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# NETWORK THEORY

EC-EE

Date of Test : 04/04/2026

## ANSWER KEY >

- |        |         |         |         |         |
|--------|---------|---------|---------|---------|
| 1. (c) | 7. (d)  | 13. (b) | 19. (c) | 25. (a) |
| 2. (b) | 8. (c)  | 14. (c) | 20. (b) | 26. (a) |
| 3. (a) | 9. (b)  | 15. (b) | 21. (a) | 27. (b) |
| 4. (a) | 10. (b) | 16. (c) | 22. (d) | 28. (c) |
| 5. (d) | 11. (b) | 17. (a) | 23. (a) | 29. (a) |
| 6. (b) | 12. (a) | 18. (a) | 24. (c) | 30. (a) |

## DETAILED EXPLANATIONS

1. (c)

$$R_L = |Z_s^*| = |5 - j10| = \sqrt{5^2 + 10^2} = 11.18 \Omega$$

2. (b)

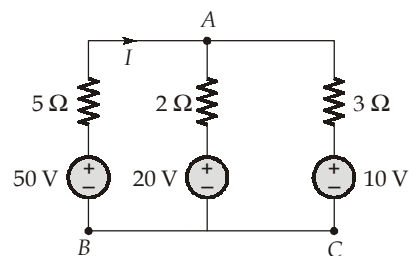
At  $t = \infty$ , capacitor is open circuited

$$V = 2 \times \frac{3}{2} = 3 \text{ V}$$

3. (a)

$$\begin{aligned} V_5 &= 20 \times 5 = 100 \text{ V} \\ \text{Voltage across } 10 \Omega &= 20 \text{ V} = V_{10} \\ V_{\text{Th}} &= V_5 - V_{10} \\ &= 100 - 20 = 80 \text{ V} \end{aligned}$$

4. (a)

Using nodal analysis,  
KCL at node 'A'

$$\frac{50 - V_A}{5} = \frac{V_A - 20}{2} + \frac{V_A - 10}{3}$$

$$V_A = 22.580 \text{ V}$$

So,

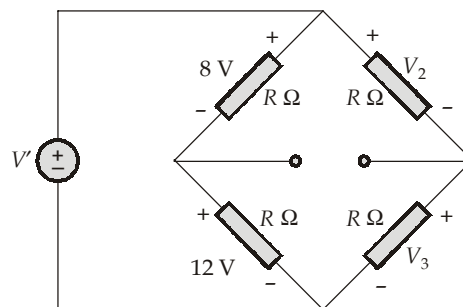
$$I = \frac{50 - 22.580}{5} = 5.483 \text{ A}$$

So, power delivered by 50 V is,

$$50 \times I = 50 \times 5.483 = 274.2 \text{ Watts}$$

5. (d)

Under bridge balance condition, (since current through 6 Ω resistor is zero).



$$V' = 8 + 12 = V_2 + V_3$$

 $\Rightarrow$ 

$$V_2 + V_3 = 20 \text{ V}$$

Option (d) satisfies the above condition.

6. (b)

The reactive power in the circuit is

$$Q \propto \sin \theta$$

If  $Q$  is positive then angle of impedance ( $\theta$ ) is positive which implies that current phasor is lagging voltage phasor i.e., load is inductive.

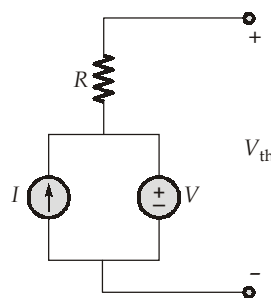
$$Z = \frac{V \angle \theta_V}{I \angle \theta_I} = \frac{V}{I} \angle \theta_V - \theta_I$$

$$\theta_V > \theta_I$$

Hence, an inductive load has lagging power factor.

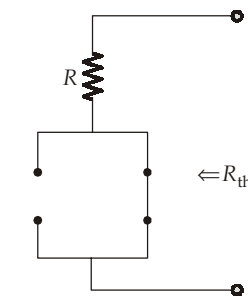
7. (d)

Given, circuit,



The open circuit voltage (Thevenin voltage  $V_{th}$ ) is equal to  $V$ .

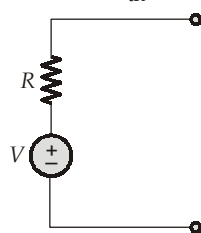
For Thevenin's resistance :  $R_{th}$  by setting all independent sources to zero, i.e., open circuit the current source and short circuit the voltage source.



∴

$$R_{th} = R$$

The equivalent circuit is



8. (c)

In series RLC circuit at resonance,

$$\text{Current, } I_R = \frac{V_s}{R}$$

Voltage across inductor is,  $V_L = j\omega_0 L I_R = j\omega_0 L \frac{V_s}{R}$

$$V_L = jQ V_s$$

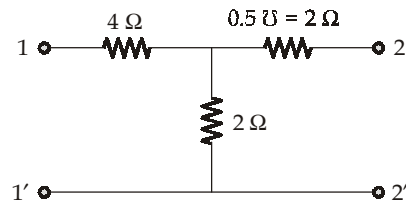
where,

$$Q = \frac{\omega_0 L}{R}$$

Since,  $Q > 1 \Rightarrow V_L > V_S$

9. (b)

Given, two port network

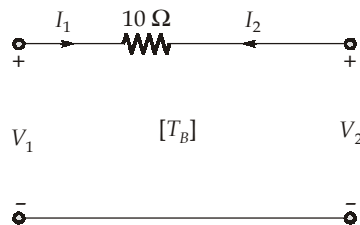


$\therefore$

$$[z] = \begin{bmatrix} 6 & 2 \\ 2 & 4 \end{bmatrix} \Omega$$

10. (b)

When  $2 \Omega$  is replaced by  $10 \Omega$ ,



From the above figure, we get,

$$V_1 - V_2 = 10I_1 \quad \dots(i)$$

and

$$V_2 - V_1 = 10I_2 \quad \dots(ii)$$

From  $T$  parameter model,

$$V_1 = AV_2 - BI_2$$

and

$$I_1 = CV_2 - DI_2$$

from equation (ii), we get,

$$V_1 = V_2 - 10I_2 \quad \Rightarrow A = 1 \text{ and } B = 10$$

from equation (i), we get,

$$\begin{aligned} V_2 - 10I_2 - V_2 &= 10I_1 \\ -I_2 &= I_1 \quad \Rightarrow C = 0 \text{ and } D = 1 \end{aligned}$$

$\therefore$

$$T_B = \begin{bmatrix} 1 & 10 \\ 0 & 1 \end{bmatrix} = \begin{bmatrix} 1 & 2 \\ 0 & 1 \end{bmatrix}^5 = [T_A]^5$$

11. (b)

$$\begin{aligned} Z_{AB} &= \left( \frac{23}{6} \right) + [(3+j4) \parallel (3-j4)] \\ &= \frac{23}{6} + \frac{(3+j4)(3-j4)}{6} = \frac{23+25}{6} = \frac{48}{6} \Omega = 8 \Omega \end{aligned}$$

$\therefore$

$$Z_{AB} = 8 \Omega$$

12. (a)

$$L_{eq} = L_1 + L_2 - 2M = 4 + 4 - 2 \times 2 = 4 \text{ mH}$$

Resonant frequency,

$$f_o = \frac{1}{2\pi\sqrt{L_{eq}C}}$$

$$f_o = \frac{1}{2\pi\sqrt{4 \times 0.1 \times 10^{-9}}} = 7.96 \text{ kHz}$$

13. (b)

$$\text{Impedance matrix for 'N'} = \frac{1}{[Y]} = \frac{1}{5} \begin{bmatrix} 3 & -2 \\ -2 & 3 \end{bmatrix} = \begin{bmatrix} 3/5 & -2/5 \\ -2/5 & 3/5 \end{bmatrix}$$

In series connection: individual impedance parameters are added

$$\therefore \text{ For individual network} = \frac{1}{2} \begin{bmatrix} 3/5 & -2/5 \\ -2/5 & 3/5 \end{bmatrix} = \begin{bmatrix} 0.3 & -0.2 \\ -0.2 & 0.3 \end{bmatrix}$$

14. (c)

$$v(t) = 2[u(t) - u(t-2)] \text{ V}$$

$$i(t) = [r(t) - r(t-2)] \text{ A}$$

$$v(t) = 2 \frac{di(t)}{dt}$$

$$\text{For inductor, } v(t) = L \frac{di(t)}{dt}$$

$\therefore$  The element is inductor of 2 H

15. (b)

From the given data:

$$V = 20\sin(2000t + 40^\circ)$$

$$i = 5\sin(2000t + 10^\circ + 90^\circ) = 5\sin(2000t + 100^\circ)$$

The current leads the voltage by  $60^\circ$ ,

$$\text{Now, } \tan\theta = \frac{1}{\omega CR}$$

$$\tan 60^\circ = \frac{1}{\omega CR}$$

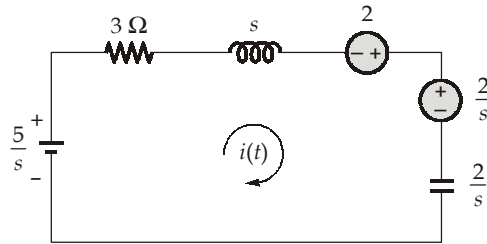
$$\frac{1}{\omega C} = \sqrt{3}R$$

$$\text{Since, } \frac{V_m}{I_m} = \frac{20}{5} = 4 = \sqrt{R^2 + \left(\frac{1}{\omega C}\right)^2}$$

$$\sqrt{R^2 + 3R^2} = 4 \Rightarrow 2R = 4$$

$$\therefore R = 2 \Omega$$

16. (c)

At  $t = 0$ , switch is closedFor  $t > 0$ , the circuit in  $s$ -domain becomes,

Applying KVL, we get,

$$\frac{5}{s} - \frac{2}{s} + 2 = \left(3 + s + \frac{2}{s}\right) I(s)$$

$$I(s) = \frac{2s + 3}{(s + 1)(s + 2)}$$

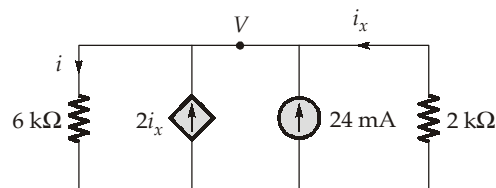
Using partial fractions,

$$I(s) = \frac{1}{(s + 1)} + \frac{1}{(s + 2)}$$

or

$$i(t) = L^{-1}[I(s)] = (e^{-t} + e^{-2t}) \text{ A ; for } t > 0$$

17. (a)



Applying KCL,

$$i = 2i_x + 24 \text{ mA} + i_x \quad \dots(i)$$

where,

$$i = \frac{V}{6000} \text{ and } i_x = \frac{-V}{2000} \quad \dots(ii)$$

Therefore, from equations (i) and (ii)

$$\frac{V}{6000} + \frac{V}{2000} - 2\left(-\frac{V}{2000}\right) = 24 \text{ mA}$$

$$\Rightarrow V = (600) (24 \times 10^{-3}) = 14.4 \text{ V}$$

Hence, power supplied by independent current source

$$P = V \times 24 \text{ mA} = 14.4 \times 24 \times 10^{-3} = 345.6 \text{ mW}$$

18. (a)

$$\frac{1}{2} v_p I_p \cos \theta = 200$$

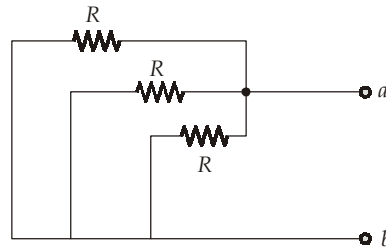
$$\Rightarrow v_p I_p = \frac{400}{0.6}$$

$$\cos \theta = 0.6 \Rightarrow \sin \theta = 0.8$$

$$\begin{aligned} \text{Now reactive power} &= \frac{1}{2} v_p I_p \sin \theta = \frac{1}{2} \times \frac{400}{0.6} \times 0.8 \\ &= 266.67 \text{ VAR} \end{aligned}$$

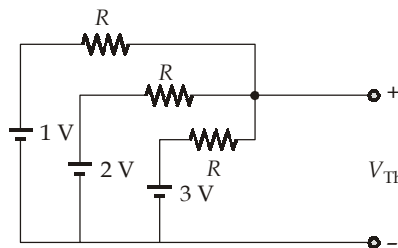
19. (c)

For finding  $R_{Th}$  across terminal  $a$  and  $b$ :



$$R_{Th} = \frac{R}{3} \Omega$$

For finding  $V_{Th}$ :



Using KCL, we get,

$$\frac{V_{Th} - 1}{R} + \frac{V_{Th} - 2}{R} + \frac{V_{Th} - 3}{R} = 0$$

$$3V_{Th} = 6$$

$$V_{Th} = 2 \text{ V}$$

∴ Maximum power transferred will be given by

$$P_{max} = \frac{V_{Th}^2}{4R_{Th}}$$

$$5 \times 10^{-3} = \frac{2 \times 2}{4 \times \frac{R}{3}}$$

$$\frac{R}{3} = \frac{10^3}{5}$$

or

$$R = \frac{3}{5} \times 10^3 = 600 \Omega$$

20. (b)

The total admittance of the network

$$\begin{aligned} Y_{eq} &= j\omega C + \frac{1}{R_1} + \frac{1}{R_2 + j\omega L} = j\omega(0.5) + \frac{1}{5} + \frac{1}{1 + j\omega(1)} \\ &= j0.5\omega + 0.2 + \frac{1}{1 + j\omega} \times \frac{(1 - j\omega)}{(1 - j\omega)} = j0.5\omega + 0.2 + \frac{1 - j\omega}{1 + \omega^2} \\ &= \left\{ 0.2 + \frac{1}{1 + \omega^2} \right\} + j \left\{ 0.5\omega - \frac{\omega}{1 + \omega^2} \right\} \end{aligned}$$

For unity power factor imaginary part of admittance is zero.

$$\begin{aligned} \operatorname{Im}\{Y_{\text{eq}}\} &= 0 \\ 0.5\omega_0 - \frac{\omega_0}{1 + \omega_0^2} &= 0 \\ \frac{1}{2}\omega_0 &= \frac{\omega_0}{1 + \omega_0^2} \\ 1 + \omega_0^2 &= 2 \\ \omega_0 &= 1 \text{ rad/sec} \end{aligned}$$

21. (a)

For  $t < 0$ , source  $2u(t) = 0$

Therefore,

$$i_L(0^-) = i_L(0^+) = 0 \text{ A}$$

$$v_c(0^-) = v_c(0^+) = 0 \text{ V}$$

For  $t > 0$ ,

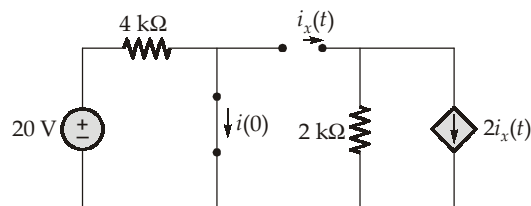
$$i_c(0^+) = 2 \text{ mA}$$

$$i_c(0^+) = C \frac{dv_c(0^+)}{dt}$$

$$\frac{dv_c(0^+)}{dt} = \frac{i_c(0^+)}{C} = \frac{2 \times 10^{-3}}{4 \times 10^{-3}} = 0.5 \text{ V/sec}$$

22. (d)

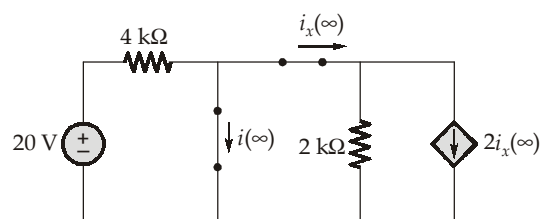
For  $t < 0$ : The switch was open, the circuit is said to be in steady state.



Since, the inductor acts as short circuit in steady state, the current through it is

$$i(0) = \frac{20}{4 \text{ k}\Omega} = 5 \text{ mA}$$

For  $t > 0$ ; at  $t = \infty$ , the circuit again reaches steady state.



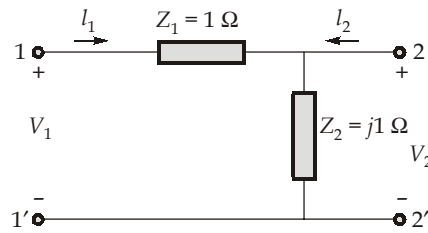
$$\therefore i_x(\infty) = 0, i(\infty) = \frac{20}{4 \text{ k}\Omega} = 5 \text{ mA}$$

$$\therefore i(t) = i(\infty) + [i(0) - i(\infty)]e^{-t/\tau}, t > 0$$

$$= 5 \text{ mA} + [(5 \text{ mA} - 5 \text{ mA})]e^{-t/\tau}; t > 0$$

$$\therefore i(t) = 5 \text{ mA}, t > 0 \text{ which is a constant current.}$$

23. (a)  
 Given, two port circuit is,



We know that, h-parameters for any two port circuit is defined as

$$V_1 = h_{11}I_1 + h_{12}V_2$$

$$I_2 = h_{21}I_1 + h_{22}V_2$$

$$\therefore h_{11} = \left. \frac{V_1}{I_1} \right|_{V_2=0} \quad h_{12} = \left. \frac{V_1}{V_2} \right|_{I_1=0}$$

$$h_{21} = \left. \frac{I_2}{I_1} \right|_{V_2=0} \quad h_{22} = \left. \frac{I_2}{V_2} \right|_{I_1=0}$$

$$\therefore h_{11} = z_1 = 1 \Omega \quad h_{21} = \frac{I_2}{I_1} = -1$$

$$h_{12} = \frac{V_1}{V_2} = 1 \quad h_{22} = -j1 \text{ } \Omega$$

$$[h] = \begin{bmatrix} 1 & 1 \\ -1 & -j1 \end{bmatrix}$$

24. (c)  
 Given,

$$\omega_o = 1000 \text{ rad/s}$$

$$\omega_o = \frac{1}{\sqrt{LC}}$$

Resonant frequency,

$$\omega_o^2 = \frac{1}{LC}$$

$\Rightarrow$

$$L = \frac{1}{\omega_o^2 \times C}$$

$\therefore$

$$L = \frac{1}{10^6 \times 0.2 \times 10^{-6}} = \frac{1}{0.2} = 5 \text{ H}$$

For parallel RLC circuit, Q-factor,

$$Q = \omega_o RC$$

$\Rightarrow$

$$R = \frac{Q}{\omega_o C}$$

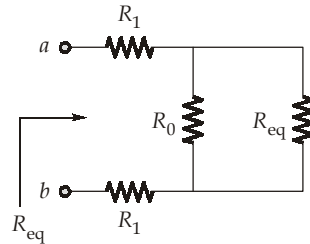
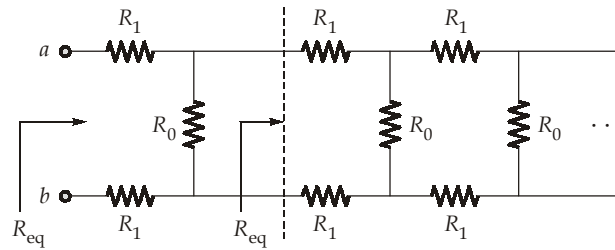
$\therefore$

$$R = \frac{80}{10^3 \times 0.2 \times 10^{-6}} = \frac{80}{0.2 \times 10^{-3}} = 400 \text{ k}\Omega$$

$\therefore$

$$\frac{L}{R} = \frac{5}{400 \times 10^3} = 12.5 \times 10^{-6} \text{ s}^{-1}$$

25. (a)  
Given, infinite resistive network,



∴

$$R_{eq} = R_1 + R_0 \parallel R_{eq} + R_1$$

$$R_{eq} = 2R_1 + R_0 \parallel R_{eq} = 2R_1 + \frac{R_0 \times R_{eq}}{R_0 + R_{eq}}$$

$$R_{eq}(R_{eq} + R_0) = 2R_1(R_0 + R_{eq}) + R_0 R_{eq}$$

$$R_0 R_{eq} + R_{eq}^2 = 2R_1 R_0 + 2R_1 R_{eq} + R_0 R_{eq}$$

∴

$$R_{eq}^2 - 2R_{eq}R_1 - 2R_1R_0 = 0$$

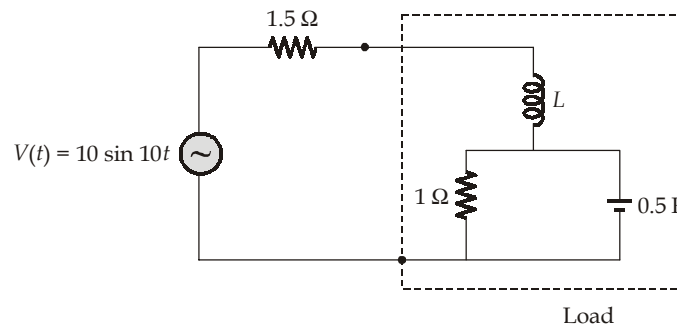
∴

$$R_{eq} = \frac{2R_1 \pm \sqrt{4R_1^2 + 8R_1R_0}}{2}$$

$$= R_1 + \sqrt{R_1^2 + 2R_1R_0}$$

$$(R_{eq} > 0)$$

26. (a)



The maximum power is transferred at the frequency at which the load is resistive and it is equal to 1.5 Ω i.e., the load is resistive means the imaginary part of the load is equal to zero.

$$Z_{load} = \frac{1 \times \frac{2}{s}}{1 + \frac{2}{s}} + Ls = \frac{2}{s+2} + Ls$$

$$= \frac{2(s-2)}{s^2-4} + Ls$$

Put  $s = j\omega$

$$Z_{\text{load}} = \frac{2(j\omega - 2)}{-\omega^2 - 4} + j\omega L = \frac{2(j10 - 2)}{-104} + j10L$$

$$Z_{\text{load}} = \frac{4}{104} + j\left(10L - \frac{20}{104}\right)$$

equating imaginary part to zero.

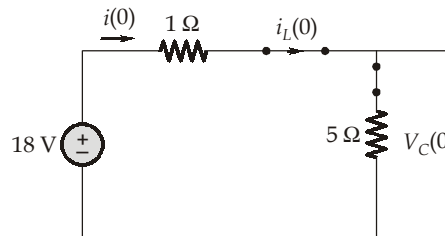
$$10L = \frac{20}{104}$$

$$\therefore L = \frac{20}{10 \times 104} = 19.23 \text{ mH}$$

27. (b)

At  $t = 0$ , the switch is closed, hence the circuit is in steady state.

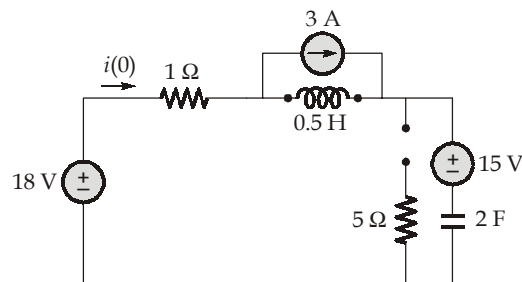
The inductor and capacitor are in steady state i.e., they are replaced as short circuit and open circuit.



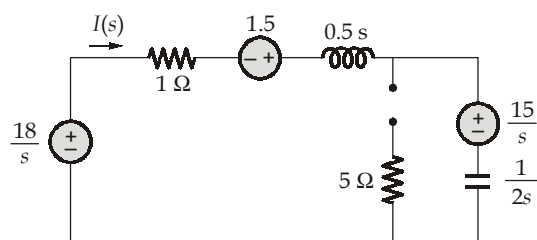
$$\therefore i_L(0) = i(0) = \frac{18}{6} = 3 \text{ A}$$

$$V_c(0) = \frac{18 \times 5}{6} = 15 \text{ V}$$

For  $t > 0$ ; the switch is in open state,



by taking Laplace transform,



$$\therefore I(s) = \frac{\frac{18}{s} - \frac{15}{s} + 1.5}{1 + 0.5s + \frac{1}{2s}} = \frac{2s \left[ \frac{3}{s} + 1.5 \right]}{2s + s^2 + 1}$$

$$\therefore I(s) = \frac{6 + 3s}{s^2 + 2s + 1}$$

$$I(s) = \frac{6}{(s+1)^2} + \frac{3s}{(s+1)^2}$$

$$= \frac{6}{(s+1)^2} + \frac{3s+3-3}{(s+1)^2} = \frac{6}{(s+1)^2} + \frac{3(s+1)}{(s+1)^2} - \frac{3}{(s+1)^2}$$

$$I(s) = \frac{3}{(s+1)^2} + \frac{3}{s+1}$$

$$\therefore i(t) = 3te^{-t}u(t) + 3e^{-t}u(t); t > 0$$

$$= (3 + 3t)e^{-t}u(t); t > 0$$

at  $t = 1$  sec

$$i(1) = (3 + 3)e^{-1}u(1) = 2.20 \text{ A}$$

28. (c)

From the given circuit,

Current,  $i = \frac{V_s}{Z}$  where  $V_s$  : Supply voltage

$$\therefore i = \frac{V_s}{1 + j\omega L + \frac{1}{j\omega 10^{-5}}} = \frac{V_s}{1 + j10^3 L + \frac{1}{j10^3 \times 10^{-5}}}$$

$$= \frac{V_s}{1 + j10^3 L - j10^2}$$

$$i = \frac{V_s}{1 + j(10^3 L - 10^2)}$$

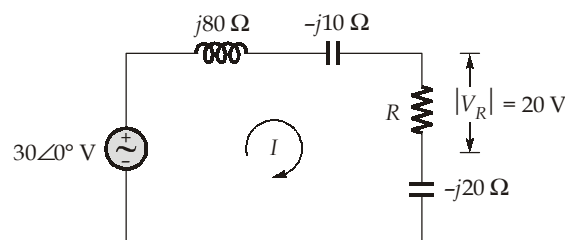
$$\therefore \frac{V_s}{i} = Z = 1 + j(10^3 L - 10^2)$$

given,  $V_s$  and  $i$  are in phase,hence,  $10^3 L - 10^2 = 0$ 

$$\Rightarrow L = \frac{10^2}{10^3} = 0.1$$

$$L = 100 \text{ mH}$$

29. (a)

Let  $I$  be the current flowing in the circuit,

$$\therefore I = \frac{30\angle 0^\circ}{j80 - j10 + R - j20} = \frac{30\angle 0^\circ}{R + j50}$$

$$|V_R| = |I|R = 20$$

$$\Rightarrow \frac{30 \times R}{\sqrt{R^2 + 50^2}} = 20$$

$$\frac{900 \times R^2}{R^2 + 2500} = 400$$

$$\Rightarrow R^2 + 2500 = \frac{9R^2}{4}$$

$$4R^2 + 10000 = 9R^2$$

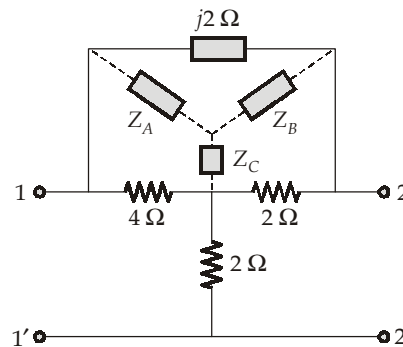
$$\Rightarrow 5R^2 = 10000$$

$$\Rightarrow R^2 = 2000$$

$$\Rightarrow R = \sqrt{2000} = 44.72 \Omega$$

30. (a)

By redrawing the given two-port network,



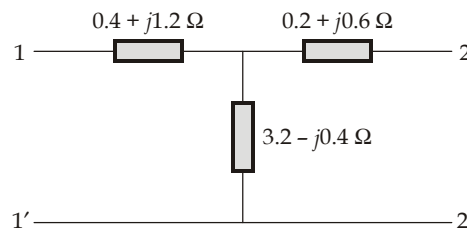
where,

$$Z_A = \frac{4 \times j2}{6 + j2} = 0.4 + j1.2 \Omega$$

$$Z_B = \frac{2 \times j2}{6 + j2} = 0.2 + j0.6 \Omega$$

$$Z_C = \frac{4 \times 2}{6 + j2} = 1.2 - j0.4 \Omega$$

After rearrangement, consider the network,



$$\therefore Z_{21} = \left. \frac{V_2}{I_1} \right|_{I_2=0} = 3.2 - j0.4 \Omega$$

$$\therefore |Z_{21}| = 3.22 \Omega$$

