

CLASS TEST - 2016

Electrical Engineering

EE

Electrical Machines

Date : 16/06/2016

ANSWERS

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|--------|---------|---------|---------|---------|
| 1. (a) | 7. (c) | 13. (a) | 19. (c) | 25. (a) |
| 2. (b) | 8. (c) | 14. (a) | 20. (c) | 26. (c) |
| 3. (b) | 9. (c) | 15. (c) | 21. (d) | 27. (b) |
| 4. (c) | 10. (b) | 16. (b) | 22. (c) | 28. (a) |
| 5. (b) | 11. (b) | 17. (d) | 23. (a) | 29. (a) |
| 6. (b) | 12. (a) | 18. (a) | 24. (d) | 30. (b) |

1. (a)

$$\begin{aligned} \text{Available flux} &= \text{Flux density} \times \text{area} \\ &= 1.5 \times (8 \times 12.5 \times 10^{-4}) \\ &= 1.5 \times 10^{-2} \text{ Wb} \end{aligned}$$

$$\text{Leakage coefficient} = 1.2 = \frac{\text{Available flux}}{\text{Useful flux}}$$

$$\begin{aligned} \text{Useful flux} &= \frac{\text{Available flux}}{1.2} \\ &= \frac{1.5 \times 10^{-2}}{1.2} = 1.25 \times 10^{-2} \text{ Wb} \end{aligned}$$

$$\begin{aligned} \text{emf} &= \frac{\phi Z N}{60} \quad (\because \text{for lap winding } P = A) \\ &= \frac{1.25 \times 10^{-2} \times 1000 \times 600}{60} = 125 \text{ V} \end{aligned}$$

4. (c)

Method-I:

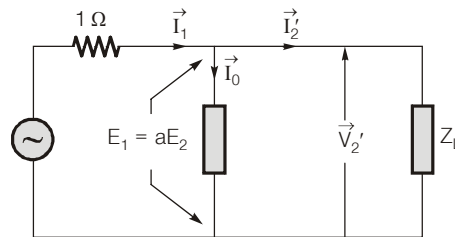
As transformer is ideal, induced emf in secondary winding = load voltage

i.e. $E_2 = V_2 = 110 \text{ Volt}$

Load current/secondary winding current,

$$I_2 = \left(\frac{4000}{110 \times 0.89} \right) = 40.86 \text{ A}$$

Transformation ratio, $a = \frac{2}{1} = 2$



(Equivalent circuit offered to primary)

Given, $\cos\phi = 0.89$ or, $\phi = 27.12^\circ$

Secondary current referred to primary,

$$I_2' = \frac{I_a}{a} = 20.43 \text{ A}$$

$$\therefore \vec{I}_2 = 20.43 \angle -27.12^\circ \text{ A,}$$

$$E_1 = aE_2 = 2 \times 110 = 220 \text{ Volt}$$

Now, $\vec{I}_1 = \vec{I}_0 + \vec{I}_2'$

Since, input pf required is unity and the load is inductive (lagging), therefore, x must be capacitive reactance.

$$\therefore \vec{I}_1 \angle 0^\circ = \frac{E_1}{-jX} + \vec{I}_2' = \frac{jE_1}{X} \times \vec{I}_2'$$

$$= \left(\frac{j220}{X} \right) + (20.43 \angle -27.12^\circ)$$

$$= \frac{j220}{X} + (18.18 - j9.31)$$

or, $\vec{I}_1 \angle 0^\circ = \sqrt{(18.18)^2 + \left(\frac{220}{X} - 9.31\right)^2} \angle \theta \dots(i)$

where, $\theta = \tan^{-1} \left[\frac{\frac{220}{X} - 9.31}{18.18} \right]$

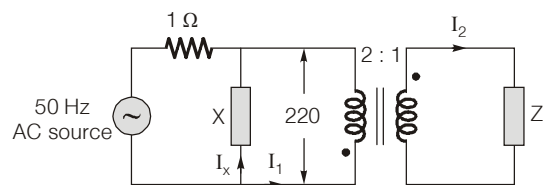
Equating angle of equation (i) on both sides, we get

$$\tan^{-1} \left(\frac{\frac{220}{X} - 9.31}{18.18} \right) = 0^\circ$$

or, $\frac{220}{X} - 9.31 = 0$

or, $X = 23.63 \Omega$

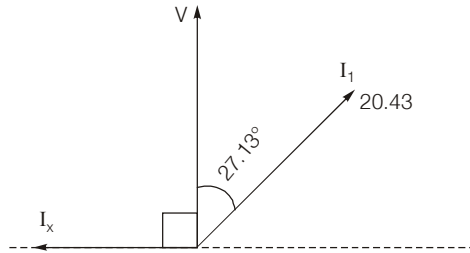
Method-II:



$$I_2 = \frac{4 \times 10^3}{110 \times 0.89} = 40.86 \angle -27.13^\circ$$

$$I_1 = \frac{40.86}{2} \angle -27.13^\circ$$

$$= 20.43 \angle -27.73^\circ$$



For unity pf horizontal component of (I_1) should be cancelled with (I_x).

$$\text{So, } 20.43 \sin(27.13^\circ) = \frac{220}{X}$$

$$X = 23.61 \Omega$$

5. (b)

$$I_{S(\text{base})} = \frac{100}{\sqrt{3} \times 115} = 502 \text{ A}$$

Since the transformer supplies a load of 80 MVA at 0.85 pf lagging, so secondary line current of the transformer is

$$I_S = \frac{80}{\sqrt{3} \times 115} = 402 \text{ A}$$

$$(I_S)_{\text{pu}} = \frac{402}{502} \angle -\cos^{-1}(0.85) = 0.8 \angle -31.8^\circ$$

per unit no load voltage of this transformer is

$$V_{\text{NL}} = 1 \angle 0^\circ + (0.8 \angle -31.8^\circ) (0.02 + j 0.055)$$

$$= 1.037 \angle 1.6^\circ$$

$$\text{V.R.} = \frac{1.037 - 1}{1} \times 100 \% = 3.7 \%$$

10. (b)

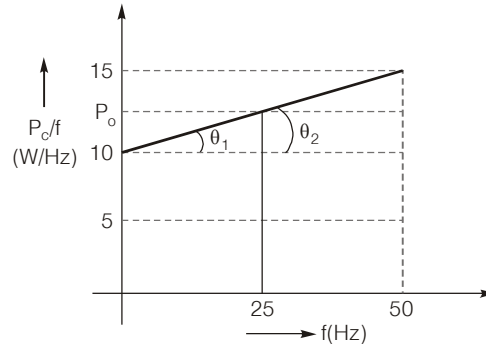
$$N\phi = Li$$

$$1000 \times 10^{-3} = L \times 1$$

$$\Rightarrow L = 1 \text{ H}$$

$$\Rightarrow W_E = \frac{1}{2} Li^2 = \frac{1}{2} \text{ J}$$

11. (b)



Let core loss at 25 Hz be P_o per Hz.

Then, equating the two slopes of figure, we have:

$$\left(\frac{P_o - 10}{25} \right) = \frac{5}{50}$$

or, $P_o = 2.5 + 10 = 12.5$ W/Hz

\therefore Core loss at 25 Hz = $12.5 \times 25 = 312.5$ Watt

Also, core loss at 50 Hz = $15 \times 50 = 750$ Watt

Let at (f_1) 25 Hz, eddy current loss = P_{e_1}

and hysteresis loss = P_{h_1}

We know that: $P_e \propto f^2$

$$\therefore \frac{P_{e_2}}{P_{e_1}} = \left(\frac{f_2}{f_1} \right)^2 = \left(\frac{50}{25} \right)^2 = 4$$

$$\therefore P_{e_2} = 4P_{e_1} \quad \dots(i)$$

Also, $P_h \propto f$

$$\therefore \frac{P_{h_2}}{P_{h_1}} = \frac{f_2}{f_1} = \frac{50}{25} = 2$$

$$\therefore P_{h_2} = 2P_{h_1} \quad \dots(ii)$$

Now, at $f_1 = 25$ Hz,

$$P_{e_1} + P_{h_1} = 312.5 \quad \dots(iii)$$

and at $f_2 = 50$ Hz

$$P_{e_2} + P_{h_2} = 750 \text{ W}$$

$$\text{or, } 4P_{e_1} + 2P_{h_1} = 750 \text{ W}$$

$$\text{or, } 2P_{e_1} + P_{h_1} = 375 \text{ W} \quad \dots(iv)$$

(Using equation (i) and (ii))

Solving equations (iii) and (iv), we get

$$P_{e_1} = 62.5 \text{ Watt and } P_{h_1} = 250 \text{ Watt}$$

Thus, at 25 Hz, hysteresis loss = 250 W and eddy current loss = 62.5 W.

12. (a)

Armature resistance is assumed negligible. Further field current is ignored in comparison to armature current, i.e.,

$$I_L = I_a$$

As per the data given

$$E = K_e N$$

$$200 = K_e \times 600 \quad \dots(i)$$

$$T = K_t I_a$$

$$= K_t \times 30 = K_L \times (600)^2 \quad \dots(ii)$$

With a 20 Ω resistor added in the armature circuit

$$(200 - 20I_a) = K_e \times N \quad \dots(iii)$$

$$K_t I_a = K_L N^2 \quad \dots(iv)$$

Dividing equation (iii) by (i) and (iv) by (ii)

$$\text{we get, } \frac{200 - 20I_a}{200} = \frac{N}{600} \quad \dots(v)$$

$$\text{and } \frac{I_a}{30} = \frac{N^2}{(600)^2} \quad \dots(vi)$$

on solving, $N = 260.5 \text{ rpm}$

15. (c)

Pole pitch = distance between two adjacent poles

$$= \frac{\text{periphery of the armature}}{\text{number of poles of the generator}}$$

$$= \frac{\pi D}{P} = \frac{\pi \times 0.35}{4} \text{ m}$$

$$\frac{\text{Poles arc}}{\text{pole pitch}} = 0.7$$

$$\text{Pole arc} = 0.7 \times \text{pole pitch} = \frac{0.7 \times \pi \times 0.35}{4}$$

$$\begin{aligned} \text{Area of pole face} &= \text{pole arc} \times \text{axial length} \\ &= \frac{0.7 \times \pi \times 0.35}{4} \times 0.2 = 0.03848 \text{ m}^2 \end{aligned}$$

$$E = \frac{NP\Phi Z}{60A}$$

$$250 = \frac{500 \times 4\Phi \times 1200}{60 \times 4}$$

$$\Phi = \frac{250 \times 60 \times 4}{500 \times 4 \times 1200} = 0.025 \text{ Wb}$$

Flux density in the air gap,

$$B = \frac{\text{Flux per pole}}{\text{area of pole shoe}} = \frac{0.025}{0.03848} = 0.65 \text{ T}$$

17. (d)

$$I_{SC} = \frac{5000}{440} = 11.36 \text{ A}$$

$$\begin{aligned} V_{SC} &= I_{SC} \times Z_{eq} = 11.36 [0.5 + (2)^2 \times 0.2] \\ &= 11.36 \{ [0.5 + (2)^2 \times 0.2] + j [0.6 + (2)^2 \times 0.15] \} \\ &= 11.36 [1.3 + j1.2] \\ &= 11.36 \times 1.77 \angle 42.70^\circ \\ &= 14.77 \angle 42.70^\circ \text{ V} \end{aligned}$$

18. (a)

$$a = \frac{3300}{100} = 33$$

I_{2t} = current in the secondary of the teaser transformer

For teaser transformer,

$$P_{2t} = V_{2t} I_{2t} \cos \phi_{2t}$$

$$I_{2t} = \frac{P_{2t}}{V_{2t} \cos \phi_{2t}} = \frac{500 \times 10^3}{100 \times 0.8} = 6250 \text{ A}$$

Since the 2-phase load is balanced, the 3-phase side is also balanced. For the mmf balance of the teaser transformer,

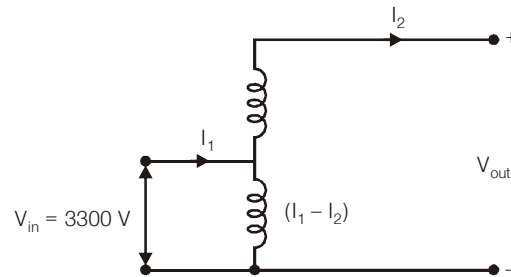
$$I_A \times \frac{\sqrt{3}}{2} T_p = I_{2t} T_s$$

$$I_A = \frac{2}{\sqrt{3}} \frac{T_s}{T_p} I_{2t} = \frac{2}{\sqrt{3}} \times \frac{6250}{33} = 218.7 \text{ A}$$

$$\therefore |I_A| = |I_B| = |I_C| = 218.7 \text{ A}$$

Line current in 3-phase sides are I_A , I_B and I_C .

21. (d)



$$V_{in} = 3300 \text{ volts}$$

$$V_{out} = 3300 + 230 = 3530 \text{ V}$$

$$I_2 = \frac{50 \times 10^3}{230} = 217.4 \text{ A}$$

These KVA rating of auto transformer

$$= V_{out} \times I_2 = (3530)(217.4) = 767.42 \text{ KVA}$$

22. (c)

$$R_1 = r_1 + r'_2 \text{ and } X_1 = x_1 + x'_2$$

The copper loss obtained during blocked rotor test = 2100 kW

$$\text{i.e., } 3I_1^2 R_1 = 2100$$

$$R_1 = \frac{2100}{3(15)^2} = 3.1 \Omega/\text{phase}$$

$$\therefore r'_2 = \frac{R_1}{2} = \frac{3.1}{2} \simeq 1.6 \Omega/\text{ph}$$

The impedance Z_1 (referred to stator)

$$= \frac{200}{\sqrt{3} \times 15} = 7.7 \Omega/\text{ph}$$

$$X_1 = x_1 + x_2' = \sqrt{(7.7)^2 - (3.1)^2}$$

$$= 7 \Omega$$

$$\therefore x_2' = \frac{X_1}{2} = 3.5 \Omega$$

$$\therefore T = \frac{sE_1^2 r_2'}{r_2'^2 + x_2'^2 s^2}$$

$$\Rightarrow \frac{0.03(E_1^2)(1.6)}{(1.6)^2 + (0.03)^2(3.5)^2} = \frac{E_1^2(0.2)(r')}{(r')^2 + (0.2)^2(3.5)^2}$$

(1970 rpm, $s = 0.03$; 800 rpm, $s = 0.2$)

$$r' = 10.7 \text{ or } 0.05 \Omega$$

extra resistance to be added with rotor is

$$= 10.7 - 1.6$$

$$= 9.1 \Omega/\text{ph}$$

23. (a)

Phase angle of the main winding current

$$\angle I_m = -\angle Z_m = -\angle(6 + 4j)$$

$$= -33.7^\circ$$

and phase angle of the auxiliary winding current with capacitor in series

$$\angle I_a = -\angle(8 + 6j - j/\omega C)$$

$$= -\tan^{-1}\left(\frac{6 - 1/\omega C}{8}\right)$$

Now, the phase difference between the currents in the main and auxiliary winding.

$$\text{Now } \alpha = \angle I_m - \angle I_a$$

$$90^\circ = -33.7 + \tan^{-1}\left(\frac{6 - 1/\omega C}{8}\right)$$

$$\Rightarrow 123.7 = \tan^{-1}\left(\frac{6 - 1/\omega C}{8}\right)$$

$$\Rightarrow \frac{6 - 1/\omega C}{8} = -1.5$$

$$\Rightarrow \frac{1}{\omega C} = 18$$

$$\Rightarrow C = \frac{1}{2\pi \times 50 \times 18}$$