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COMMUNICATIONS

ELECTRONICS ENGINEERING

Date of Test : 30/09/2025

ANSWER KEY ➤

1. (c)	7. (a)	13. (c)	19. (a)	25. (b)
2. (b)	8. (c)	14. (d)	20. (c)	26. (b)
3. (c)	9. (a)	15. (b)	21. (b)	27. (b)
4. (d)	10. (b)	16. (b)	22. (b)	28. (b)
5. (a)	11. (b)	17. (c)	23. (d)	29. (d)
6. (c)	12. (b)	18. (d)	24. (c)	30. (b)

Detailed Explanations

1. (c)

$$E_{\max} = 40 \text{ V and } E_{\min} = 10 \text{ V}$$

$$\text{Modulation index, } \mu = \frac{E_{\max} - E_{\min}}{E_{\max} + E_{\min}} = \frac{40 - 10}{40 + 10} = 0.60$$

2. (b)

From the given angle modulated signal,

$$\text{Modulation index, } \beta = 4$$

$$\text{Message signal frequency, } f_m = \frac{2000\pi}{2\pi} = 1000 \text{ Hz} = 1 \text{ kHz}$$

According to Carson's rule,

$$\begin{aligned} \text{BW} &= (1 + \beta)2f_m = (1 + 4)(2 \times 1) \text{ kHz} \\ &= 10 \text{ kHz} \end{aligned}$$

3. (c)

$$\text{Given, } \beta = 0.25; \quad P_{FM} = 10 \text{ kW}$$

The normalized power of the carrier component in FM signal,

$$P_N = J_0^2 P_{FM} (0.98)^2 \times 10 \text{ kW} = 9.604 \text{ kW}$$

Since, there is only one significant sideband whose relative amplitude is equal to $J_1 = 0.12$. The power of each sideband = $(0.12)^2 \times 10 \text{ kW}$

$$= 0.144 \text{ kW}$$

The normalized power in the carrier and first order sideband,

$$\begin{aligned} &= 9.604 + 2(0.144) \\ &= 9.892 \text{ kW} \end{aligned}$$

4. (d)

Given that,

$$\begin{aligned} f_{LO} &> f_c \\ \text{IF or } f_{IF} &= 450 \text{ kHz} \\ f_c &= 1200 \text{ kHz} \\ f_{\text{image}} &= f_c + 2f_{IF} = 1200 + 2(450) \text{ kHz} = 2100 \text{ kHz} \end{aligned}$$

5. (a)

Speech-processing circuits in AM transmitters are used for,

- Prevention of overmodulation.
- Prevention of excessive signal bandwidth.
- Increasing the average transmitted power.

6. (c)

For coherent BFSK, the condition for orthogonality is,

$$f_1 - f_0 = \frac{n}{2T_b} = n \frac{R_b}{2}; \quad n = 1, 2, 3, \dots$$

Given that, $f_0 = 50 \text{ kHz}$ and $f_1 = 60 \text{ kHz}$. So,

$$f_1 - f_0 = 10 \text{ kHz}$$

For option (a) $\Rightarrow \frac{R_b}{2} = 4 \text{ kbps} \Rightarrow 10 \neq n(4)$

For option (b) $\Rightarrow \frac{R_b}{2} = 3 \text{ kbps} \Rightarrow 10 \neq n(3)$

For option (c) $\Rightarrow \frac{R_b}{2} = 2 \text{ kbps} \Rightarrow 10 = 5(2) = n(2)$

For option (d) $\Rightarrow \frac{R_b}{2} = 1.5 \text{ kbps} \Rightarrow 10 \neq n(1.5)$

So, only option (c) satisfies the required condition.

7. (a)

$$(BW)_{M\text{-PSK}} = \frac{(BW)_{\text{BPSK}}}{\log_2(M)}$$

$$(BW)_{32\text{-PSK}} = \frac{100}{\log_2(32)} \text{ kHz} = \frac{100}{5} \text{ kHz} = 20 \text{ kHz}$$

8. (c)

Given,

$$P(X_1) = 0.5$$

$$P(X_2) = 0.5$$

The transmission probability matrix of the given binary channel is,

$$P[Y/X] = \begin{bmatrix} 0.7 & 0.3 \\ 0.4 & 0.6 \end{bmatrix}$$

The joint probability matrix is,

$$\begin{aligned} P[X, Y] &= P[X_d] \cdot P[Y/X] \\ &= \begin{bmatrix} 0.5 & 0 \\ 0 & 0.5 \end{bmatrix} \begin{bmatrix} 0.7 & 0.3 \\ 0.4 & 0.6 \end{bmatrix} = \begin{bmatrix} 0.35 & 0.15 \\ 0.2 & 0.3 \end{bmatrix} \end{aligned}$$

9. (a)

It is given that, X and Y are statistically independent.

So,

$$\sigma_Z^2 = \sigma_X^2 + \sigma_Y^2$$

$$\sigma_X^2 = \frac{(8-0)^2}{12} = \frac{64}{12} = \frac{16}{3}$$

$$\sigma_Y^2 = \frac{(4-0)^2}{12} = \frac{16}{12} = \frac{4}{3}$$

$$\sigma_Z^2 = \frac{16}{3} + \frac{4}{3} = \frac{20}{3} = 6.66$$

10. (b)

The entropy of the source is,

$$H(x) = -\sum_{i=1}^6 P_i \log_2 P_i$$

$$= -[\log_2 P_1 + \log_2 P_2 + \log_2 P_3 + \log_2 P_4 + \log_2 P_5 + \log_2 P_6]$$

given,

$$P_1 = 0.2; P_2 = 0.1, P_3 = 0.15, P_4 = 0.05, P_5 = 0.3 \text{ and } P_6 = 0.2$$

$$= -[0.2 \log_2 0.2 + 0.1 \log_2 0.1 + 0.15 \log_2 0.15 + 0.05 \log_2 0.05 + 0.3 \log_2 0.3 + 0.2 \log_2 0.2]$$

$$H(x) = 2.4087 \text{ bits/sample}$$

The sampling rate is, $f_s = 2000 + 2 \times 6000 = 14000 \text{ Hz}$
 i.e., 14000 samples are taken per second.

Hence, the entropy of the source,

$$H(x) = 2.4087 \times 14000 = 33721.8 \text{ bits/second}$$

11. (b)

For 16-PSK, theoretical minimum channel BW required is,

$$\begin{aligned} (\text{BW})_{16\text{-PSK}} &= \frac{2(R_b / 2)}{\log_2(16)} = \frac{R_b}{4} \\ &= \frac{10^8}{4} \text{ Hz} = 25 \text{ MHz} \end{aligned}$$

Note :

- Minimum bandwidth required for base-band data transmission = $\frac{R_b}{2}$
- Minimum bandwidth required for BPSK transmission = $2 \times (\text{Base-band BW}) = 2 \times \left(\frac{R_b}{2}\right) = R_b$
- Minimum bandwidth required for M-PSK transmission = $\frac{\text{BW of BPSK}}{\log_2(M)} = \frac{R_b}{\log_2(M)}$

12. (b)

$$f_m = 10 \text{ kHz, G.B} = 1 \text{ kHz}$$

$$\text{Sampling frequency } (f_s) = 2 \times 1.25 \times 10 + 1$$

$$= 26 \text{ kHz}$$

$$\therefore \text{ Symbol rate; } r = f_s = 26 \text{ K symbols/sec}$$

$$\text{Now, Entropy; } H = -\sum_{i=1}^4 P(x_i) \log_2 P(x_i)$$

$$\text{From graph; } P(1) = \frac{1}{2}$$

$$P(2) = P(3) = \frac{1}{8}$$

$$P(4) = \lambda = \frac{1}{4}$$

$$\therefore H = -\left[\frac{1}{2} \times \log_2 2^{-1} + 2 \times \frac{1}{8} \log_2 8^{-1} + \frac{1}{4} \log_2 4^{-1} \right]$$

$$H = \frac{7}{4} = 1.75 \text{ bits/symbol}$$

$$\begin{aligned} \text{Now, Data rate} &= r.H = 26 \times 1.75 \\ &= 45.5 \text{ kbps} \end{aligned}$$

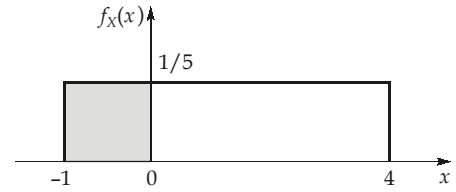
13. (c)

$$\begin{aligned} P(Y \leq 3) &= P(2X + 3 \leq 3) \\ &= P(2X \leq 0) = P(X \leq 0) \end{aligned}$$

$$P(X \leq 0) = \text{Shaded area} = \frac{1}{5}$$

So,

$$P(Y \leq 3) = P(X \leq 0) = \frac{1}{5} = 0.2$$



14. (d)

Given that, $X(t)$ is a WSS process.

So, the following two conditions should be satisfied by $X(t)$.

Condition - 1 : $E[X(t)]$ should be independent of t .

Condition - 2 : $E[X(t + \tau)X(t)]$ should be a function of τ only.

Check for condition - 1:

$$\begin{aligned} E[X(t)] &= E[A \cos t + B \sin t] \\ &= E[A \cos t] + E[B \sin t] \\ &= E[A] \cos t + E[B] \sin t \end{aligned}$$

$E[X(t)]$ will be independent of t , when and only $E[A] = E[B] = 0$.

Check for condition - 2:

$$\begin{aligned} E[X(t + \tau)X(t)] &= E[(A \cos(t + \tau) + B \sin(t + \tau))(A \cos t + B \sin t)] \\ &= E[A^2 \cos(t + \tau) \cos t + AB \sin(t + \tau) \cos t + AB \cos(t + \tau) \sin t + B^2 \sin(t + \tau) \sin t] \\ &= E\left[\frac{A^2}{2} \cos(\tau) + \frac{A^2}{2} \cos(2t + \tau) + \frac{AB}{2} \sin(2t + \tau) + \frac{AB}{2} \sin(\tau) + \frac{AB}{2} \sin(2t + \tau) \right. \\ &\quad \left. - \frac{AB}{2} \sin(\tau) + \frac{B^2}{2} \cos(\tau) - \frac{B^2}{2} \cos(2t + \tau)\right] \\ &= E\left[\left(\frac{A^2 + B^2}{2}\right) \cos(\tau) + \left(\frac{A^2 - B^2}{2}\right) \cos(2t + \tau) + AB \sin(2t + \tau)\right] \\ &= E\left[\frac{A^2 + B^2}{2}\right] \cos(\tau) + E\left[\frac{A^2 - B^2}{2}\right] \cos(2t + \tau) + E[AB] \sin(2t + \tau) \end{aligned}$$

If the above equation is to be a function of " τ " only, then

$$E\left[\frac{A^2 - B^2}{2}\right] = 0 \Rightarrow E[A^2] = E[B^2]$$

and $E[AB] = 0$

So, if $X(t)$ is a WSS process, then the conditions to be satisfied are,

$$E[A] = E[B] = 0$$

$$E[AB] = 0$$

$$E[A^2] = E[B^2]$$

15. (b)

For AWGN channels, as per Shanon Hartley Theorem

Channel capacity,

$$\begin{aligned}
 C &= B \log_2(1 + \text{SNR}) \\
 &= 5 \log_2(1 + 15) \text{ Mbps} \\
 &= 5 \times 4 = 20 \text{ Mbps}
 \end{aligned}$$

16. (b)

Sampling rate,

$$f_s = 2 \times 4000 = 8000 \text{ Hz} = 8 \text{ kHz}$$

Entropy of alphabets,

$$\begin{aligned}
 H(X) &= \frac{1}{2} \log_2(2) + \frac{1}{4} \log_2(4) + \frac{1}{8} \log_2(8) + \frac{2}{16} \log_2(16) \text{ bits/symbol} \\
 &= \frac{1}{2} + \frac{1}{2} + \frac{3}{8} + \frac{1}{2} \text{ bits/symbol} = 1.875 \text{ bits/symbol}
 \end{aligned}$$

Bit rate,

$$R_b = H(X) f_s = 8 \times 1.875 = 15 \text{ kbps}$$

Minimum baseband channel bandwidth required is,

$$(\text{BW})_{\min} = \frac{R_b}{2} = 7.50 \text{ kHz}$$

17. (c)

For Hilbert transform, the impulse response is,

$$\begin{aligned}
 h(t) &= \frac{1}{\pi t} \\
 H(\omega) &= -j \text{sgn}(\omega) \\
 |H(\omega)|^2 &= 1
 \end{aligned}$$

So, the input and output power spectral densities are related as,

$$S_Y(\omega) = |H(\omega)|^2 S_X(\omega) = (1) S_X(\omega) = S_X(\omega)$$

Since $S_Y(\omega) = S_X(\omega)$, $R_Y(\tau) = R_X(\tau) \Rightarrow$ Hence, R_1 is correct

$$\begin{aligned}
 R_{XY}(\tau) &= R_X(\tau) * h(\tau) \\
 R_{XY}(-\tau) &= R_X(-\tau) * h(-\tau)
 \end{aligned}$$

$$h(\tau) = \frac{1}{\pi \tau} \text{ and } h(-\tau) = -h(\tau)$$

$$R_X(-\tau) = R_X(\tau)$$

 \therefore ACF of a WSS process is an even function

So,

$$R_{XY}(-\tau) = R_X(\tau) * [-h(\tau)] = -[R_X(\tau) * h(\tau)]$$

$$R_{XY}(-\tau) = -R_{XY}(\tau) \Rightarrow \text{Hence, } R_2 \text{ is correct}$$

So, both the given relations are correct.

18. (d)

$$P(z_0) = P(x_0) + 0.50P(x_1) = 0.40 + (0.50 \times 0.20) = 0.50$$

$$P(z_1) = P(x_2) + 0.50P(x_1) = 0.40 + (0.50 \times 0.20) = 0.50$$

$$P(y_0) = 0.50P(z_0) = 0.50 \times 0.50 = 0.25$$

$$P(y_1) = 0.50P(z_0) + 0.50P(z_1) = (0.50 \times 0.50) + (0.50 \times 0.50) = 0.50$$

$$P(y_2) = 0.50P(z_1) = 0.50 \times 0.50 = 0.25$$

So, the entropy of the output symbols can be given by,

$$H(Y) = -[0.25 \log_2(0.25) + (0.25) \log_2(0.25) + 0.50 \log_2(0.50)] \text{ bits/}$$

symbol

$$= 1.50 \text{ bits/symbol}$$

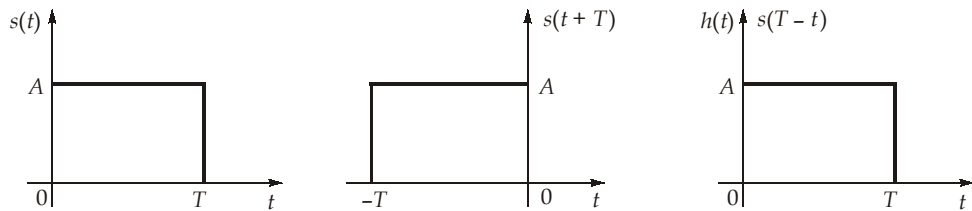
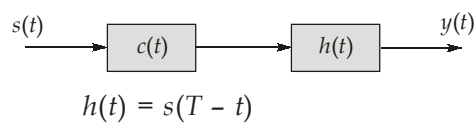
19. (a)

$$E[Y] = E\left[\int_0^{10} X(t) dt\right] = \int_0^{10} E[X(t)] dt$$

$$= \int_0^{10} 8 dt = 80$$

$$\therefore E[X(t)] = 8$$

20. (c)

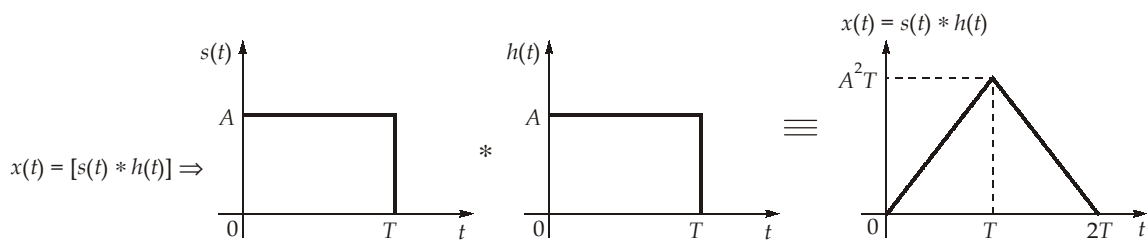


Given that,

$$c(t) = \delta(t) + \frac{1}{2}\delta\left(t - \frac{T}{2}\right)$$

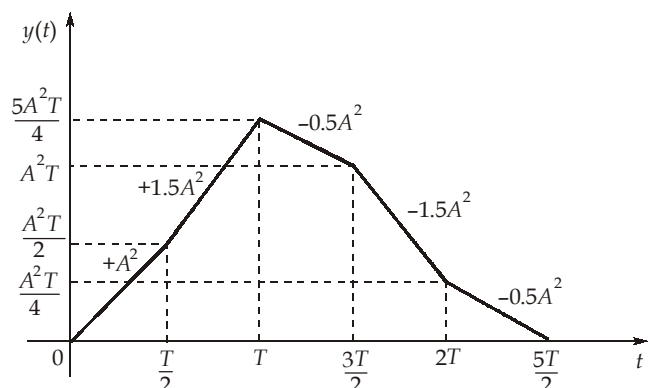
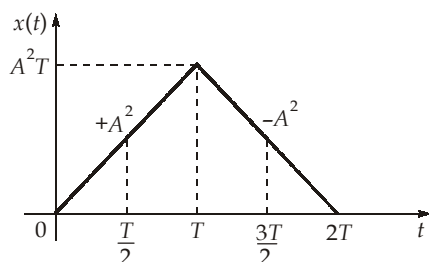
So,

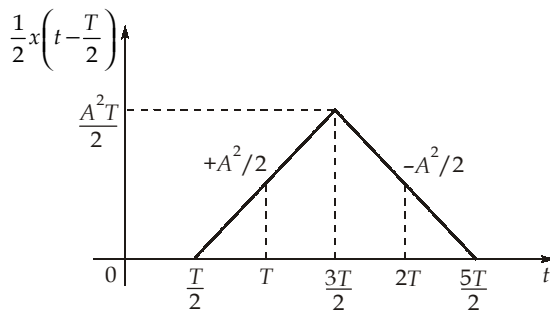
$$y(t) = s(t) * c(t) * h(t) = [s(t) * h(t)] * c(t)$$



$$y(t) = x(t) * c(t)$$

$$= x(t) * \left[\delta(t) + \frac{1}{2}\delta\left(t - \frac{T}{2}\right) \right] = x(t) + \frac{1}{2}x\left(t - \frac{T}{2}\right)$$





So, the peak value of the filter output is,

$$y_{\max} = \frac{5}{4} A^2 T$$

21. (b)

Sampling rate,

$$f_{m(\max)} = 15 \text{ kHz}$$

Bits/sample,

$$f_s = 2f_{m(\max)} = 30 \text{ kHz}$$

Bit rate,

$$n = \log_2(256) = 8 \text{ bits/sample}$$

Theoretical minimum channel BW required is,

$$R_b = nf_s = 240 \text{ kbps}$$

$$(BW)_{\min} = \frac{R_b}{2} = 120 \text{ kHz}$$

22. (b)

$$H(X) = -\sum_{i=0}^2 P(x_i) \log_2 P(x_i)$$

$$= -[0.25 \log_2(0.25) + 0.25 \log_2(0.25) + 0.50 \log_2(0.50)] \text{ bits/symbol}$$

$$= [0.50 \log_2(4) + 0.50 \log_2(2)] \text{ bits/symbol}$$

$$= 1.50 \text{ bits/symbol}$$

23. (d)

$$P(y_0 | x_0) = (0.8 \times 0.9) + (0.2 \times 0.2) = 0.72 + 0.04 = 0.76$$

$$P(y_0 | x_1) = (0.1 \times 0.9) + (0.9 \times 0.2) = 0.09 + 0.18 = 0.27$$

$$P(y_0) = P(y_0 | x_0)P(x_0) + P(y_0 | x_1)P(x_1)$$

$$= (0.76)(0.60) + (0.27)(0.40) = 0.456 + 0.108 = 0.564$$

24. (c)

$$H(X) = H(X | A) P(A) + H(X | B) P(B) + H(X | C) P(C)$$

$$H(X | A) = \text{Entropy of the probabilities } \{0, 1, 0\} = 0 \text{ bits/symbol}$$

$$H(X | B) = \text{Entropy of the probabilities } \{0, 0.50, 0.50\} = 1 \text{ bits/symbol}$$

$$H(X | C) = \text{Entropy of the probabilities } \{0, 1, 0\} = 0 \text{ bits/symbol}$$

$$P(A) = P(A | A) P(A) + P(A | B) P(B) + P(A | C) P(C) = 0.50 P(B)$$

$$P(C) = P(C | A) P(A) + P(C | B) P(B) + P(C | C) P(C) = 0.50 P(B)$$

$$P(A) + P(B) + P(C) = 0.50 P(B) + P(B) + 0.50 P(B) = 1 \Rightarrow P(B) = 0.50$$

$$P(A) = P(C) = 0.50 P(B) = 0.25$$

So,

$$H(X) = [0 \times P(A)] + [1 \times P(B)] + [0 \times P(C)] \text{ bits/symbol}$$

$$= 1 \times P(B) = 1 \times 0.50 = 0.50 \text{ bits/symbol}$$

25. (b)

Let, the bit rate at the output of information source is R_b and at the output of channel encoder is R_c .

For error-free transmission,

$$\begin{aligned} R_c &\leq C \\ R_{c(\max)} &= C = B \log_2(1 + \text{SNR}) \\ &= 50 \log_2(1 + 127) \text{ kbps} = 50 \log_2(128) \text{ kbps} \\ &= 350 \text{ kbps} \end{aligned}$$

For a (7, 4) Hamming code,

$$R_c = \frac{7}{4} R_b$$

So, $\frac{7}{4} R_{b(\max)} = R_{c(\max)} = 350 \text{ kbps}$

$$R_{b(\max)} = \frac{4}{7} \times 350 = 200 \text{ kbps}$$

26. (b)

The channel capacity of an AWGN channel can be given by,

$$C = B \log_2(1 + \text{SNR})$$

Given that, $B = 4 \text{ kHz}$ and $\text{SNR} = 1023$.

So, $C = 4 \log_2(1 + 1023) \text{ kbps} = 40 \text{ kbps}$

For error-free transmission, $R_b \leq C$.

So, $R_{b(\max)} = C = 40 \text{ kbps}$

Bit rate, $R_b = R_s H(X)$; R_s = symbol rate

Entropy, $H(X) = \log_2(32) = 5 \text{ bits/symbol}$

\therefore symbols are equiprobable

So, $R_{s(\max)} H(X) = R_{b(\max)} = 40 \text{ kbps}$

Maximum symbol rate,

$$R_{s(\max)} = \frac{R_{b(\max)}}{H(X)} = \frac{40 \times 1000 \text{ bits/sec}}{5 \text{ bits/symbol}} = 8000 \text{ symbols/sec}$$

27. (b)

When $m(t)$ is applied as message signal:

$m(t)$ is bandlimited to, $f_1 = 4 \text{ kHz}$

BW of FM signal, $B_1 = 200 \text{ kHz}$

$$B_1 = (1 + D_1) 2f_1 ; D_1 = \text{deviation ratio}$$

$$(1 + D_1) = \frac{B_1}{2f_1} = \frac{200}{8} = 25$$

$$D_1 = \frac{\Delta f_{\max(1)}}{f_1} = 24$$

When $m(2t)$ is applied as message signal:

$m(2t)$ is bandlimited to, $f_2 = 2f_1 = 8 \text{ kHz}$

$$|m(2t)|_{\max} = |m(t)|_{\max}$$

So, $\Delta f_{\max(2)} = k_f |m(2t)|_{\max} = k_f |m(t)|_{\max} = \Delta f_{\max(1)}$

$$D_2 = \frac{\Delta f_{\max(2)}}{f_2} = \frac{\Delta f_{\max(1)}}{2f_1} = \frac{D_1}{2} = 12$$

Bandwidth,

$$B_2 = (1 + D_2)f_2 = (1 + 12) (2) (8) \text{ kHz} \\ = 208 \text{ kHz}$$

28. (b)

The standard form of an FM signal can be given as,

$$s(t) = A_c \cos \left[2\pi f_c t + 2\pi k_f \int_{-\infty}^t m(t) dt \right]$$

The maximum phase deviation of the FM signal can be given as,

$$\Delta \phi_{\max} = 2\pi k_f \left| \int_{-\infty}^t m(t) dt \right|_{\max}$$

Since $m(t)$ has only positive values,

$$\left| \int_{-\infty}^t m(t) dt \right|_{\max} = \text{Area under } m(t) = 4 \times 2 = 8 \text{ V-sec}$$

$$\Delta \phi_{\max} = 2\pi k_f \left| \int_{-\infty}^t m(t) dt \right|_{\max} \\ = 2\pi (0.5 \text{ Hz/V}) (8 \text{ V-sec}) \text{ rad} = 2\pi (0.5) (8) \text{ rad} = 8\pi \text{ rad}$$

29. (d)

For a uniform quantizer, the maximum quantization error is $\frac{\Delta}{2}$.

So, $\frac{\Delta}{2} \leq \frac{0.1}{100}(a)$; Where, Δ = step-size

$$\Delta \leq \frac{a}{500}$$

$$\frac{2a}{2^n} \leq \frac{a}{500} \quad \dots(i)$$

$$\therefore \Delta = \frac{a - (-a)}{L} = \frac{2a}{2^n}$$

Where, L = Number of quantization levels and n = number of bits per sample.

So, from equation (i), $2^n \geq 1000$

$$n \geq 10$$

$\therefore n$ can be an integer only

$$n_{\min} = 10$$

30. (b)

Given,

$$m(t) = 10 \cos 16 \pi t$$

$$u(t) = 10 \cos \left[4000\pi t + 2\pi K_f \int_{-\infty}^t m(\tau) d\tau \right]$$

$$\text{modulation index, } \beta = \frac{K_f \cdot A_m}{f_m}$$

where,

$$A_m = 10 \text{ V}$$

$$f_m = 8 \text{ Hz}$$

$$\therefore \beta = \frac{10 \times 10}{8} = 12.5$$

\therefore Output of FM modulator can be written as,

$$u(t) = 10 \cos \left[2\pi \times 2000 \times t + 2\pi K_f \int_{-\infty}^t 10 \cos(2\pi \times 8\tau) d\tau \right]$$

$$= \sum_{n=-\infty}^{\infty} 10 J_n(\beta) \cos(2\pi(2000 + n \cdot 8)t + \phi_n)$$

The BPF passes the frequencies in the range 1969 Hz to 2031 Hz. At the output of BPF, the frequency components of signal present are (2000 - 24), (2000 - 16), (2000 - 8), 2000, (2000 + 8), (2000 + 16) and (2000 + 24) Hz. These components are the terms of $u(t)$ for which, $n = -3, \dots, 3$. The power of the output signal is then,

$$P_{SB} = \frac{10^2}{2} J_0^2(\beta) + 2 \sum_{n=1}^3 \frac{10^2}{2} J_n^2(\beta)$$

where,

$$\beta = 12.5$$

$$= \frac{10^2}{2} J_0^2(12.5) + 2 \sum_{n=1}^3 \frac{10^2}{2} J_n^2(12.5)$$

$$= 2 \left[50 \times [J_1^2(12.5) + J_2^2(12.5) + J_3^2(12.5)] \right]$$

$$P_{SB} = 100[(0.2)^2 + (0.3)^2 + (0.25)^2]$$

$$P_{SB} = 19.25 \text{ W}$$

Since the total transmitted power is, $P_{tot} = \frac{10^2}{2} = 50 \text{ W}$, the power at the output of BPF is 19.25 W

(38.5%) of the transmitted power.

