



MADE EASY

India's Best Institute for IES, GATE & PSUs

Delhi | Bhopal | Hyderabad | Jaipur | Pune | Bhubaneswar | Kolkata

Web: www.madeeasy.in | E-mail: info@madeeasy.in | Ph: 011-45124612

EC | EE

Control Systems

Duration : 1:00 hr.

Maximum Marks : 50

Read the following instructions carefully

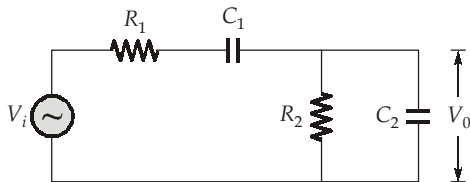
1. This question paper contains **30** objective questions. **Q.1-10** carry one mark each and **Q.11-30** carry two marks each.
2. Answer all the questions.
3. Questions must be answered on Objective Response Sheet (**ORS**) by darkening the appropriate bubble (marked **A, B, C, D**) using HB pencil against the question number. Each question has only one correct answer. In case you wish to change an answer, erase the old answer completely using a good soft eraser.
4. There will be **NEGATIVE** marking. For each wrong answer **1/3rd** of the full marks of the question will be deducted. More than one answer marked against a question will be deemed as an incorrect response and will be negatively marked.
5. Write your name & Roll No. at the specified locations on the right half of the **ORS**.
6. No charts or tables will be provided in the examination hall.
7. Choose the **Closest** numerical answer among the choices given.
8. If a candidate gives more than one answer, it will be treated as a **wrong answer** even if one of the given answers happens to be correct and there will be same penalty as above to that questions.
9. If a question is left blank, i.e., no answer is given by the candidate, there will be **no penalty** for that question.

DO NOT OPEN THIS TEST BOOKLET UNTIL YOU ARE ASKED TO DO SO

Q.No. 1 to Q.No. 10 carry 1 mark each

Q.1 For the system shown in below figure, what is the transfer function $\frac{V_0(s)}{V_i(s)}$?

(Given $R_1 = R_2 = C_1 = C_2 = 1$ unit)



(a) $\frac{V_0(s)}{V_i(s)} = \frac{s}{s^2 + 3s + 1}$

(b) $\frac{V_0(s)}{V_i(s)} = \frac{1}{s^2 + 3s + 1}$

(c) $\frac{V_0(s)}{V_i(s)} = \frac{s^2}{s^2 + 2s + 1}$

(d) $\frac{V_0(s)}{V_i(s)} = \frac{s + 1}{s^2 + 2s + 1}$

Q.2 The damping ratio ξ , natural frequency ω_n and damping frequency ω_d of a system with transfer function,

$$\frac{Y(s)}{R(s)} = \frac{64}{s^2 + 8s + 64}$$

(a) $\xi = 0.5; \omega_n = 8; \omega_d = 6.93$

(b) $\xi = 1, \omega_n = 8; \omega_d = 0$

(c) $\xi = 0.5; \omega_n = 64; \omega_d = 55.42$

(d) $\xi = 0.8; \omega_n = 8; \omega_d = 6.93$

Q.3 What is the impulse response $y(t)$ of a system with transfer function

$$\frac{Y(s)}{R(s)} = \frac{1}{(s + 1)(s + 10)}$$

(a) $y(t) = \frac{1}{9}e^{-t} + \frac{1}{9}e^{-10t}$

(b) $y(t) = \frac{1}{9}e^{-t} - \frac{1}{9}e^{-10t}$

(c) $y(t) = e^{-t} - e^{-10t}$

(d) $y(t) = e^{-t} + e^{-10t}$

Q.4 The forward transfer function of a unity feedback system is $G(s) = \frac{K(s^2 + 1)}{(s + 1)(s + 2)}$. The

system is stable for:

(a) $K < -1$ (b) $K > -1$

(c) $K < -2$ (d) $K > -2$

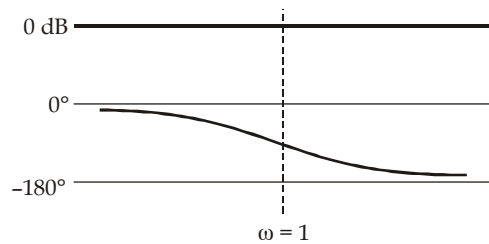
Q.5 The phase margin of a system with the open loop transfer function,

$$G(s)H(s) = \frac{(1 - s)}{(1 + s)(3 + s)}$$
 is

(a) ∞ (b) 90°

(c) 68.3° (d) 0

Q.6 For the Bode plot shown in figure, the transfer function is,



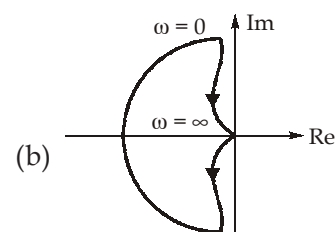
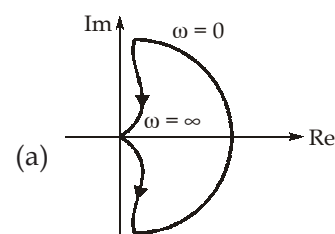
(a) $\frac{1 - s}{1 + s}$ (b) $\frac{1}{(1 + s)^2}$

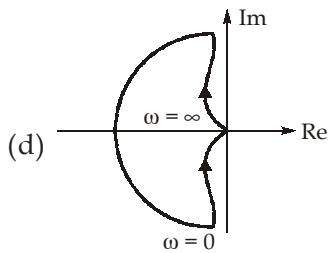
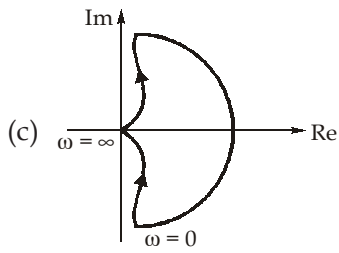
(c) $\frac{1}{s^2}$ (d) $\frac{1}{s(s + 1)}$

Q.7 The OLTF of a feedback control system is

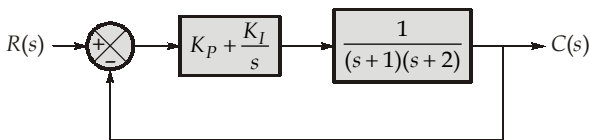
$$G(s)H(s) = \frac{-1}{2s(1 - 20s)}$$
, the Nyquist plot of

the system is





Q.8 A stable closed loop control system with PI controller is shown below,



The suitable values of controller parameters are

- (a) $K_p > 0$ and $K_I > \frac{K_p}{3} - 3$
- (b) $K_I > 0$ and $K_p > \left(\frac{K_I}{3} - 2\right)$
- (c) $K_I > \frac{K_p}{3} - 2$ (or) $K_p > 0$
- (d) $K_I > 0, K_p > 0$

Q.9 The state model of a system is given as,

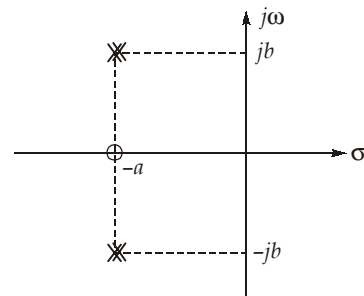
$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ 1 & 1 \end{bmatrix} \begin{bmatrix} x_2 \\ x_1 \end{bmatrix}$$

$$x^T(0) = [1 \ 0]$$

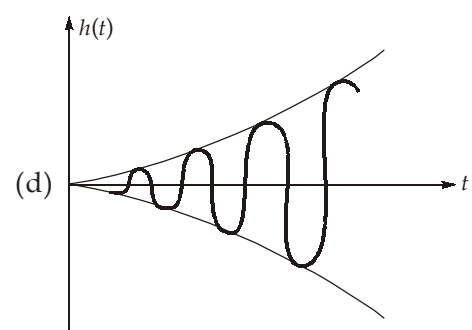
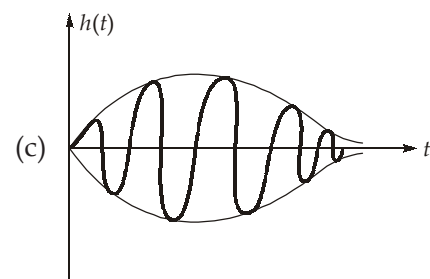
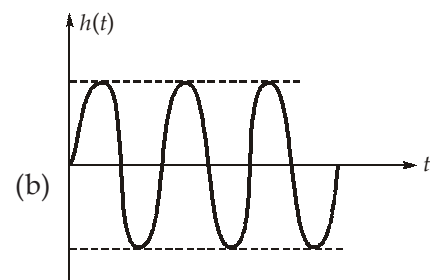
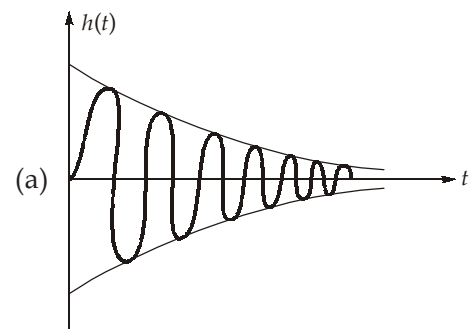
The solution of homogenous equation is given by,

- (a) $x(t) = \begin{bmatrix} e^t \\ 1 \end{bmatrix}$
- (b) $x(t) = \begin{bmatrix} e^t \\ te^t \end{bmatrix}$
- (c) $x(t) = \begin{bmatrix} 0 \\ e^t \end{bmatrix}$
- (d) $x(t) = \begin{bmatrix} e^t \\ e^t \end{bmatrix}$

Q.10 A system shown below has multiple poles and a zero as shown below.



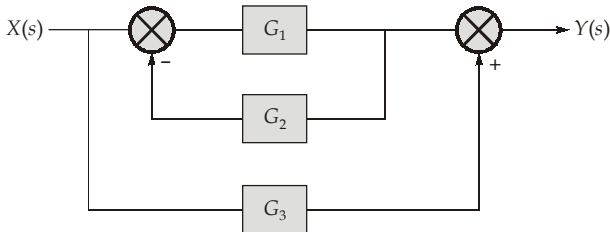
The impulse response of the system is



Q. No. 11 to Q. No. 30 carry 2 marks each

Q.11 The equivalent transfer function of the system shown in block diagrams is:

$$G_1 = \frac{1}{s+1}, \quad G_2 = \frac{1}{s+4} \quad \text{and} \quad G_3 = \frac{s+2}{s+5}$$



- (a) $\frac{s^3 + 4s^2 + 18s + 20}{s^3 + 5s^2 + 10s + 14}$
- (b) $\frac{s^3 + 8s^2 + 24s + 30}{s^3 + 10s^2 + 30s + 25}$
- (c) $\frac{s^3 + 4s^2 + 18s + 20}{s^3 + 10s^2 + 30s + 25}$
- (d) $\frac{s^3 + 8s^2 + 24s + 30}{s^2 + 5s + 10s + 14}$

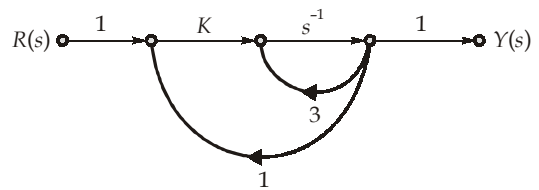
Q.12 A control system is represented by the given below differential equation,

$$\frac{d^2y}{dt^2} + 5\frac{dy}{dt} + 4y = x(t)$$

has all initial conditions zero. If an input $x(t) = 8 u(t)$, then output $y(t)$ is

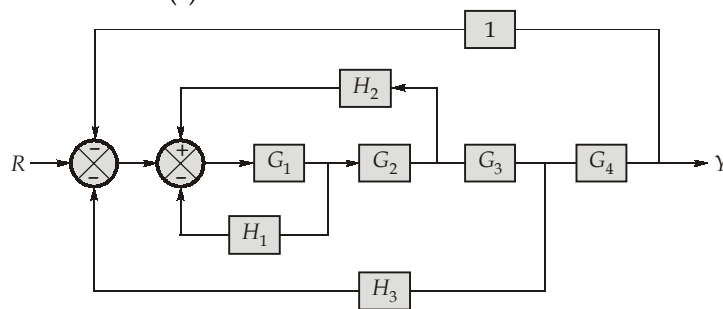
- (a) $\left[1 - \frac{1}{4}e^{-4t} + \frac{1}{2}e^{-t}\right]u(t)$
- (b) $\left[1 + \frac{1}{4}e^{-4t} + \frac{1}{3}e^{-t}\right]u(t)$
- (c) $\left[1 + \frac{1}{8}e^{-4t} - \frac{1}{3}e^{-t}\right]u(t)$
- (d) $2\left[1 + \frac{1}{3}e^{-4t} - \frac{4}{3}e^{-t}\right]u(t)$

Q.13 The system shown in the figure remains stable when



- (a) $K < -1$
- (b) $-1 < K < 1$
- (c) $1 < K < 3$
- (d) $K < -3$

Q.14 Find the transfer function $\frac{Y(s)}{R(s)}$ for the system with following block diagram:



- (a) $\frac{Y(s)}{R(s)} = \frac{G_1G_2G_3G_4}{1 + G_1H_1 - G_1G_2H_2 + G_1G_2G_3H_3 + G_1G_2G_3G_4}$
- (b) $\frac{Y(s)}{R(s)} = \frac{G_1G_2G_3G_4}{1 - G_1H_1 + G_1G_2H_2 + G_1G_2G_3H_3}$
- (c) $\frac{Y(s)}{R(s)} = \frac{G_1G_2G_3G_4}{1 + G_1H_1 + G_1G_2H_2 + G_1G_2G_3H_3}$
- (d) $\frac{Y(s)}{R(s)} = \frac{G_1G_2G_3G_4}{1 - G_1H_1 - G_1G_2H_2 + G_1G_2G_3H_3 + G_1G_2G_3G_4}$

Q.15 The closed loop transfer function of a system is,

$$T(s) = \frac{s^3 + 4s^2 + 8s + 16}{s^5 + 3s^4 + 5s^2 + 4s + 3}$$

The number of poles in the right half plane and in left half plane are

- (a) 3, 2 (b) 1, 4
 (c) 2, 3 (d) 4, 1

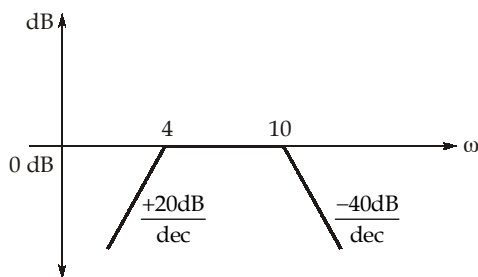
Q.16 Consider a unity feedback system with forward path transfer function $G(s) =$

$$\frac{K(s+3)(s+5)}{(s-2)(s-4)}$$

The range of K to ensure stability is

- (a) $K > \frac{3}{4}$
 (b) $K < -1$ (or) $K > \frac{3}{4}$
 (c) $K < -1$
 (d) $-1 < K < \frac{3}{4}$

Q.17 For the Bode plot shown below the transfer function is,

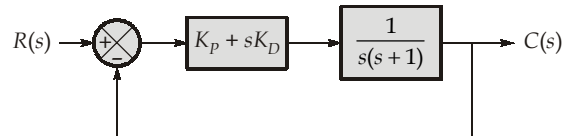


- (a) $\frac{100s}{(s+4)(s+10)^2}$
 (b) $\frac{100(s+4)}{s(s+10)^2}$

(c) $\frac{100}{(s+4)(s+10)}$

(d) $\frac{100}{s^2(s+4)(s+10)}$

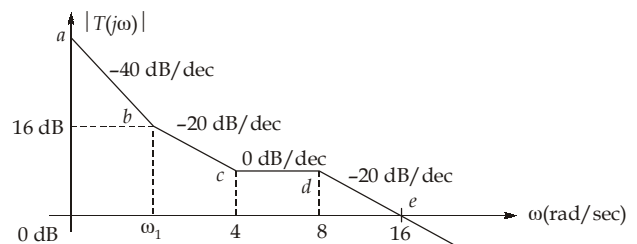
Q.18 The block diagram of a control system is shown below,



The control specification are such that the damping ratio of the closed loop system is 0.6 and damped frequency, ω_d is 10 rad/sec. The parameters of PD controller K_p and K_D are

- (a) 156.25, 14
 (b) 277.78, 19
 (c) 2, 4
 (d) 11, 100

Q.19 The asymptotic log-magnitude curve for open loop transfer function is sketched below,



Open loop transfer function is

(a) $T(s) = \frac{10(s+8)(s+4)}{s^2(s+1.268)}$

(b) $T(s) = \frac{16(s+1.268)(s+4)}{s^2(s+8)}$

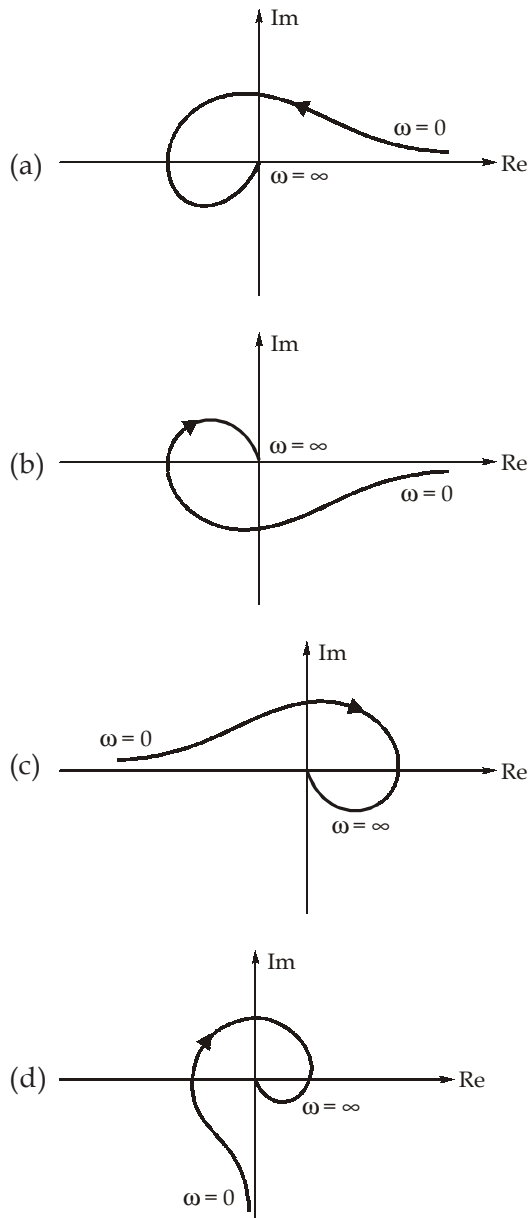
(c) $T(s) = \frac{10(s+1.268)(s+8)}{s^2(s+4)}$

(d) $T(s) = \frac{8(s+1.268)(s+8)}{s^2(s+4)}$

Q.20 Consider a control system having open-loop transfer function of,

$$G(s)H(s) = \frac{1}{s^2(s-a)(s-b)(s-c)}$$

If a, b, c are positive, then the polar plot of the system will be



Q.21 A control system is represented by following state space representation

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} -5 & 1 \\ 0 & -4 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u(t)$$

$$y = \begin{bmatrix} 1 & 4 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

If two such systems are connected in parallel then overall transfer function of the combined system is

- (a) $\frac{4s+21}{(s+4)(s+5)}$ (b) $\frac{4s+36}{(s+4)(s+5)}$
 (c) $\frac{2(s+9)}{(s+4)(s+5)}$ (d) $\frac{2(4s+21)}{(s+4)(s+5)}$

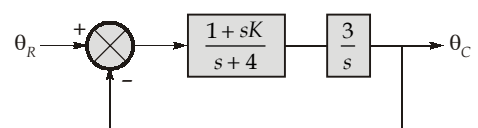
Q.22 If the transfer function of a control system is given as below:

$$\frac{Y(s)}{U(s)} = \frac{1}{s^3 + 4s^2 + 5s + 3}$$

The matrix $[A]$ and $[B]$ in state variable representation of the system will be

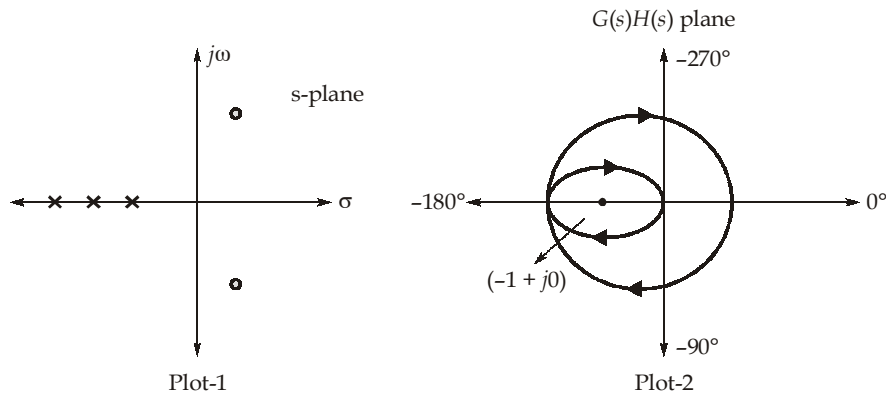
- (a) $[A] = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -4 & -5 & -3 \end{bmatrix}; [B] = \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix}$
 (b) $[A] = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -3 & -5 & -4 \end{bmatrix}; [B] = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$
 (c) $[A] = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -8 & -10 & -6 \end{bmatrix}; [B] = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$
 (d) $[A] = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -6 & -10 & -18 \end{bmatrix}; [B] = \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix}$

Q.23 A control system with damping ratio $\xi = \sqrt{3}$ is represented by the block diagram which employs proportional plus error rate control. The value of error rate constant K when unit step is given as input to system will be



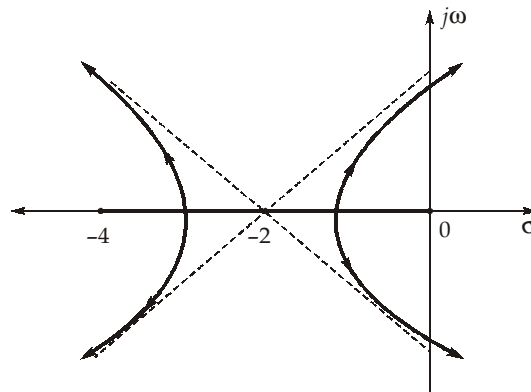
- (a) $\frac{1}{3}$ (b) $\frac{1}{2}$
 (c) $\frac{1}{4}$ (d) $\frac{2}{3}$

Q.24 The pole zero map and the Nyquist plot of the open-loop transfer function $G(s)H(s)$ of a feed back system are shown below, for this,

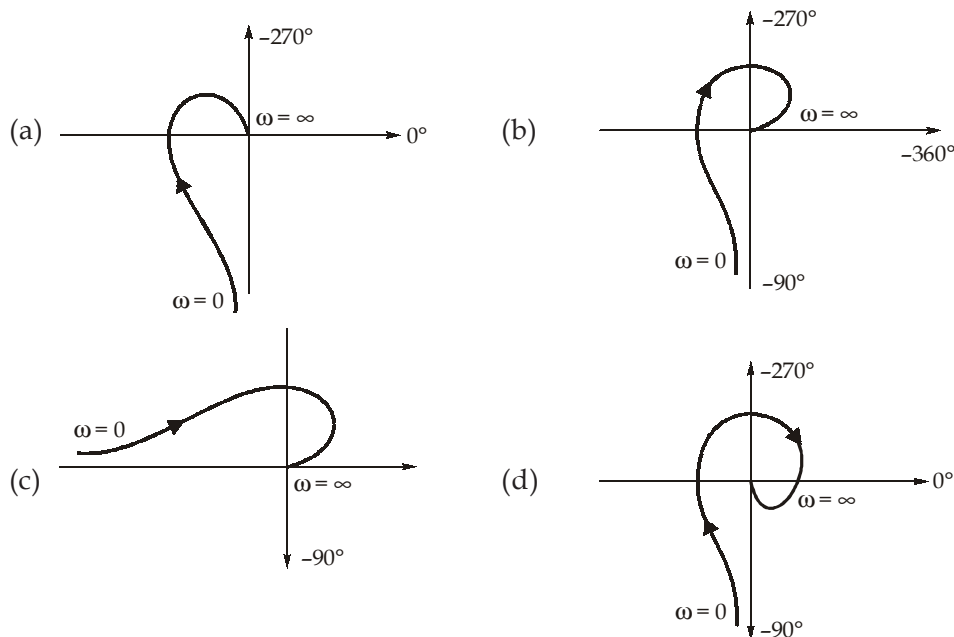


- (a) Both open-loop and closed-loop systems are stable.
- (b) Open-loop system is stable but closed-loop system is unstable.
- (c) Open-loop system is unstable but closed-loop system is stable.
- (d) Both open-loop and closed-loop systems are unstable.

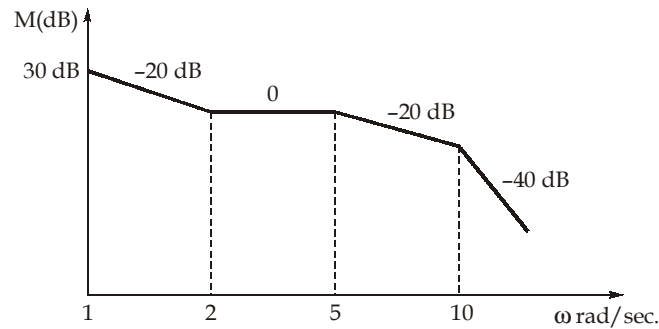
Q.25 The root locus diagram of a control system is shown below,



The polar plot of the same system is



Q.26 The open-loop transfer function of the system is,



- (a) $\frac{30\left(\frac{s}{5} + 1\right)}{s\left(\frac{s}{10} + 1\right)\left(\frac{s}{2} + 1\right)}$
- (b) $\frac{30\left(\frac{s}{2} + 1\right)}{s^2\left(\frac{s}{5} + 1\right)\left(\frac{s}{10} + 1\right)}$
- (c) $\frac{1.65(0.2s + 1)}{s(0.5s + 1)(0.1s + 1)}$
- (d) $\frac{31.62(0.5s + 1)}{s(0.2s + 1)(0.1s + 1)}$

Q.27 If $G(s)H(s) = \frac{10}{(s+1)(s+20)}$ then the gain margin and phase margin of the system are respectively.

- (a) $\frac{1}{2}, \infty$
- (b) $0, 180^\circ$
- (c) ∞, ∞
- (d) $\infty, 180^\circ$

Q.28 The state space equation of the system is given by

$$\dot{x} = \begin{bmatrix} -4 & 0 \\ 0 & -3 \end{bmatrix} x$$

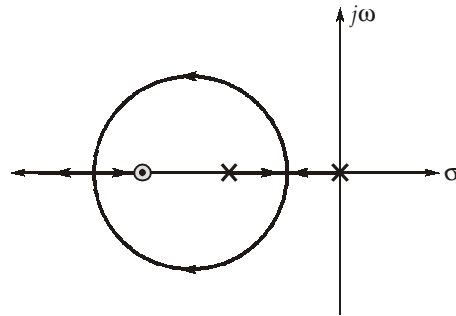
and the initial state value is $x(0) = [4 \ -4]^T$. The zero input response of the system will be

- (a) $\begin{bmatrix} 4e^{-4t} \\ -4e^{-3t} \end{bmatrix}$
- (b) $\begin{bmatrix} -4e^{-4t} \\ -4e^{-3t} \end{bmatrix}$
- (c) $\begin{bmatrix} 4e^{-4t} \\ 4e^{-3t} \end{bmatrix}$
- (d) $\begin{bmatrix} -4e^{-4t} \\ 4e^{-3t} \end{bmatrix}$

Q.29 The open-loop transfer function of a system is

$$G(s)H(s) = \frac{K(s+3)}{s(s+2)}$$

The root locus of the system consists of a circle as shown in the figure below. The equation of circle is



- | | |
|-------------------------------------|--|
| (a) $(\sigma + 4)^2 + \omega^2 = 4$ | (b) $(\sigma - 3)^2 + \omega^2 = 9$ |
| (c) $(\sigma + 3)^2 + \omega^2 = 3$ | (d) $(\sigma - 4)^2 + \omega^2 = \sqrt{4}$ |

Q.30 Consider a unity gain closed-loop transfer function with forward path gain,

$$G(s) = \frac{K(s+1)}{s^3 + 0.5s^2 + 3s + 1}$$

If the system is producing undamped oscillations, then value of K and corresponding frequency of oscillations are respectively

- | | |
|---------------------|-------------------|
| (a) 2.5 and 1 rad/s | (b) 1 and 2 rad/s |
| (c) 1 and 2.5 rad/s | (d) 2 and 1 rad/s |





MADE EASY

India's Best Institute for IES, GATE & PSUs

Delhi | Bhopal | Hyderabad | Jaipur | Pune | Bhubaneswar | Kolkata

Web: www.madeeasy.in | E-mail: info@madeeasy.in | Ph: 011-45124612

CONTROL SYSTEMS

EC | EE

Date of Test : 16/10/2023

ANSWER KEY >

- | | | | | |
|--------|---------|---------|---------|---------|
| 1. (a) | 7. (b) | 13. (d) | 19. (b) | 25. (b) |
| 2. (a) | 8. (b) | 14. (a) | 20. (a) | 26. (d) |
| 3. (b) | 9. (b) | 15. (c) | 21. (d) | 27. (c) |
| 4. (b) | 10. (c) | 16. (b) | 22. (b) | 28. (a) |
| 5. (a) | 11. (b) | 17. (a) | 23. (d) | 29. (c) |
| 6. (a) | 12. (d) | 18. (a) | 24. (b) | 30. (b) |

DETAILED EXPLANATIONS

1. (a)

After taking Laplace transform, the series combination of R_1 and C_1 gives: $R_1 + \frac{1}{sC_1}$.

Similarly, the parallel combination of R_2 and C_2 gives: $\frac{\frac{R_2}{sC_2}}{R_2 + \frac{1}{sC_2}}$

Substituting, $R_1 = R_2 = C_1 = C_2 = 1$ makes the above combinations $1 + \frac{1}{s}$ and $\frac{1}{1 + \frac{1}{s}}$

On applying the voltage divider rule, we get

$$V_0(s) = \frac{\frac{1}{s}}{1 + \frac{1}{s}} \cdot \frac{1}{1 + \frac{1}{s} + \frac{s}{1 + \frac{1}{s}}} V_i(s)$$

$$\frac{V_0(s)}{V_i(s)} = \frac{\frac{1}{s+1}}{\frac{s(s+1) + (s+1) + s}{s(s+1)}}$$

$$\frac{V_0(s)}{V_i(s)} = \frac{s}{s^2 + 3s + 1}$$

2. (a)

The general form of transfer function of a 2nd order system is given by,

$$\frac{Y(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\xi\omega_n s + \omega_n^2}$$

On comparing:

$$\omega_n^2 = 64$$

Natural frequency, $\omega_n = 8 \text{ rad/sec}$

$$2\xi\omega_n = 8$$

$$\text{Damping ratio, } \xi = \frac{8}{16} = 0.5$$

$$\omega_d = \omega_n \sqrt{1 - \xi^2} = 8\sqrt{1 - 0.5^2}$$

$$\omega_d = 6.93 \text{ rad/sec}$$

3. (b)

For impulse response, $R(s) = 1$

The impulse response is given by,

$$y(t) = L^{-1} \left\{ \frac{1}{(s+1)(s+10)} \right\}$$

$$y(t) = L^{-1} \left\{ \frac{1}{9(s+1)} - \frac{1}{9(s+10)} \right\}$$

$$y(t) = \frac{1}{9}e^{-t} - \frac{1}{9}e^{-10t}$$

4. (b)

Close loop transfer function of the system is

$$T(s) = \frac{G(s)}{1+G(s)} = \frac{K(s^2+1)}{(K+1)s^2+3s+K+2}$$

∴ the characteristic equation is given by

$$(K+1)s^2+3s+K+2=0$$

For the system to be stable all the coefficient should be of same sign

$$\therefore K+1 > 0 = K > -1$$

$$\text{and } K+2 > 0 = K > -2$$

∴ The system is stable for $K > -1$.

5. (a)

$$G(s)H(s) = \frac{(1-s)}{(1+s)(3+s)}$$

$$\text{So, } |G(j\omega)H(j\omega)| = \frac{\sqrt{1+\omega^2}}{\sqrt{1+\omega^2}\sqrt{9+\omega^2}} = \frac{1}{\sqrt{9+\omega^2}}$$

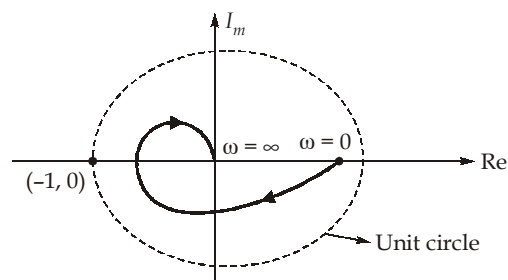
Phase angle is,

$$\phi = -2 \tan^{-1} \omega - \tan^{-1} \frac{\omega}{3}$$

$$\text{At } \omega = 0; \quad |G(j\omega)H(j\omega)| = \frac{1}{3}; \quad \phi = 0^\circ$$

$$\text{At } \omega = 4; \quad |G(j\omega)H(j\omega)| = \frac{1}{5}; \quad \phi = -205^\circ$$

$$\text{At } \omega = \infty; \quad |G(j\omega)H(j\omega)| = 0; \quad \phi = -270^\circ$$



Polar plot does not cross unit circle, for all values of ω (0 to ∞), magnitude will be less than unity. In this case, gain crossover frequency does not exist and phase margin becomes ∞ .

6. (a)

The gain of function is 1, given system is all pass system, Hence option (a) is correct answer.
 We can cross check:

$$GH(s) = \frac{1-s}{1+s}$$

$$GH(j\omega) = \frac{1-j\omega}{1+j\omega}$$

$$\angle GH(j\omega) = -\tan^{-1}(\omega) - \tan^{-1}(\omega) = -2 \tan^{-1}(\omega)$$

By varying $\omega: 0 \rightarrow \infty$

The phase varies from 0° to -180° as shown in the plot.

7. (b)

$$\text{OLTF is } G(s)H(s) = \frac{-1}{2s(1-20s)}$$

(or)
$$G(j\omega)H(j\omega) = \frac{-1}{2j\omega(1-20j\omega)}$$

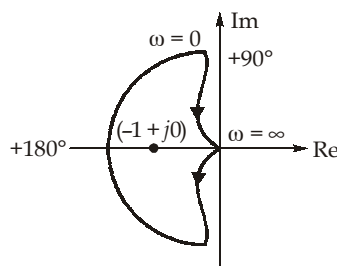
$$M = |G(j\omega)H(j\omega)| = \frac{1}{2\omega\sqrt{1+400\omega^2}}$$

$$\phi = 180^\circ - 90^\circ - \tan^{-1}\left(\frac{-20\omega}{1}\right)$$

$$= 90^\circ + \tan^{-1}(20\omega)$$

At $\omega = 0$;	$M = \infty$;	$\phi = 90^\circ$
At $\omega = 0.1$;	$M = 2.24$;	$\phi = 153.43^\circ$
At $\omega = \infty$;	$M = 0$;	$\phi = 180^\circ$

We get the Nyquist plot as,



8. (b)

The characteristic equation is,

$$1 + G(s)H(s) = 0$$

$$1 + \left(K_P + \frac{K_I}{s}\right) \left(\frac{1}{(s+1)(s+2)}\right) = 0$$

(or)
$$s^3 + 3s^2 + (2 + K_P)s + K_I = 0$$

Forming Routh Array:

$$\begin{array}{l}
 s^3 \quad \left| \begin{array}{cc} 1 & 2 + K_P \\ 3 & K_I \end{array} \right. \\
 s^2 \quad \left| \begin{array}{cc} 3 & K_I \end{array} \right. \\
 s^1 \quad \left| \begin{array}{cc} \frac{6 + 3K_P - K_I}{3} & \\ K_I & \end{array} \right. \\
 s^0 \quad \left| \begin{array}{cc} K_I & \end{array} \right.
 \end{array}$$

For stable system, $K_I > 0$

and $K_P > \frac{K_I}{3} - 2$

9. (b)

$$x(t) = \phi(t) x(0)$$

Where, $\phi(t)$ = state transition matrix

Given,
$$A = \begin{bmatrix} 1 & 0 \\ 1 & 1 \end{bmatrix}$$

$$[sI - A] = \begin{bmatrix} s & 0 \\ 0 & s \end{bmatrix} - \begin{bmatrix} 1 & 0 \\ 1 & 1 \end{bmatrix} = \begin{bmatrix} s-1 & 0 \\ -1 & s-1 \end{bmatrix}$$

$$[sI - A]^{-1} = \frac{1}{(s-1)(s-1)} \begin{bmatrix} s-1 & 0 \\ 1 & s-1 \end{bmatrix}$$

$$= \begin{bmatrix} \frac{1}{s-1} & 0 \\ \frac{1}{(s-1)^2} & \frac{1}{(s-1)} \end{bmatrix}$$

$$L^{-1}[sI - A]^{-1} = \phi(t) = \begin{bmatrix} e^t & 0 \\ te^t & e^t \end{bmatrix}$$

$$\therefore x(t) = \begin{bmatrix} e^t & 0 \\ te^t & e^t \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

$$x(t) = \begin{bmatrix} e^t \\ te^t \end{bmatrix}$$

10. (c)

The transfer function of the system is,

$$\begin{aligned}
 H(s) &= \frac{(s+a)}{[(s-(-a-jb))(s-(-a+jb))]^2} \\
 &= \frac{s+a}{[(s+a)^2 + b^2]^2} = \frac{1}{2} \cdot \frac{2(s+a)}{[(s+a)^2 + b^2]^2}
 \end{aligned}$$

$$H(s) = \frac{1}{2} \left[-\frac{dF(s)}{ds} \right]$$

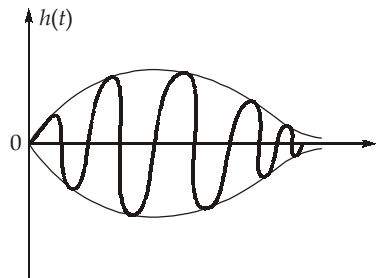
where,

$$F(s) = \frac{1}{(s+a)^2 + b^2} \xrightarrow{\text{I.L.T.}} f(t) = \frac{1}{b} e^{-at} \sin(bt)$$

$$-\frac{d}{ds} F(s) \longleftrightarrow t f(t)$$

$$h(t) = \frac{t}{2} f(t) = \frac{t}{2b} e^{-at} \sin(bt)$$

$$h(t) = K t e^{-at} \sin(bt)$$

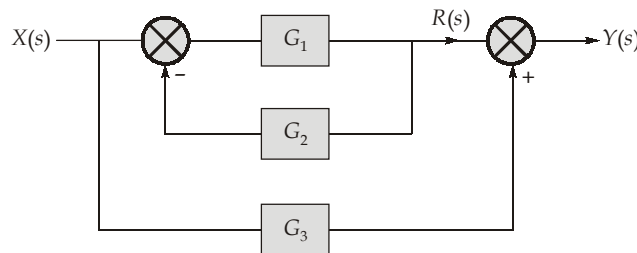


11. (b)

For given block diagram.

In inner loop feedback we can write,

$$\frac{R(s)}{X(s)} = \frac{G_1(s)}{1 + G_1(s)G_2(s)}$$



∴ Overall transfer function,

$$Y(s) = X(s) R(s) + G_3(s) X(s)$$

$$\frac{Y(s)}{X(s)} = R(s) + G_3(s)$$

$$= \frac{1}{s+1} = \frac{(s+4)}{(s+1)(s+4)+1} = \frac{s+4}{s^2+5s+4+1}$$

$$\frac{s+4}{s^2+5s+5} + \frac{s+2}{s+5} \Rightarrow \frac{(s+4)(s+5) + (s+2)(s^2+5s+5)}{(s^2+5s+5)(s+5)}$$

$$= \frac{s^2+9s+20+s^3+5s^2+5s+2s^2+10s+10}{s^3+5s^2+5s^2+25s+5s+25}$$

$$\frac{s^3+8s^2+24s+30}{s^3+10s^2+30s+25}$$

12. (d)

$$\frac{d^2y}{dt^2} + 5\frac{dy}{dt} + 4y = x(t)$$

$$s^2Y(s) + 5sY(s) + 4Y(s) = X(s)$$

$$x(t) = 8u(t) \quad \text{or} \quad X(s) = \frac{8}{s}$$

$$(s^2 + 5s + 4)Y(s) = \frac{8}{s}$$

$$Y(s) = \frac{8}{s(s^2 + 5s + 4)} = \frac{A}{s} + \frac{B}{s+4} + \frac{C}{s+1}$$

$$Y(s) = \frac{2}{s} - \frac{8}{3(s+1)} + \frac{2}{3(s+4)}$$

$$y(t) = 2\left[1 + \frac{1}{3}e^{-4t} - \frac{4}{3}e^{-t}\right]u(t)$$

13. (d)

$$\frac{Y(s)}{R(s)} = \frac{K/s}{1 - \left(\frac{3}{s} + \frac{K}{s}\right)} = \frac{K}{s - (3+K)}$$

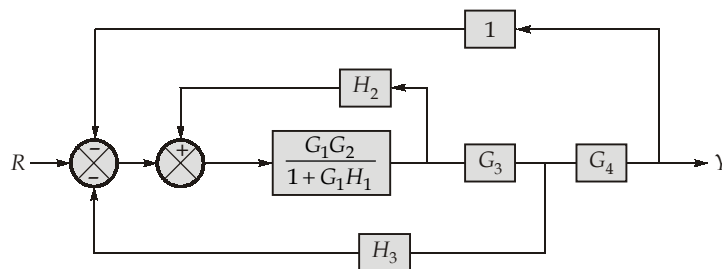
For system to be stable,

$$3 + K < 0$$

$$K < -3$$

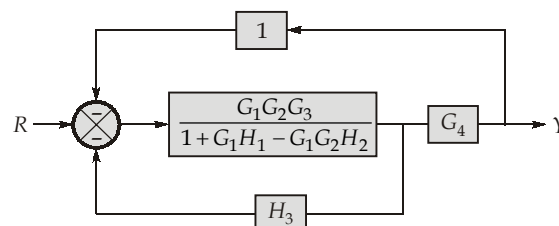
14. (a)

G_1 and H_1 form a negative feedback loop and that loop is in cascade with G_2 . The block diagram can be simplified as:

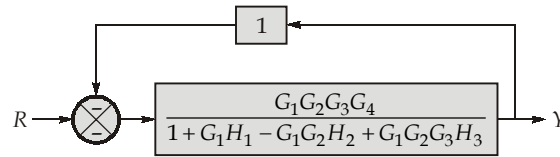


Now $\frac{G_1G_2}{1+G_1H_1}$ forms a positive feedback loop with H_2 and that loop is in cascade with G_3 . The

block diagram can be further simplified as:



Now there is another negative feedback loop with H_3 and cascade with G_4 . Block diagram becomes



$$\frac{Y(s)}{R(s)} = \frac{G_1G_2G_3G_4}{1 + G_1H_1 - G_1G_2H_2 + G_1G_2G_3H_3 + G_1G_2G_3G_4}$$

15. (c)

The Routh table for the characteristic equation corresponding to the given transfer function is:

s^5	1	0	4
s^4	3	5	3
s^3	-1.67	3	
s^2	10.4	3	
s^1	3.48		
s^0	3		

There are two sign changes in the first column of Routh table. So there are 2 poles in RHP. Order of the characteristic equation is 5 = number of poles

∴ No. of poles in LHP = 5 - 2 = 3

16. (b)

Closed loop transfer function is,

$$T(s) = \frac{G(s)}{1 + G(s)} = \frac{K(s+3)(s+5)}{(K+1)s^2 + (8K-6)s + (15K+8)}$$

For stability all coefficients should be positive or all negative

Case-I: All coefficient are positive,

$$K + 1 > 0 \Rightarrow K > -1$$

$$8K - 6 > 0 \Rightarrow K > 0.75$$

$$15K + 8 > 0 \Rightarrow K > \frac{-8}{15}$$

∴ $K > 0.75$

Case-II : All coefficients are negative:

$$K + 1 < 0 \Rightarrow K < -1$$

$$8K - 6 < 0 \Rightarrow K < 0.75$$

$$15K + 8 < 0 \Rightarrow K < \frac{-8}{15}$$

∴ $K < -1$

So, for stability either $K > 0.75$ or $K < -1$

17. (a)

Transfer function is,

$$G(s)H(s) = \frac{Ks}{\left(\frac{s}{4}+1\right)\left(\frac{s}{10}+1\right)^2} = \frac{400Ks}{(s+4)(s+10)^2}$$

Magnitude of initial plot is given by,

$$M = 20 \log \omega + 20 \log K$$

and

$$M = 0 \text{ dB at } \omega = 4$$

$$0 = 20 \log_{10} 4 + 20 \log K$$

$$(or) \quad 20 \log \frac{1}{K} = 20 \log 4$$

$$(or) \quad \frac{1}{K} = 4$$

$$(or) \quad K = \frac{1}{4}$$

$$Hence, \quad G(s)H(s) = \frac{100s}{(s+4)(s+10)^2}$$

18. (a)

$$\text{Damped frequency,} \quad \omega_d = \omega_n \sqrt{1-\xi^2}$$

$$\Rightarrow \quad \omega_n = \frac{\omega_d}{\sqrt{1-\xi^2}} = \frac{10}{\sqrt{1-(0.6)^2}}$$

The desired second order C.E. is,

$$s^2 + 2\xi\omega_n s + \omega_n^2 = 0$$

$$s^2 + 2(0.6)(12.5)s + (12.5)^2 = 0$$

$$s^2 + 15s + 156.25 = 0$$

The C.E. of given system is,

$$1 + (K_p + sK_D) \left(\frac{1}{s(s+1)} \right) = 0$$

$$(or) \quad s^2 + (1 + K_D)s + K_p = 0$$

$$\text{Hence,} \quad K_p = 156.25$$

$$\text{and} \quad 1 + K_p = 15$$

$$\Rightarrow \quad K_D = 14$$

19. (b)

From the above Bode plot,

For section *de*, slope is -20 dB/dec

$$\therefore \quad -20 = \frac{y-0}{\log 8 - \log 16}$$

$$y = 6.02 \text{ dB}$$

Now, for section *bc*, slope is -20 dB/dec

$$\therefore -20 = \frac{16 - 6.02}{\log \omega_1 - \log 4}$$

$$\omega_1 = 1.268 \text{ rad/sec}$$

To find value of gain K

$$y = mx + c$$

$$16 = -40 \log 1.268 + 20 \log K$$

$$K = 10.14$$

From all the result, transfer function is,

$$T(s) = \frac{10.14 \left(\frac{s}{1.268} + 1 \right) \left(\frac{s}{4} + 1 \right)}{s^2 \left(\frac{s}{8} + 1 \right)}$$

$$T(s) = \frac{16(s + 1.268)(s + 4)}{s^2(s + 8)}$$

20. (a)

Given,
$$G(s)H(s) = \frac{1}{s^2(s-a)(s-b)(s-c)}$$

Put $s = j\omega$,
$$G(j\omega)H(j\omega) = \frac{1}{(j\omega)^2(j\omega-a)(j\omega-b)(j\omega-c)}$$

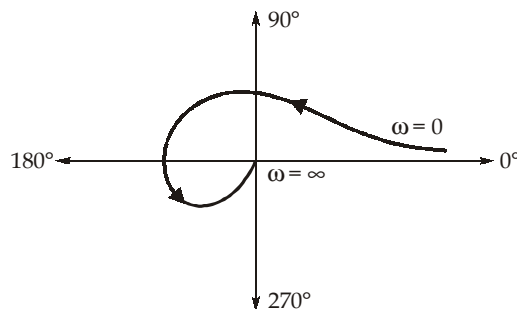
$$|G(j\omega)H(j\omega)| = \frac{1}{\omega^2 \sqrt{\omega^2 + a^2} \cdot \sqrt{\omega^2 + b^2} \cdot \sqrt{\omega^2 + c^2}}$$

when, $\omega = 0$ Magnitude = ∞ ; Phase = 0°
 $\omega = \infty$ Magnitude = 0 ; Phase = 270°

$$\angle G(j\omega)H(j\omega) = - \left[180^\circ + \left(180^\circ - \tan^{-1} \left(\frac{\omega}{a} \right) \right) + \left(180^\circ - \tan^{-1} \left(\frac{\omega}{b} \right) \right) + \left(180^\circ - \tan^{-1} \left(\frac{\omega}{c} \right) \right) \right]$$

$$\phi = \tan^{-1} \left(\frac{\omega}{a} \right) + \tan^{-1} \left(\frac{\omega}{b} \right) + \tan^{-1} \left(\frac{\omega}{c} \right)$$

Note: Since a, b and c are positive values, for $\omega = 1$, phase of $G(j\omega)H(j\omega)$ will be a positive quantity. So, the polar plot will start into the first quadrant.



21. (d)

The transfer function from steady state model can be written as

$$T(s) = C[sI - A]^{-1} B$$

$$[sI - A]^{-1} = \frac{\text{adj}[sI - A]}{|sI - A|}$$

$$[sI - A] = \begin{bmatrix} s+5 & -1 \\ 0 & s+4 \end{bmatrix}$$

$$[sI - A]^{-1} = \frac{1}{(s+5)(s+4)} \begin{bmatrix} s+4 & 1 \\ 0 & s+5 \end{bmatrix}$$

$$T(s) = \frac{1}{(s+5)(s+4)} \begin{bmatrix} 1 & 4 \end{bmatrix} \begin{bmatrix} s+4 & 1 \\ 0 & s+5 \end{bmatrix} \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

$$T(s) = \frac{1}{(s+5)(s+4)} \begin{bmatrix} 1 & 4 \end{bmatrix} \begin{bmatrix} 1 \\ s+5 \end{bmatrix}$$

$$T(s) = \frac{1+4(s+5)}{(s+4)(s+5)} = \frac{4s+21}{(s+4)(s+5)}$$

If two systems are connected in parallel

Overall transfer function,

$$T'(s) = T(s) + T(s) = \frac{2(4s+21)}{(s+4)(s+5)}$$

22. (b)

For the given system we can write,

$$Y(s) [s^3 + 4s^2 + 5s + 3] = U(s)[1] \quad \dots(\text{i})$$

Let, $x_1 = y$

So, $x_2 = \dot{x}_1 = \dot{y} \quad \dots(\text{ii})$

$$x_3 = \dot{x}_2 = \ddot{y} \quad \dots(\text{iii})$$

The equation (i) can be written as

$$\dot{x}_3 + 4x_3 + 5x_2 + 3x_1 = u$$

i.e. $\dot{x}_3 = -3x_1 - 5x_2 - 4x_3 + u \quad \dots(\text{iv})$

Using equation (ii), (iii) and (iv) we can write in state variable form as below,

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -3 & -5 & -4 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} u$$

So, $[A] = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -3 & -5 & -4 \end{bmatrix}$

and $[B] = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$

Hence option (b) is correct.

23. (d)

The closed-loop transfer function for given system,

$$\begin{aligned} \frac{G(s)}{1+G(s)} &= \frac{\frac{3(1+sK)}{s(s+4)}}{1+\frac{3(1+sK)}{s(s+4)}} = \frac{3(1+sK)}{s(s+4)+3(1+sK)} \\ &= \frac{3(1+sK)}{s^2+4s+3(1+sK)} = \frac{3+3sK}{s^2+(4+3K)s+3} \end{aligned}$$

Comparing the transfer function with standard second order transfer function.

$$\omega_n^2 = 3$$

⇒

$$\omega_n = \sqrt{3} \text{ rad/sec}$$

$$2\xi\omega_n = 4 + 3K$$

$$2\sqrt{3}\xi = 3K + 4$$

$$\frac{2\sqrt{3}\xi - 4}{3} = K$$

Given, $\xi = \frac{1}{\sqrt{3}}$,

$$K = \frac{6-4}{3} = \frac{2}{3}$$

24. (b)

From plot -1

Number of poles on the right side of s-plane = 0

$$P = 0$$

Open-loop system is stable.

From plot -2

Number of encirclement about (-1, 0) is = 2 in clockwise

$$N = -2$$

$$N = P - Z$$

$$-2 = 0 - Z$$

$$Z = 2$$

Two closed loop poles on the right side of s-plane. Close-loop system is unstable.

25. (b)

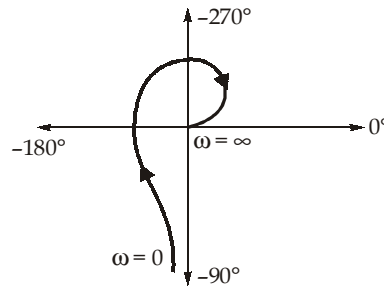
From the figure, the transfer function of the system is,

$$G(s) H(s) = \frac{K}{s(s+2)^2(s+4)}$$

$$|G(j\omega)H(j\omega)| = \frac{K}{\omega(4+\omega^2)\sqrt{16+\omega^2}}$$

$$\angle G(j\omega)H(j\omega) = -90^\circ - 2 \tan^{-1}\left(\frac{\omega}{2}\right) - \tan^{-1}\left(\frac{\omega}{4}\right)$$

ω	M	ϕ
0	∞	-90°
\vdots	\vdots	\vdots
∞	0	-360°



26. (d)

$$\text{T.F.} = \frac{K\left(\frac{s}{2} + 1\right)}{s\left(\frac{s}{5} + 1\right)\left(\frac{s}{10} + 1\right)}$$

To obtain the value of K ,

$$30 \text{ dB} = 20 \log K - 20 r \log(1) \quad [\because r = 1]$$

$$1.5 = \log K$$

$$K = 10^{1.5} = 31.62$$

$$\text{T.F.} = \frac{31.62(0.5s + 1)}{s(0.2s + 1)(0.1s + 1)}$$

27. (c)

$$\angle G(j\omega)H(j\omega) = \angle \frac{10}{(j\omega + 1)(j\omega + 20)} = -180^\circ$$

$$= -\tan^{-1}(\omega) - \tan^{-1}\left(\frac{\omega}{20}\right) = -180^\circ$$

$$-\tan^{-1}\left(\frac{\omega + \frac{\omega}{20}}{1 - \frac{\omega^2}{20}}\right) = -180^\circ$$

$$\frac{\omega + \frac{\omega}{20}}{1 - \frac{\omega^2}{20}} = 0$$

The above expression to become zero $1 - \frac{\omega^2}{20} = \infty$.

$$\therefore \omega_{pc} = \infty$$

$$|G(j\omega)H(j\omega)| = \frac{10}{\sqrt{1+\omega^2}\sqrt{(20)^2+\omega^2}}$$

$$|G(j\omega)H(j\omega)|_{\omega_{pc}=\infty} = 0$$

$$\text{Gain margin} = \frac{1}{|G(j\omega)H(j\omega)|_{\omega_{pc}}} = \infty$$

\therefore Gain crossover frequency does not exist.

\therefore Phase margin is $= \infty$

28. (a)

For given state model,

$$A = \begin{bmatrix} -4 & 0 \\ 0 & -3 \end{bmatrix}$$

$$[sI - A] = \begin{bmatrix} s & 0 \\ 0 & s \end{bmatrix} - \begin{bmatrix} -4 & 0 \\ 0 & -3 \end{bmatrix} = \begin{bmatrix} s+4 & 0 \\ 0 & s+3 \end{bmatrix}$$

$$[sI - A]^{-1} = \frac{1}{|sI - A|} \text{Adj}[sI - A] = \frac{1}{(s+3)(s+4)} \begin{bmatrix} s+3 & 0 \\ 0 & s+4 \end{bmatrix}$$

$$= \begin{bmatrix} \frac{1}{s+4} & 0 \\ 0 & \frac{1}{s+3} \end{bmatrix}$$

$$e^{At} = \begin{bmatrix} e^{-4t} & 0 \\ 0 & e^{-3t} \end{bmatrix}$$

So zero input response of the given system will be

$$x(t) = e^{At} \cdot x(0)$$

$$= \begin{bmatrix} e^{-4t} & 0 \\ 0 & e^{-3t} \end{bmatrix} \begin{bmatrix} 4 \\ -4 \end{bmatrix} = \begin{bmatrix} 4e^{-4t} \\ -4e^{-3t} \end{bmatrix}$$

29. (c)

Method-I

For root locus point,

$$\angle G(s) H(s) = 180^\circ$$

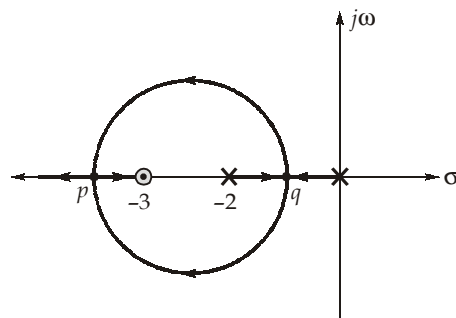
 \therefore Substituting, $s = (\sigma + j\omega)$ we get,

$$G(\sigma + j\omega) H(\sigma + j\omega) = \frac{K(\sigma + 3 + j\omega)}{(\sigma + j\omega)(\sigma + 2 + j\omega)}$$

$$\tan^{-1}\left(\frac{\omega}{\sigma + 3}\right) - \tan^{-1}\left(\frac{\omega}{\sigma}\right) = 180^\circ + \tan^{-1}\left(\frac{\omega}{\sigma + 2}\right)$$

$$\frac{\frac{\omega}{\sigma + 3} - \frac{\omega}{\sigma}}{1 + \left(\frac{\omega}{\sigma + 3}\right)\left(\frac{\omega}{\sigma}\right)} = \frac{\omega}{\sigma + 2}$$

$$\therefore (\sigma + 3)^2 + \omega^2 = 3$$

Method-II:

p and q are breakaway and breakin points, to obtain them we have to perform, $\frac{dK}{ds} = 0$.

$$K = -\frac{s(s+2)}{s+3}$$

$$\frac{dK}{ds} = \frac{d}{ds} \left[-\frac{s(s+2)}{s+3} \right] = 0$$

$$s^2 + 6s + 6 = 0$$

$$q = -1.268 \quad \text{to} \quad p = -4.732$$

As we know two points on the diameter, center of the circle is $(-3, 0)$ and radius is 1.732.

Equation of the circle is $(\sigma + 3)^2 + \omega^2 = (\sqrt{3})^2$.

30. (b)

The characteristic equation is,

$$q(s) = s^3 + 0.5s^2 + (K + 3)s + (K + 1) = 0$$

$$\begin{array}{l|ll} s^3 & 1 & K + 3 \\ s^2 & 0.5 & K + 1 \\ s^1 & (3 + K) - 2(K + 1) & 0 \\ s^0 & (K + 1) & \end{array}$$

For a system to oscillate a row should become zero.

$$\therefore K + 3 - 2K - 2 = 0$$

$$K = 1$$

Given system is third order system $(s + a)(s^2 + bs + c) = 0$

For a marginally stable system, $\xi = 0$

$$s^2 + 2\xi\omega_n s + \omega_n^2 = 0$$

$$s^2 + \omega_n^2 = 0$$

Take the coefficients of s^2 row.

$$0.5s^2 + (K + 1) = 0$$

$$0.5s^2 + 2 = 0$$

$$s = \pm j2$$

$$\omega = 2 \text{ rad/s}$$

