which includes unit circle.

\therefore The impulse response will be stable.

10. (b)

For sequence $X_1[n] \xrightarrow{z} X_1[z]$; ROC = R

For sequence $X_2[n] = X_1[-n] \xrightarrow{z} X_1[1/z]$; ROC = $1/R$

\therefore ROC's are reciprocal of each other.

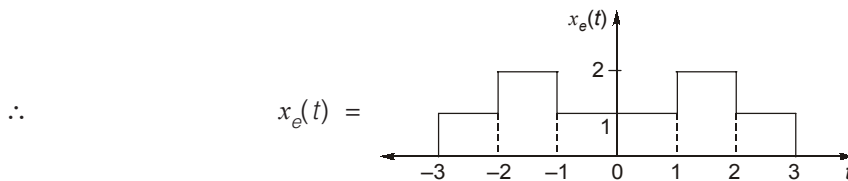
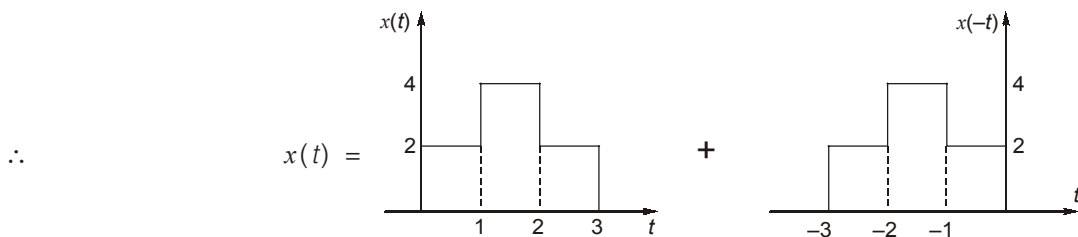
11. (c)

$$x(t) \xrightarrow{L.T} X(s) = \frac{s+3}{(s+3)^2+4}$$

$$y(t) \Rightarrow \int x(\tau) d\tau \Rightarrow \frac{s+3}{s[(s+3)^2+4]}$$

12. (a)

$$x_e(t) = \frac{x(t) + x(-t)}{2}$$



13. (b)

$$Y(e^{j\omega}) = X(e^{j\omega}) * X(e^{j\omega - \pi/2})$$

∴ In time domain

$$Y(e^{j\omega}) = 2\pi x_1[n] \cdot x_2[n]$$

now, If

$$X(e^{j\omega}) \longleftrightarrow n \left(\frac{3}{4}\right)^{|n|}$$

Then

$$X(e^{j\omega - \pi/2}) \longleftrightarrow n \cdot e^{j\pi/2} \left(\frac{3}{4}\right)^{|n|}$$

∴

$$y[n] \longleftrightarrow 2\pi n^2 e^{j\pi n/2} \left(\frac{3}{4}\right)^{2|n|}$$

14. (b)

$$X(z) = \frac{z}{z-1} \quad |z| > 1$$

$$Y(z) = \frac{2z}{z - \frac{1}{3}} \quad |z| > \frac{1}{3}$$

$$H(z) = \frac{2(z-1)}{z-\frac{1}{3}} \quad |z| > \frac{1}{3}$$

$$X'(z) = \frac{z}{z-\frac{1}{2}} \quad |z| > \frac{1}{2}$$

$$Y(z) = H(z) \cdot X'(z)$$

$$= \frac{2z(z-1)}{\left(z-\frac{1}{2}\right)\left(z-\frac{1}{3}\right)} \quad |z| > \frac{1}{2}$$

Taking inverse z transform

$$y[n] = \left[-6\left(\frac{1}{2}\right)^n + 8\left(\frac{1}{3}\right)^n \right] u[n]$$

$$k_1 = -6, \quad k_2 = 8$$

so,

$$k_1 + k_2 = 2$$

15. (b)

If $x[n]$ is real

$$\text{odd}[x[n]] \xrightarrow{FT} j\text{Im}[X(e^{j\omega})]$$

$$\therefore \text{odd}[x[n]] = F^{-1} \left[\frac{1}{2} (e^{j\omega} - e^{-j\omega} - e^{2j\omega} + e^{-2j\omega}) \right]$$

$$= \frac{1}{2} [\delta[n+1] - \delta[n-1] - \delta[n+2] + \delta[n-2]]$$

$$\therefore \text{odd}[x[n]] = \frac{x[n] - x[-n]}{2}$$

Since,

$$x[n] = 0 \text{ for } n > 0$$

$$x[n] = 2 \text{ odd}[x[n]]$$

$$= \delta[n+1] - \delta[n+2] \text{ for } n < 0$$

using Parseval's theorem

$$\frac{1}{2\pi} \int_{-\infty}^{\infty} |X(e^{j\omega})|^2 d\omega = \sum_{n=-\infty}^{\infty} |x[n]|^2$$

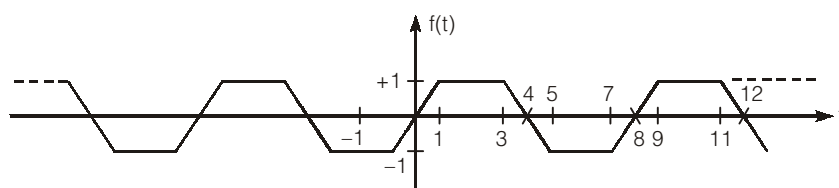
$$(x[0])^2 - 2 = 3$$

$$x[0] = \pm 1$$

$$\therefore x[0] > 0$$

$$\therefore x[n] = \delta[n] + \delta[n+1] - \delta[n+2]$$

16. (c)



The signal $f(t)$ has hidden, odd and half-wave symmetry.

$$\begin{aligned} \text{So,} \quad a_0 &\neq 0 \\ a_n &= 0; \quad \forall n \\ b_n &\neq 0; \quad n = 1, 3, 5 \end{aligned}$$

Therefore, non zero Fourier series coefficients are a_0 and $b_n, n = 1, 3, 5, \dots$

17. (a)

$$\begin{aligned} \text{As,} \quad x(t) \cos^2 t &= x(t) \left[\frac{1}{2} + \frac{1}{2} \cos 2t \right] \\ &= \frac{1}{2} x(t) + \frac{1}{2} \cos 2t \end{aligned}$$

$$\begin{aligned} \text{Now,} \quad g(t) &= x(t) \cdot \cos^2 t * \frac{\sin t}{\pi t} \\ &= x(t) \left[\frac{1 + \cos 2t}{2} \right] * \frac{\sin t}{\pi t} \end{aligned}$$

$$\therefore G(j\omega) = \left[\frac{1}{2} X(j\omega) + \frac{1}{4} X(\omega - 2) + \frac{1}{4} X(\omega + 2) \right] \times \text{rect} \left(\frac{\omega}{2} \right)$$

Thus the given solution will be

$$\therefore G(j\omega) = \frac{1}{2} X(j\omega)$$

$$\text{or} \quad g(t) = \frac{1}{2} x(t)$$

Thus to get the desired result

$$h(t) = \frac{1}{2} \delta(t)$$

18. (a)

$$\begin{aligned} x[n] &= \delta[n] \\ X(e^{j\omega}) &= 1 \end{aligned}$$

$$\frac{dX(e^{j\omega})}{d\omega} = 0$$

$$\therefore Y(e^{j\omega}) = e^{-j\omega} X(e^{j\omega})$$

$$h[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} e^{-j\omega} \cdot e^{j\omega n} d\omega = \frac{\sin \pi(n-1)}{\pi(n-1)}$$

19. (a)

$$\therefore x^*(t) \xleftrightarrow{F} X^*(-j\omega)$$

$$\text{and} \quad \int_{-\infty}^t x(\tau) d\tau \xleftrightarrow{F} \frac{X(j\omega)}{j\omega} + \pi X(0) \delta(\omega)$$

20. (d)

$$H(j\omega) = \frac{1 + 2e^{-j\omega}}{1 + \frac{1}{2e^{-j\omega}}} = \frac{1 + 2e^{-j\omega}}{2e^{-j\omega} + 1} \cdot 2e^{-j\omega}$$

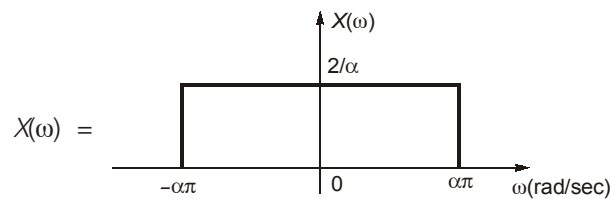
$$|H(j\omega)| = 2$$

21. (c)

We know that,

Energy,
$$E = \int_{-\infty}^{\infty} x^2(t) dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} |X(\omega)|^2 d\omega$$

Now,



$$E = \frac{1}{2\pi} \int_{-\alpha\pi}^{\alpha\pi} \left| \frac{2}{\alpha} \right|^2 d\omega$$

$$0.5 = \frac{1}{2\pi} \times \frac{4}{\alpha^2} [2\alpha\pi]$$

$$0.5 = \frac{4}{\alpha}$$

$$\therefore \alpha = 8$$

22. (d)

Given,
$$x[n] = \cos\left(\frac{\pi}{4}n\right) + \sin\left(\frac{\pi}{3}n + \frac{1}{2}\right)$$

Let
$$x[n] = x_1[n] + x_2[n]$$

Let N_1 be the period of $x_1[n]$.

$$\frac{\pi/4}{2\pi} = \frac{m}{N_1}$$

$$\Rightarrow N_1 = 8$$

Let N_2 be the period of $x_2[n]$.

$$\frac{\pi/3}{2\pi} = \frac{m}{N_2}$$

$$\Rightarrow N_2 = 6$$

Overall time period, $N = \text{LCM}(N_1, N_2) = \text{LCM}(6, 8)$

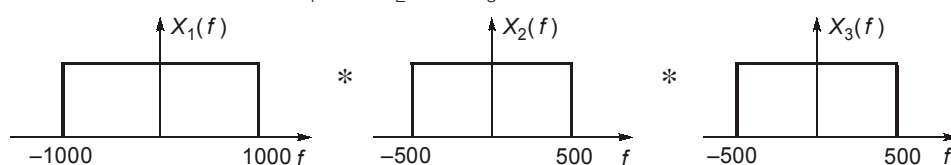
$$\therefore N = 24$$

23. (d)

We know that,

For

$$X(f) = X_1(f) * X_2(f) * X_3(f)$$



Sampling frequency, $f_s = 2(1000 + 500 + 500)$
 $f_s = 4000$ samples/sec

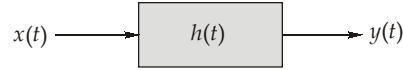
24. (c)

Given, the Causal LTI system,

$$H(j\omega) = \frac{1}{3 + j\omega}$$

and output,

$$y(t) = e^{-3t} u(t) - e^{-4t} u(t)$$



We know that,

$$H(j\omega) = \frac{Y(j\omega)}{X(j\omega)}$$

$$Y(j\omega) = \frac{1}{3 + j\omega} - \frac{1}{4 + j\omega} = \frac{1}{(3 + j\omega)(4 + j\omega)}$$

∴

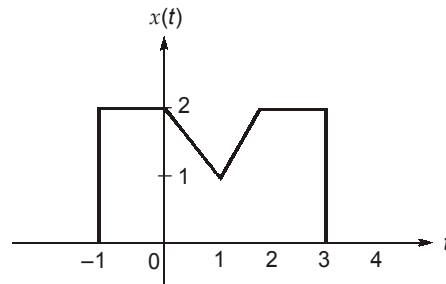
$$X(j\omega) = \frac{Y(j\omega)}{H(j\omega)} = \frac{1}{4 + j\omega}$$

By inverse Fourier transform of $X(j\omega)$, we have,

$$x(t) = e^{-4t} u(t)$$

25. (b)

Given,



By the definition of Fourier transform,

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) \cdot e^{-j\omega t} dt$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) \cdot e^{j\omega t} d\omega$$

at $t = 0$,

$$x(0) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) d\omega$$

$$\therefore \int_{-\infty}^{\infty} X(j\omega) d\omega = 2\pi x(0) = 2\pi(2) = 4\pi \approx 12.57$$

26. (a)

$$X(z) = \sum_{n=-\infty}^{\infty} x(n)z^{-n} = \sum_{n=0}^{\infty} \frac{2^{-n}}{n!} z^{-n}$$

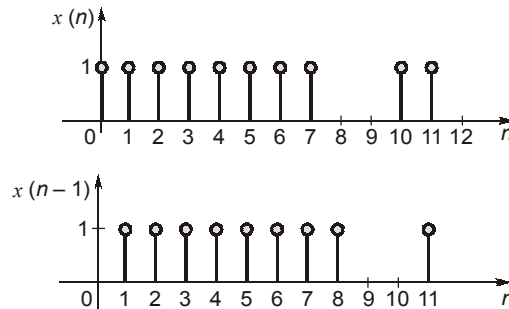
$$= \sum_{n=0}^{\infty} \frac{(2z)^{-n}}{n!} = \sum_{n=0}^{\infty} \frac{\left(\frac{1}{2z}\right)^n}{n!}$$

$$X(z) = 1 + \frac{1}{2z} + \frac{\left(\frac{1}{2z}\right)^2}{2!} + \frac{\left(\frac{1}{2z}\right)^3}{3!} + \dots$$

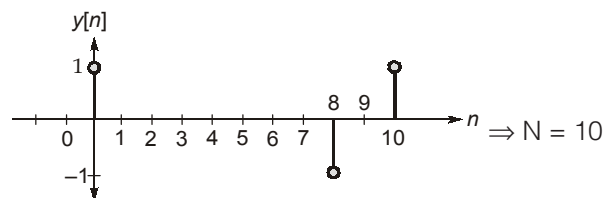
$$X(z) = e^{1/2z}$$

$$X(1) = e^{1/2} = \sqrt{e} = 1.648 \approx 1.65$$

27. (d)



Subtracting the two signal, we get



28. (c)

We know that the Laplace transform of

$$\cos(at)u(t) = \frac{s}{s^2 + a^2}$$

$$\therefore \cos(\pi t)u(t) = \frac{s}{s^2 + \pi^2}$$

now, the given function $x(t)$ can be written as,

$$\begin{aligned} &= \cos\pi t[u(t) - u(t-1)] \\ &= \cos(\pi t)u(t) - \cos\pi t u(t-1) \\ &= \cos\pi t u(t) - \cos\pi(t-1+1)u(t-1) \\ &= \cos\pi t u(t) - \cos[\pi(t-1) + \pi] u(t-1) \\ x(t) &= \cos(\pi t)u(t) + \cos[\pi(t-1)] u(t-1) \end{aligned}$$

By taking Laplace transform,

$$X(s) = \frac{s}{s^2 + \pi^2} + \frac{se^{-s}}{s^2 + \pi^2} \quad [\because x(t-t_0) = X(s) \cdot e^{-st_0}, \text{ by shifting property}]$$

$$X(s) = \frac{s[1 + e^{-s}]}{s^2 + \pi^2}$$

29. (b)

Given signals, $x(t) = \sin\omega_0 t$

$$h(t) = \text{sgnt}$$

from the multiplication property of Fourier transform,

$$x(t)h(t) = \frac{1}{2\pi} [X(\omega) * H(\omega)]$$

Fourier transform of $x(t)$ is $X(\omega)$,

$$X(\omega) = \frac{\pi}{j} [\delta(\omega - \omega_0) - \delta(\omega + \omega_0)]$$

Fourier transform of $h(t)$ is $H(\omega)$,

$$H(\omega) = \frac{2}{j\omega}$$

$$\begin{aligned} \therefore x(t)h(t) &\xrightarrow{\text{FT}} \frac{1}{2\pi} \left[\frac{\pi}{j} (\delta(\omega - \omega_0) - \delta(\omega + \omega_0)) * \frac{2}{j\omega} \right] \\ &\xrightarrow{\text{FT}} \frac{1}{2\pi} \left[\left[\frac{\pi}{j} \times \frac{2}{j(\omega - \omega_0)} \right] - \left[\frac{\pi}{j} \times \frac{2}{j(\omega + \omega_0)} \right] \right] \\ &\quad (\because X(\omega) * \delta(\omega - \omega_0) = X(\omega - \omega_0)) \\ &\xrightarrow{\text{FT}} \left[\frac{-1}{\omega - \omega_0} + \frac{1}{\omega + \omega_0} \right] = \frac{-\omega - \omega_0 + \omega - \omega_0}{\omega^2 - \omega_0^2} \\ \therefore x(t)h(t) &\xrightarrow{\text{FT}} \frac{-2\omega_0}{\omega^2 - \omega_0^2} \end{aligned}$$

30. (b)

Given,
$$X(z) = \frac{1}{1 - 2.5z^{-1} + z^{-2}} = \frac{1}{(z^{-1} - 2)\left(z^{-1} - \frac{1}{2}\right)}$$

$$\frac{1}{(z^{-1} - 2)\left(z^{-1} - \frac{1}{2}\right)} = \frac{A}{z^{-1} - 2} + \frac{B}{z^{-1} - \frac{1}{2}}$$

$$\therefore A = \frac{1}{2 - \frac{1}{2}} = \frac{1}{3/2} = \frac{2}{3}$$

$$B = \frac{1}{\frac{1}{2} - 2} = -\frac{2}{3}$$

$$\begin{aligned} \therefore X(z) &= \frac{\frac{2}{3}}{z^{-1} - 2} + \frac{-\frac{2}{3}}{z^{-1} - \frac{1}{2}} \\ &= \frac{-\frac{1}{3}}{1 - \frac{1}{2}z^{-1}} + \frac{\frac{4}{3}}{1 - 2z^{-1}} \end{aligned}$$

Given $X(z)$ is a causal system, the ROC is right of the right most pole.

$$\therefore |z| > 2$$

hence,
$$x[n] = -\frac{1}{3} \left(\frac{1}{2}\right)^n u[n] + \frac{4}{3} (2)^n u[n]$$

$$\therefore x(0) = -\frac{1}{3} + \frac{4}{3} = \frac{3}{3} = 1$$

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